

September 2003

Features

- High Output Drive
 - 18.8 Vpp differential output voltage, $R_L = 50\Omega$
 - 9.4 Vpp single-ended output voltage, $R_L = 25\Omega$
- High Output Current
 - $\pm 200\text{mA}$ @ $V_o = 9.4\text{ Vpp}$, $V_s = 12\text{V}$
- Low Distortion
 - 85dB SFDR (Spurious Free Dynamic Range) @ 100KHz, $V_o = 2\text{Vpp}$, $R_L = 25\Omega$
- High Speed
 - 192MHz 3dB bandwidth ($G=2$)
 - 240V / μs slew rate
- Low Noise
 - 3.8nV / $\sqrt{\text{Hz}}$: input noise voltage
 - 2.7pA / $\sqrt{\text{Hz}}$: input noise current
- Low supply current: 7mA/amp
 - Single-supply operation: 5V to 12V

Ordering Information

ZL40167/DCA (tubes) 8 lead SOIC
 ZL40167/DCB (tape and reel) 8 lead SOIC
-40°C to +85°C

- High ESD (Electro-Static Discharge) immunity
 - 4kV for Supply and Output pins
- Low differential gain and phase
 - 0.005% and -0.07deg

Applications

- ADSL PCI modem cards
- xDSL external modem
- Line Driver

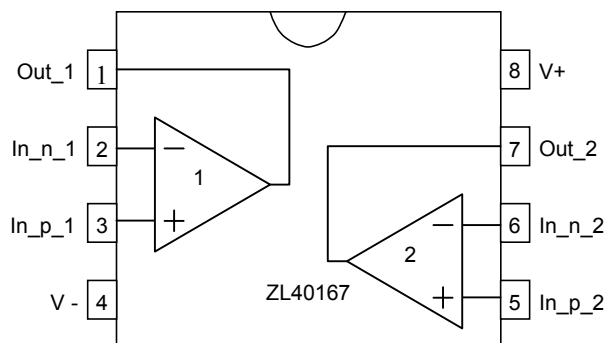


Figure 1 - Functional Block Diagram and Pin Connection

Description

The ZL40167 is a low cost voltage feedback opamp capable of driving signals to within 1V of the power supply rails. It features low noise and low distortion accompanied by a high output current which makes it ideally suited for the application as an xDSL line driver. The dual opamp can be connected as a differential line driver delivering signals up to 18.8Vpp swing into a 25Ω load, fully supporting the peak upstream power levels for upstream full-rate ADSL (Asymmetrical Digital Subscriber Line).

The wide bandwidth, high power output and low differential gain and phase figures make the ZL40167 ideally suited for a wide variety of video driver applications.

Application Notes

The ZL40167 is a high speed, high output current, dual operational amplifier with a high slew rate and low distortion. The device uses conventional voltage feedback for ease of use and more flexibility. These characteristics make the ZL40167 ideal for applications where driving low impedances of 25 to 100Ω such as xDSL and active filters.

The figure below shows a typical ADSL application utilising a 1:2 transformer, the feedback path provides a Gain = +2.

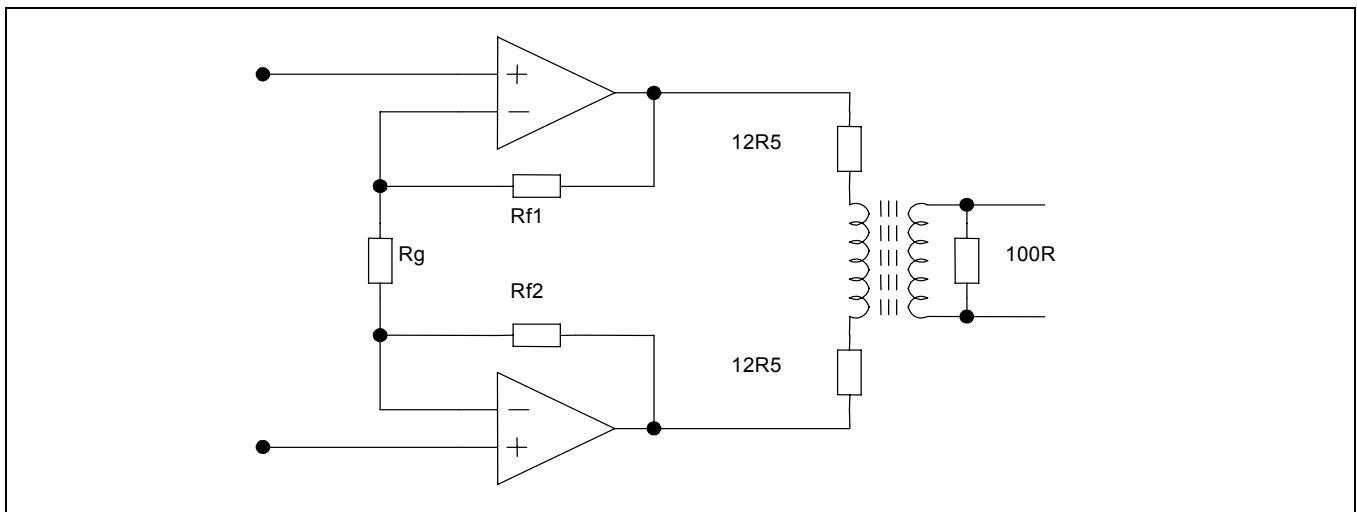


Figure 2 - A Typical ADSL Application

A class AB output stage allows the ZL40167 to deliver high currents to low impedance loads with low distortion while consuming low quiescent current.

Note: the high ESD immunity figure of 4kV may mean that in some designs fewer additional EMC protection components are needed thus reducing total system costs.

The ZL40167 is not limited to ADSL applications and can be used as a general purpose opamp configured with either inverting or non-inverting feedback. The figure below shows non-inverting feedback arrangement that has typically been used to obtain the data sheet specifications.

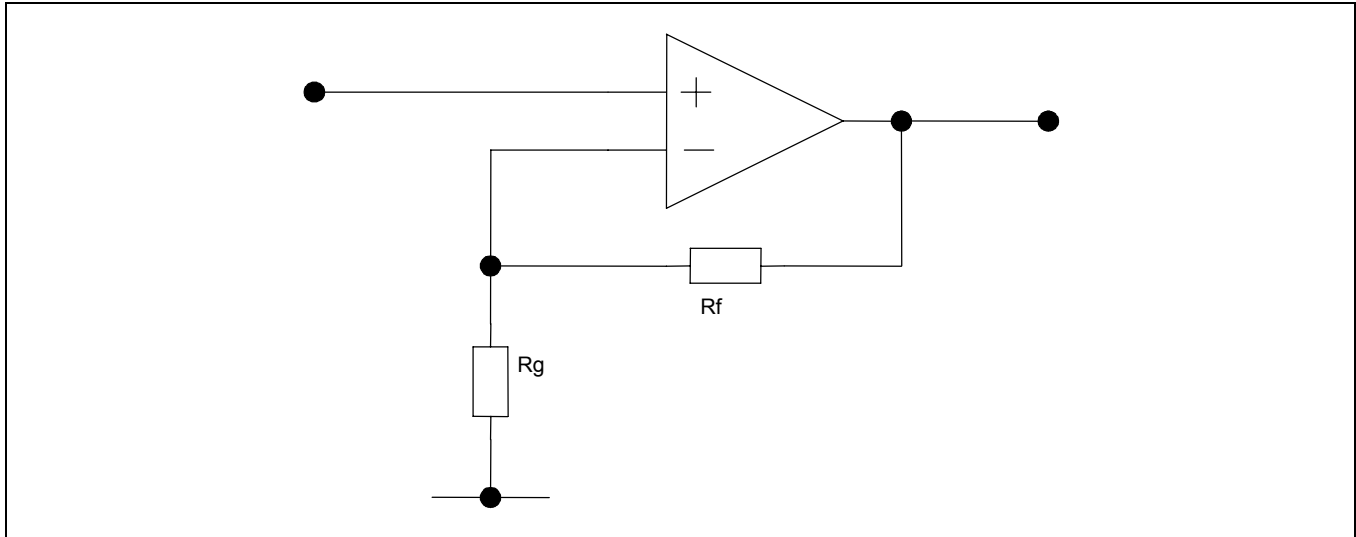


Figure 3 - A Non-Inverting Feedback Amplifier Example

Video transmitter and receiver for twisted wire pair

Composite video signals can be transmitted down twisted pair cable, i.e. Ethernet (CAT 5), using a differential transmitter and receiver. The transmitter must be able to drive high currents into the low impedance twisted pair cable. For video, the amplifiers require flat gain and low phase-shift over the video signal band. To ensure this, the amplifiers will have 3dB bandwidths well in excess of this. The ZL40167 (dual amplifier) has all of these attributes.

With reference to the differential video driver shown in Figure , the input coax is assumed to have a characteristic impedance of 75 Ohms, this is terminated with a parallel combination of 110 Ohms and the input impedance of amplifier IC1 (b) of 255 Ohms, giving 77 Ohms. Low values of feedback resistors are used around the op-amps to reduce phase-shift due to parasitic capacitors and to minimise the addition of noise.

Baseband PAL or NTSC video signals generally have an amplitude of 2V pk-pk. A gain of two is used to ensure that the signal level at the end of the (terminated with 100 Ohms) differential pair will be the same as the input level, neglecting any losses due to the use of long cable lengths.

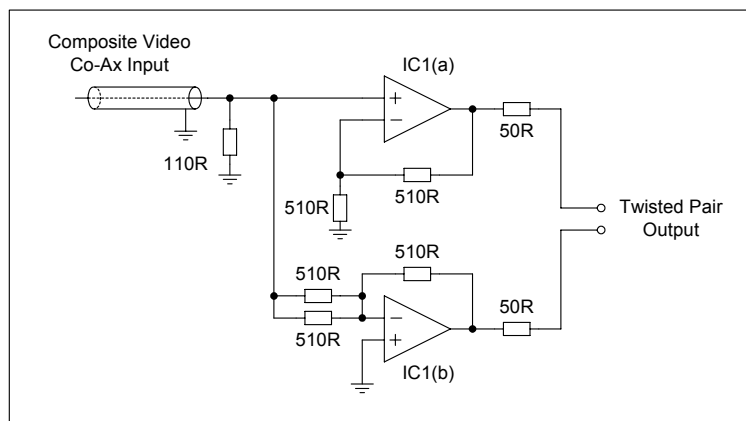


Figure 4 - Differential Video Driver

The differential receiver is shown in Figure 5 has a 100 Ohm line termination resistor, followed by a differential amplifier. Long cables will tend to attenuate the signal with greater losses at the higher frequencies, so the second amplifier is used to equalise these losses. Initially the amplifier should be built without fitting components R1 and

C1. Select the value of R2 to give the required gain at low frequency. Adjust the values of R1 and C1 to correct for the frequency dependant attenuation of the cable.

To drive a coax cable the output of the amplifier is connected via a series matching 75 Ohm resistor, again this second (dual amplifier) ZL40167 provides the required power output for the restored 2Vpk-pk video signal.

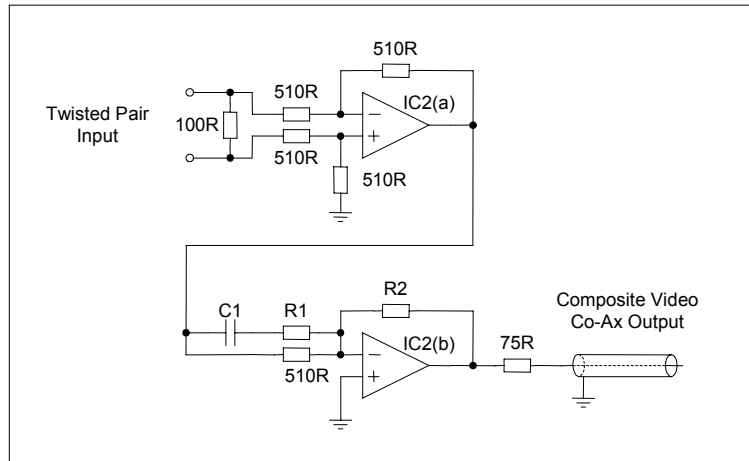


Figure 5 - Differential Video Receiver

Absolute Maximum Ratings - (See Note 1)

Parameter	Symbol	Min	Max	Units
Vin Differential	V_{IN}		± 1.2	V
Output Short Circuit Protection	$V_{OS/C}$		See Apps Note in this data sheet	
Supply Voltage	V_{+}, V_{-}		± 13.2	V
Voltage at Input Pins	$V_{(+IN)}, V_{(-IN)}$	(V-) -0.8	(V+) +0.8	V
Voltage at Output Pins	V_O		± 5.5	V
ESD Protection (HBM Human Body Model) (See Note 2)		4	(Note 3)	kV
Storage Temperature		-55	+150	°C
Latch-up test		+/-100mA for 100ms	(Note 4)	
Supply transient test		20% pulse for 100ms	(Note 5)	

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Human body model, 1.5k Ω in series with 100pF. Machine model, 200 Ω in series with 100pF.

Note 3: 1.25kV between the pairs of +INA, -INA and +INB, -INB pins only. 4kV between supply pins, OUTA or OUTB pins and any input pin.

Note 4: +/-100mA applied to input and output pins to force the device to go into "latch-up". The device passes this test to JEDEC spec 17.

Note 5: Positive and Negative supply transient testing increases the supplies by 20% for 100ms.

Operating Ratings - (See Note 1)

Parameter	Symbol	Min	Max	Units
Supply Voltage	V_{+}, V_{-}	± 2.5	± 6.5	V
Junction Temperature Range		-40	150	°C
Junction to Ambient Resistance	$R_{th(j-a)}$	150		°C 4 layer FR5 board
Junction to Case Resistance	$R_{th(j-c)}$	60		°C 4 layer FR5 board

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Electrical Characteristics - TA = 25°C, G = +2, Vs = ± 6V, Rf = Rg = 510Ω, RL = 100Ω / 2pF; Unless otherwise specified.

Symbol	Parameter	Conditions	Min (Note 1)	Typ (Note 2)	Max (Note 3)	Units	Test Type
Dynamic Performance							
	-3dB Bandwidth	Vo = 200mVp-p		192		MHz	C
	-0.1dB Bandwidth	Vo = 200mVp-p		32		MHz	C
	Slew Rate	4V Step O/P, 10-90%		240		V/μs	C
	Rise and Fall Time	4V Step O/P, 10-90%		13.3		ns	C
	Rise and Fall Time	200mV Step O/P, 10-90%		1.7		ns	C
	Differential Gain	NTSC, RL = 150Ω		0.005		%	C
	Differential Phase	NTSC, RL = 150Ω		-0.07		deg	C
Distortion and Noise Response							
	2 nd Harmonic Distortion	Vo = 8.4Vpp, f = 100KHz, RL = 25Ω/2pF		-65.4		dBc	C
		Vo = 8.4Vpp, f = 1MHz, RL = 100Ω/2pF		-83.8		dBc	C
		Vo = 2Vpp, f = 100kHz, RL = 25Ω/2pF		-93.6		dBc	C
		Vo = 2Vpp, f = 1MHz, RL = 100Ω/2pF		-86		dBc	C
	3 rd Harmonic Distortion	Vo = 8.4Vpp, f = 100KHz, RL = 25Ω/2pF		-70		dBc	C
		Vo = 8.4Vpp, f = 1MHz, RL = 100Ω/2pF		-77.7		dBc	C
		Vo = 2Vpp, f = 100KHz, RL = 25Ω/2pF		-85		dBc	C
		Vo = 2Vpp, f = 1MHz, RL = 100Ω/2pF		-73.5		dBc	C
MTPR	Multi-Tone Power Ratio	47.4375 KHz		-75		dBc	C
		69 KHz		-76.3		dBc	C
		90.5625 KHz		-73.8		dBc	C
		112.125 KHz		-71.5		dBc	C
	Input Noise Voltage	f = 100KHz		3.85		nV/√Hz	C
	Input Noise Current	f = 100KHz		2.7		pA/√Hz	C

Symbol	Parameter	Conditions	Min (Note 1)	Typ (Note 2)	Max (Note 3)	Units	Test Type
Input Characteristics							
Vos	Input Offset Voltage	Tj = -40°C to 150°C	- 4.2	- 0.3	4.2	mV	A
Ib	Input Bias Current	Tj = -40°C to 150°C		-10	-20	μA	A
Ios	Input Offset Current	Tj = -40°C to 150°C	-2	-0.2	2	μA	A
CMVR	Common Mode Voltage Range	Tj = -40°C to 150°C	- 4.9		4.9	V	A
CMRR	Common Mode Rejection Ratio	Tj = -40°C to 150°C	70	79		dB	A
Transfer Characteristics							
Avol	Voltage Gain	RL = 1k, Tj = -40°C to 150°C	4.7	10		V/mV	A
		RL = 25Ω, Tj = -40°C to 150°C	1.6	5.5			A
	Output Swing	RL = 25Ω, Tj = -40°C to 150°C	- 4.5	± 4.7	4.5	V	A
	Output Swing	RL = 1k, Tj = -40°C to 150°C	- 5	± 5.1	5	V	A
Isc	Output Current (Note 3)	Vo = 0, Tj = -40°C to 150°C	570	1000		mA	B
Power Supply							
Is	Supply Current / Amp	Tj = -40°C to 150°C		7	9	mA	A
PSRR	Power Supply Rejection Ratio	Tj = -40°C to 150°C	73	81		dB	A

Note 1: The maximum power dissipation is a function of Tj(max), θJA and TA. The maximum allowable power dissipation at any ambient temperature is $PD = (Tj(max) - TA) / \theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

Note 2: Typical values represent the most likely parametric norm.

Note 3: Test Types:

- 100% tested at 25°C. Over temperature limits are set by characterisation, simulation and statistical analysis.
- Limits set by characterisation, simulation and statistical analysis.
- Typical value only for information.

± 2.5V Electrical Characteristics - TA = 25°C, G = +2, Vs = ± 2.5V, Rf = Rg = 510Ω, RL = 100Ω / 2pF; Unless otherwise specified.

Symbol	Parameter	Conditions	Min (Note 1)	Typ (Note 2)	Max (Note 3)	Units	Test Type
Dynamic Performance							
	-3dB Bandwidth			176.5		MHz	C
	-0.1dB Bandwidth			83.8		MHz	C
	Slew Rate	1V Step O/P, 10-90%		216		V/μs	C
	Rise and Fall Time	1V Step O/P, 10-90%		3.7		ns	C
	Rise and Fall Time	200mV Step O/P, 10-90%		1.7		ns	C
Distortion and Noise Response							
	2 nd Harmonic Distortion	Vo = 2Vpp,f = 100KHz, RL = 25Ω		-92.6		dBc	C
		Vo = 2Vpp, f = 1MHz, RL = 100Ω		-85		dBc	C
	3 rd Harmonic Distortion	Vo = 2Vpp, f = 100KHz, RL = 25Ω		-86.3		dBc	C
		Vo = 2Vpp, f = 1MHz, RL = 100Ω		-74.8		dBc	C
Input Characteristics							
Vos	Input Offset Voltage	Tj = -40°C to 150°C	- 4.2	- 0.3	4.2	mV	B
Ib	Input Bias Current	Tj = -40°C to 150°C		- 10	-20	μA	B
CMVR	Common Mode Voltage Range		-1.55		1.55	V	B
CMRR	Common Mode Rejection Ratio	Tj = -40°C to 150°C	70	80		dB	B
Transfer Characteristics							
Avol	Voltage Gain	RL = 1k, Tj = -40°C to 150°C	5.5	10.5		V/mV	B
		RL = 25Ω, Tj = -40°C to 150°C	1.6	5.8			B
Output Characteristics							
	Output Swing	RL = 25Ω, Tj = -40°C to 150°C	-1.4	±1.45	1.4	V	B
		RL = 1k, Tj = -40°C to 150°C	-1.6	±1.65	1.6		B

Symbol	Parameter	Conditions	Min (Note 1)	Typ (Note 2)	Max (Note 3)	Units	Test Type
Power Supply							
Is	Supply Current/Amp	Tj = -40°C to 150°C		6.75	8.5	mA	A
PSRR	Power Supply Rejection Ratio	Tj = -40°C to 150°C	73	83		dB	B

Note 1: The maximum power dissipation is a function of Tj(max), θ_{JA} and TA. The maximum allowable power dissipation at any ambient temperature is $PD = (Tj(max) - TA) / \theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

Note 2: Typical values represent the most likely parametric norm.

Note 3: Test Types:

- 100% tested at 25°C. Over temperature limits are set by characterisation, simulation and statistical analysis.
- Limits set by characterisation, simulation and statistical analysis.
- Typical value only for information.

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, gain = +2, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

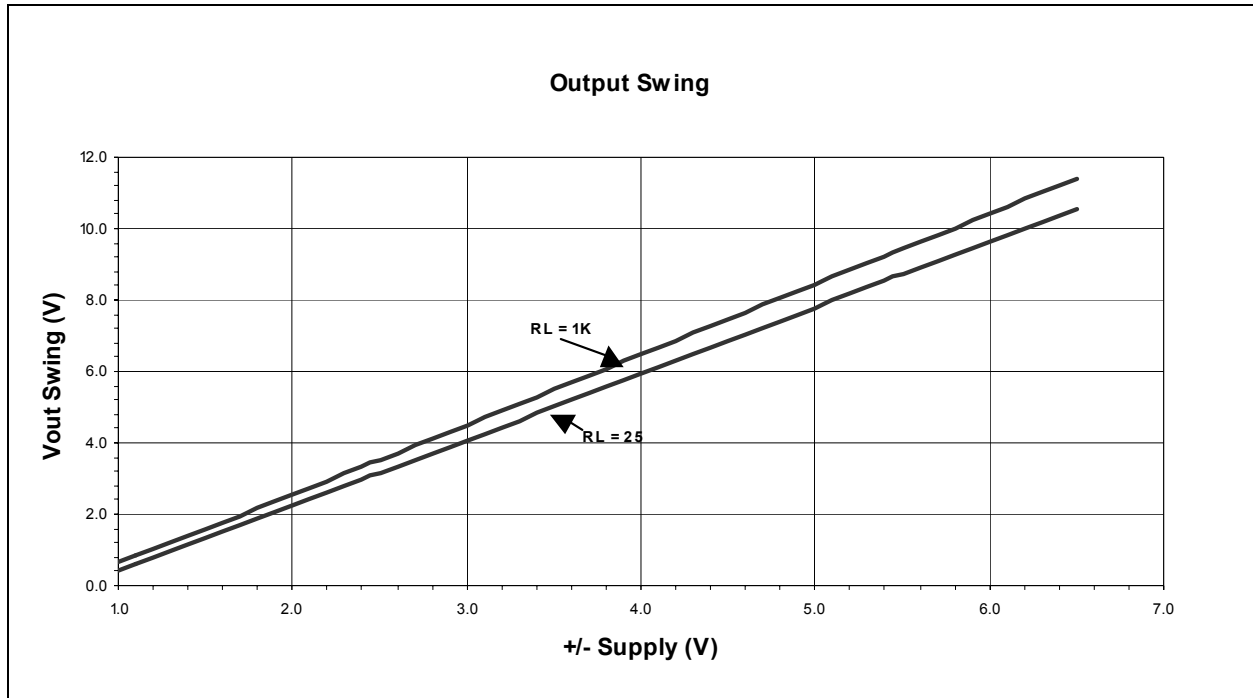


Figure 6 - Output Swing

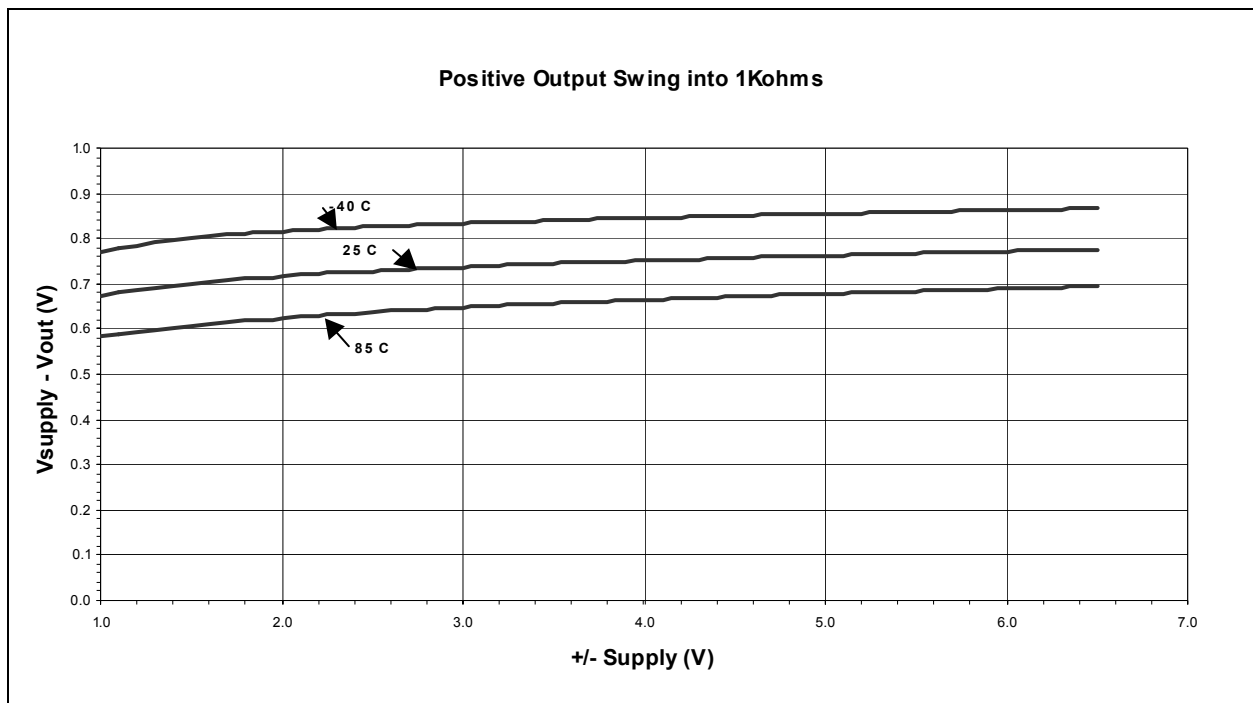


Figure 7 - Positive Output Swing into $1\text{K}\Omega$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

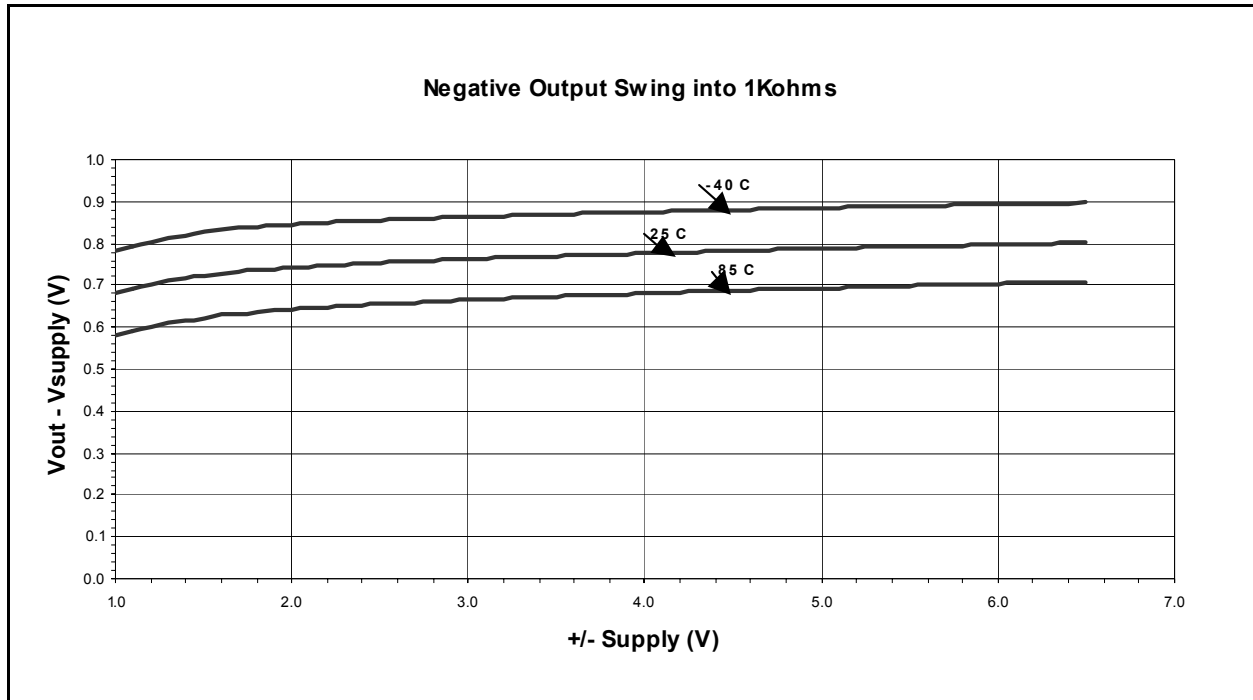


Figure 8 - Negative Output Swing into 1K Ω

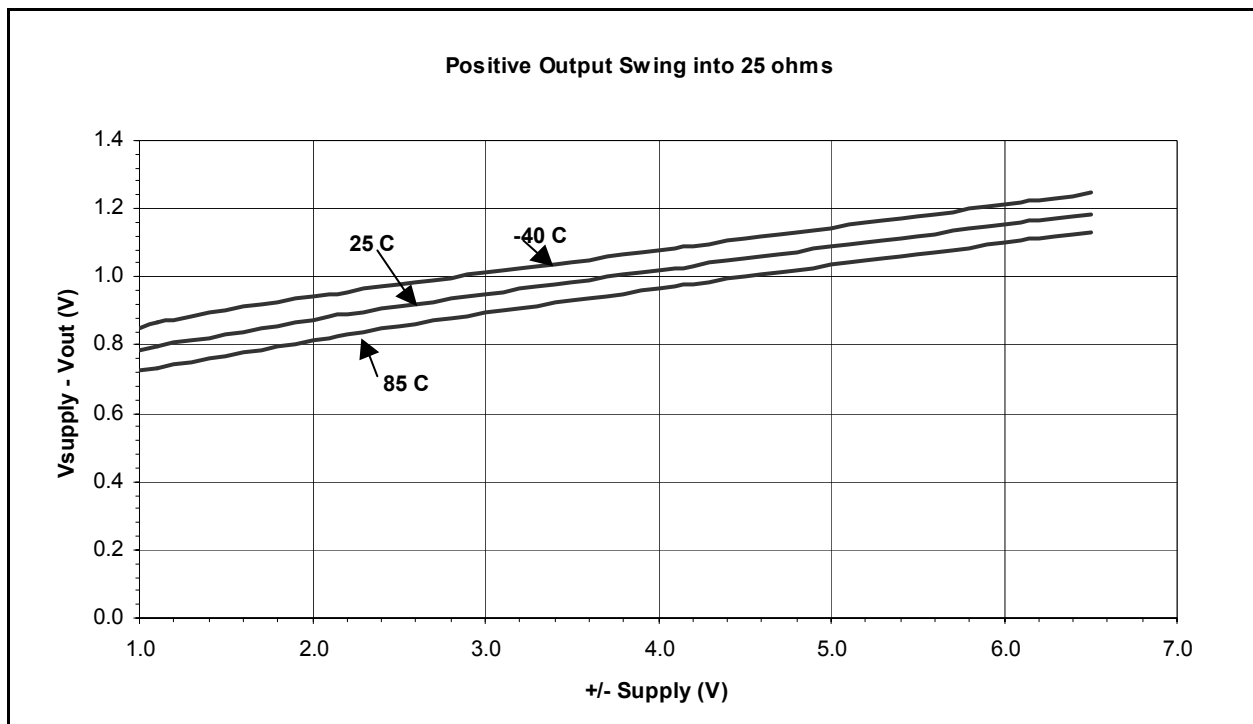


Figure 9 - Positive Output Swing into 25 Ω

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

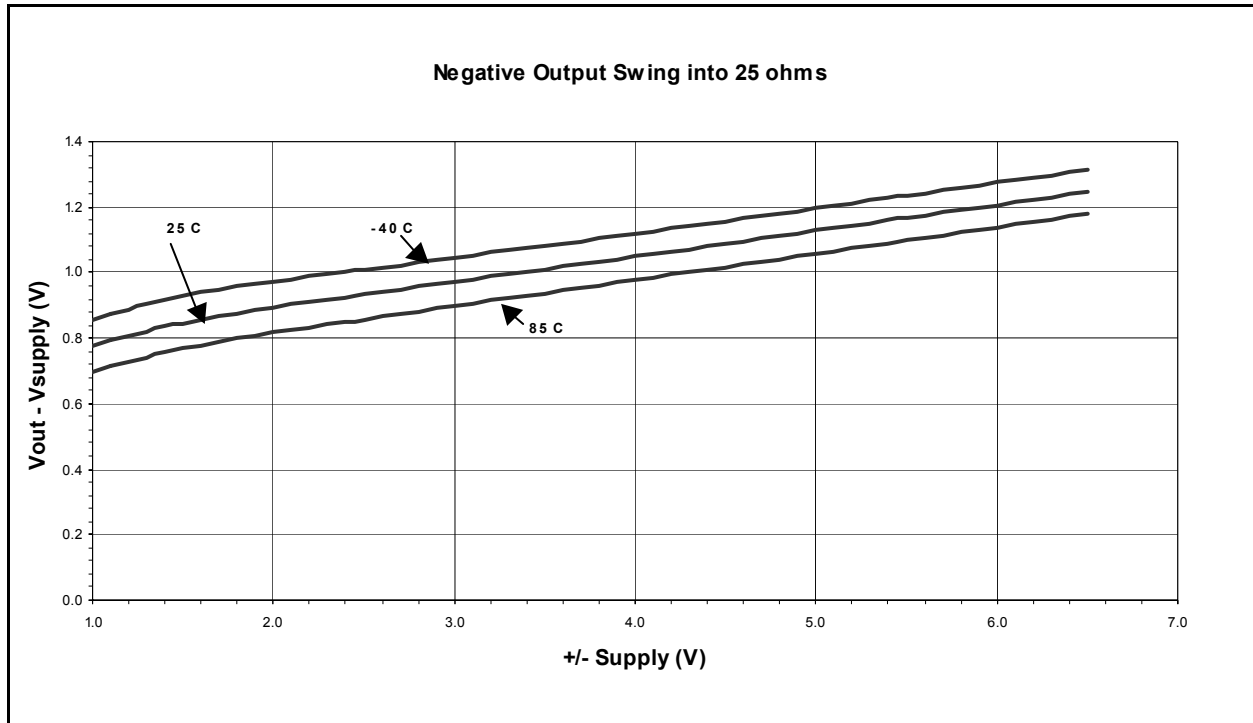


Figure 10 - Negative Output Swing into 25 Ω

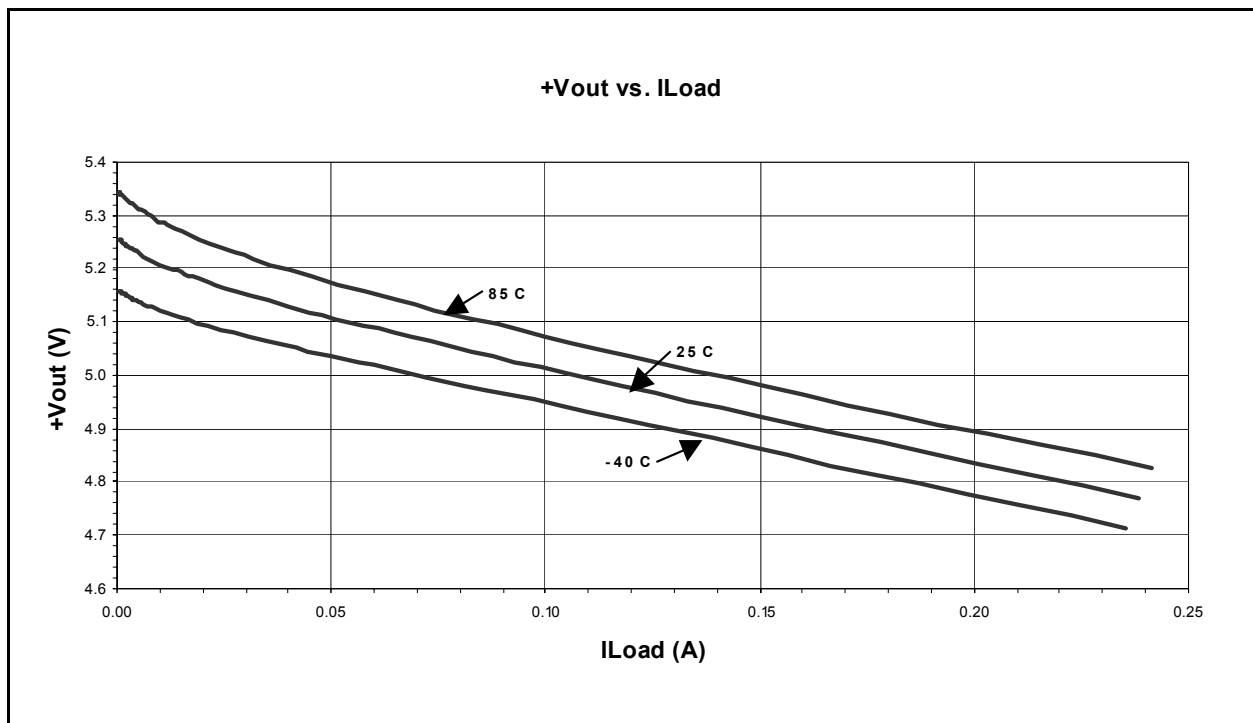


Figure 11 - +Vout vs. ILoad

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

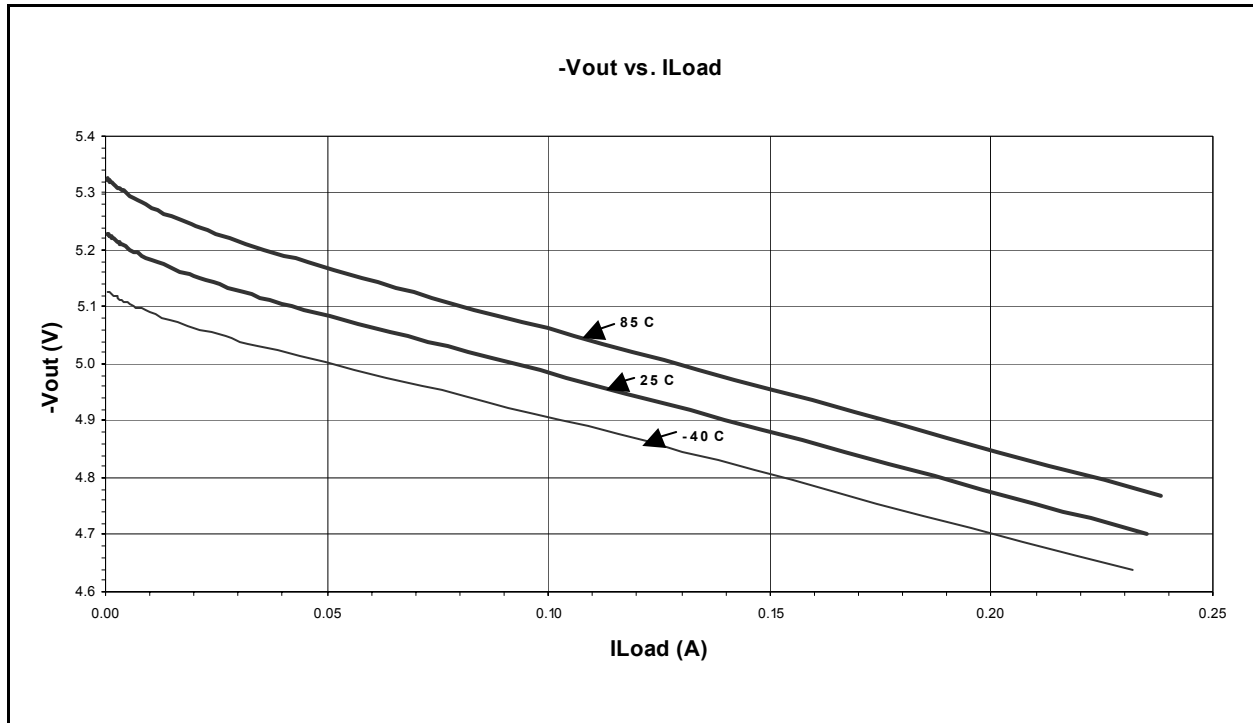


Figure 12 - -Vout vs. ILoad

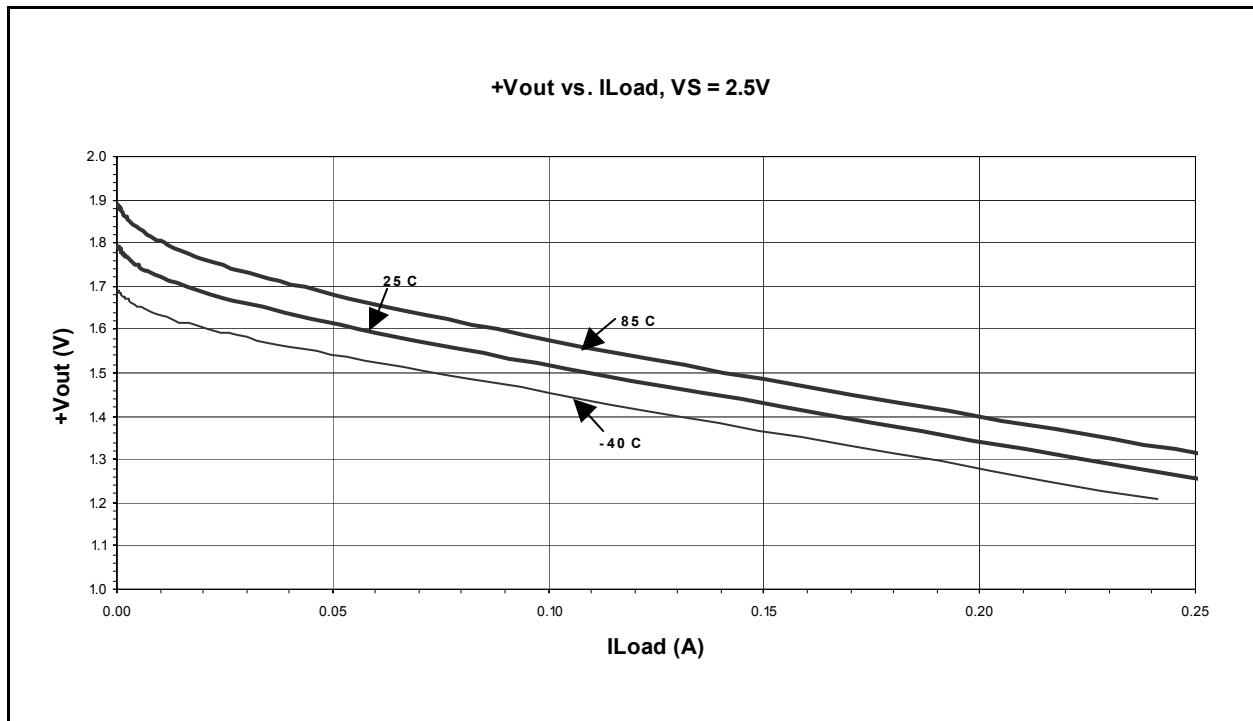


Figure 13 - +Vout vs. ILoad, Vs = $\pm 2.5\text{V}$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, gain = +2, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

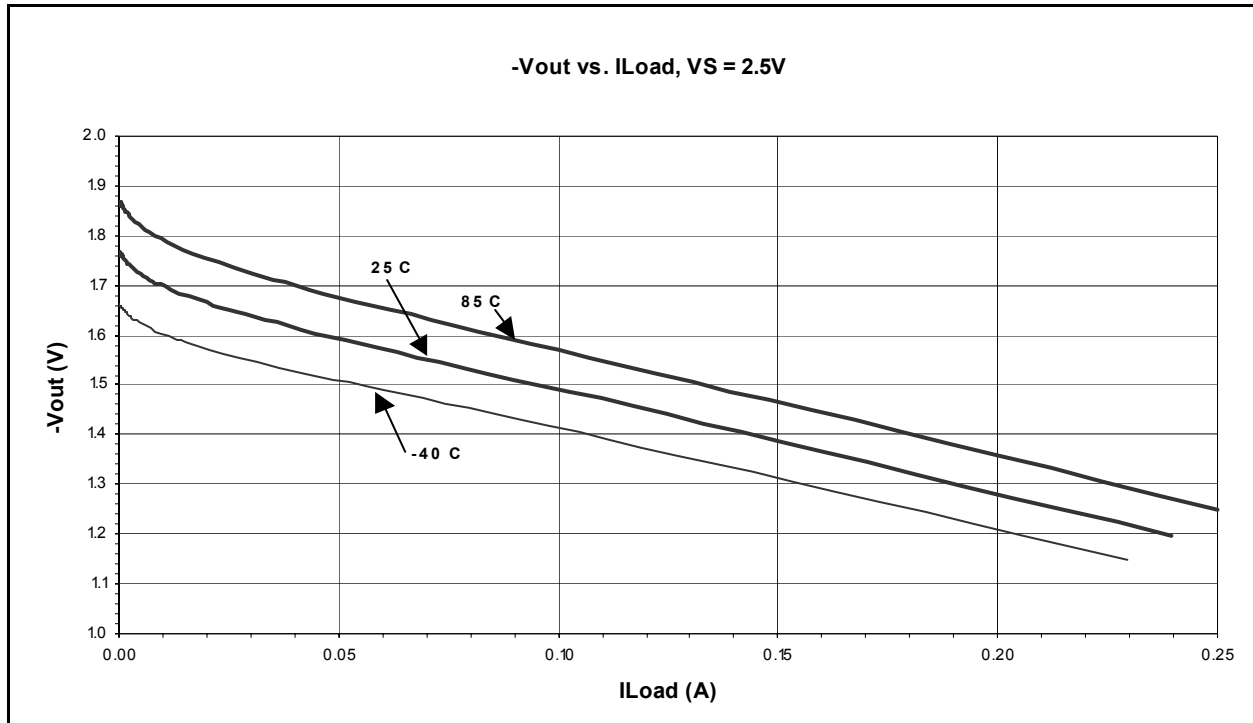


Figure 14 - -Vout vs. ILoad, Vs $\pm 2.5\text{V}$

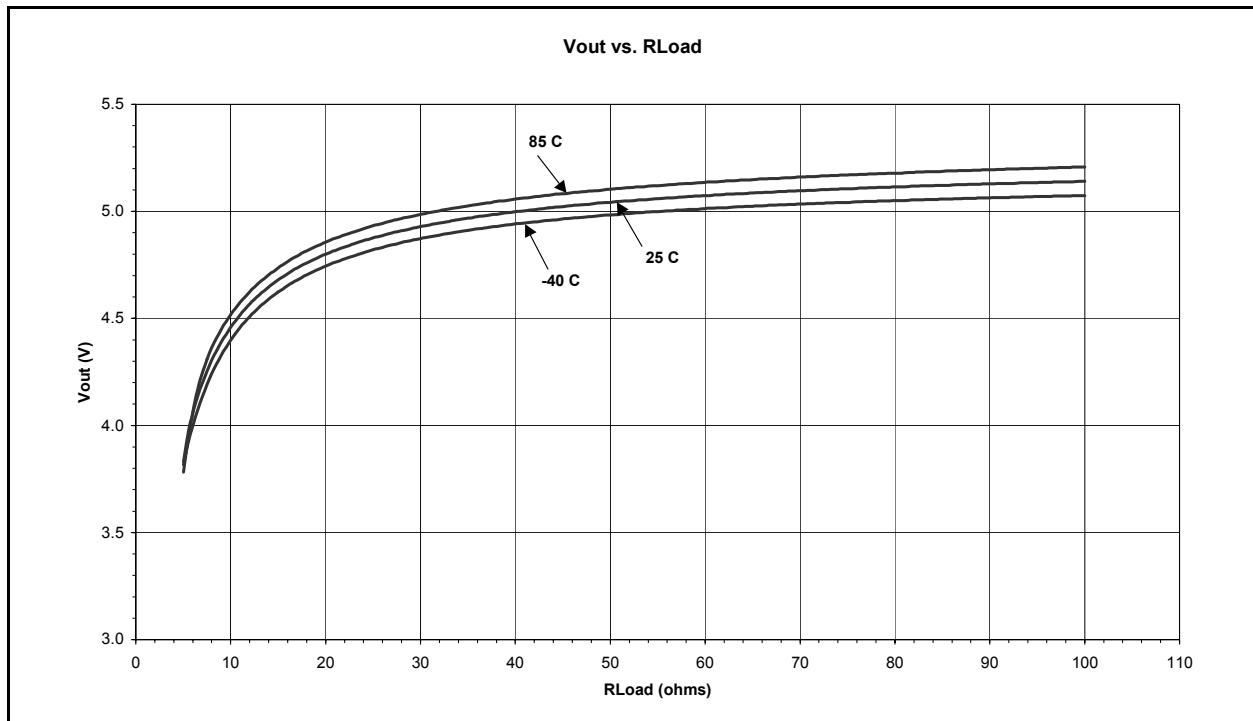


Figure 15 - Vout vs Rload

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

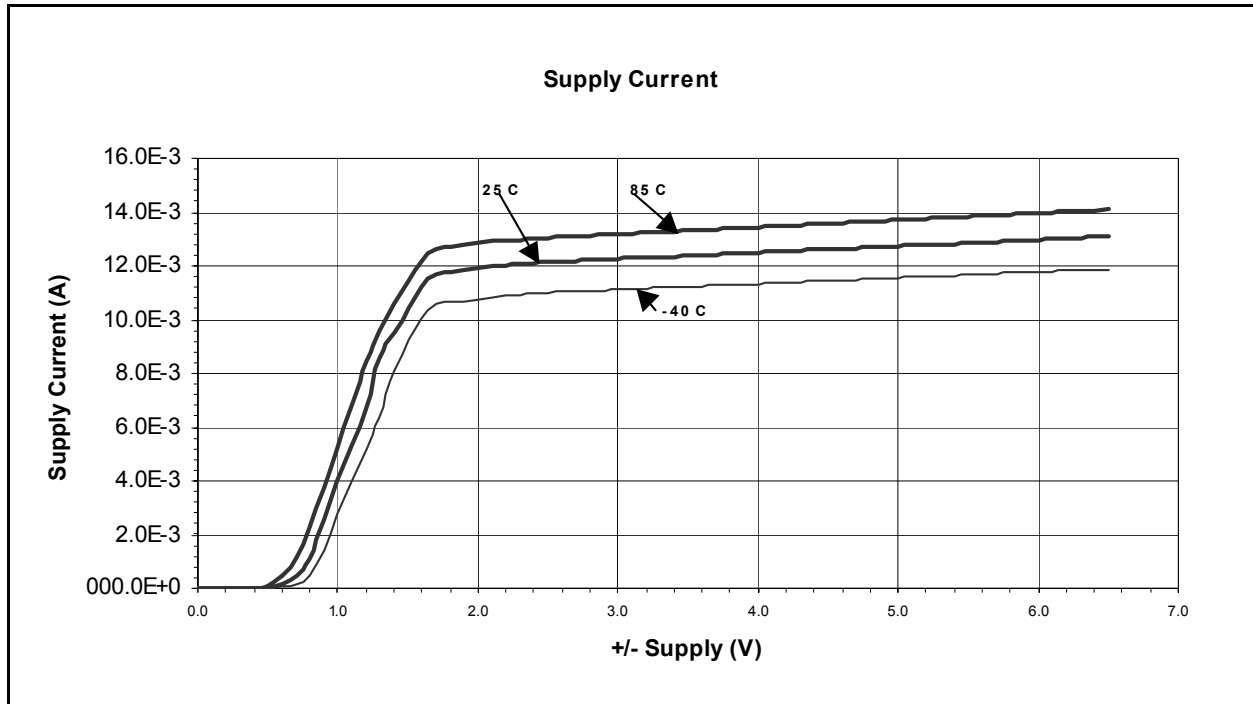


Figure 16 - Supply Current vs. Supply Voltage

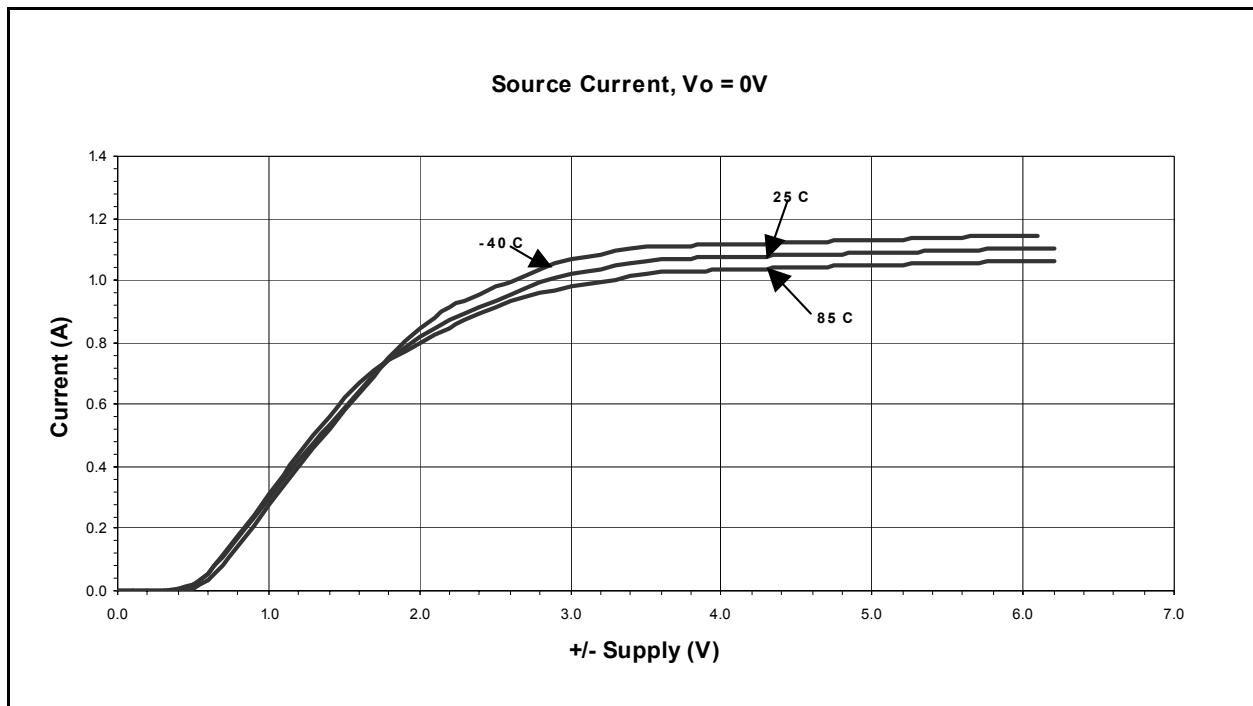


Figure 17 - Sourcing Current vs. Supply Voltage

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

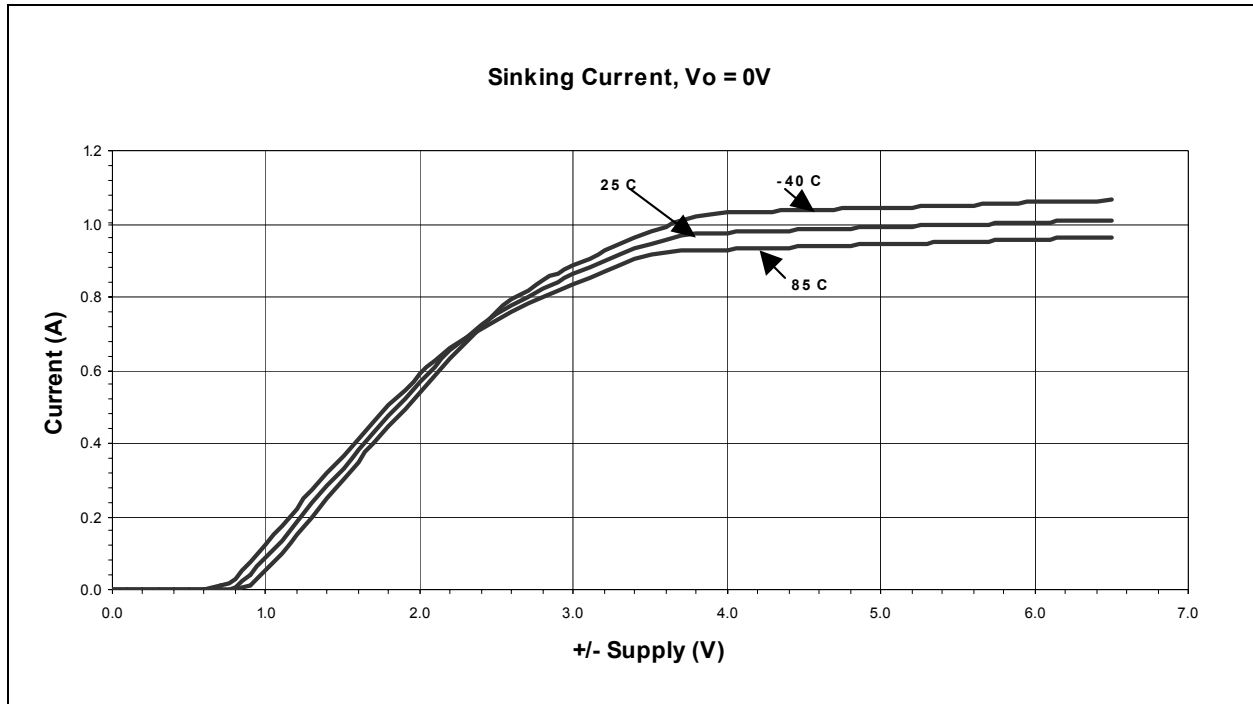


Figure 18 - Sinking Current vs. Supply Voltage

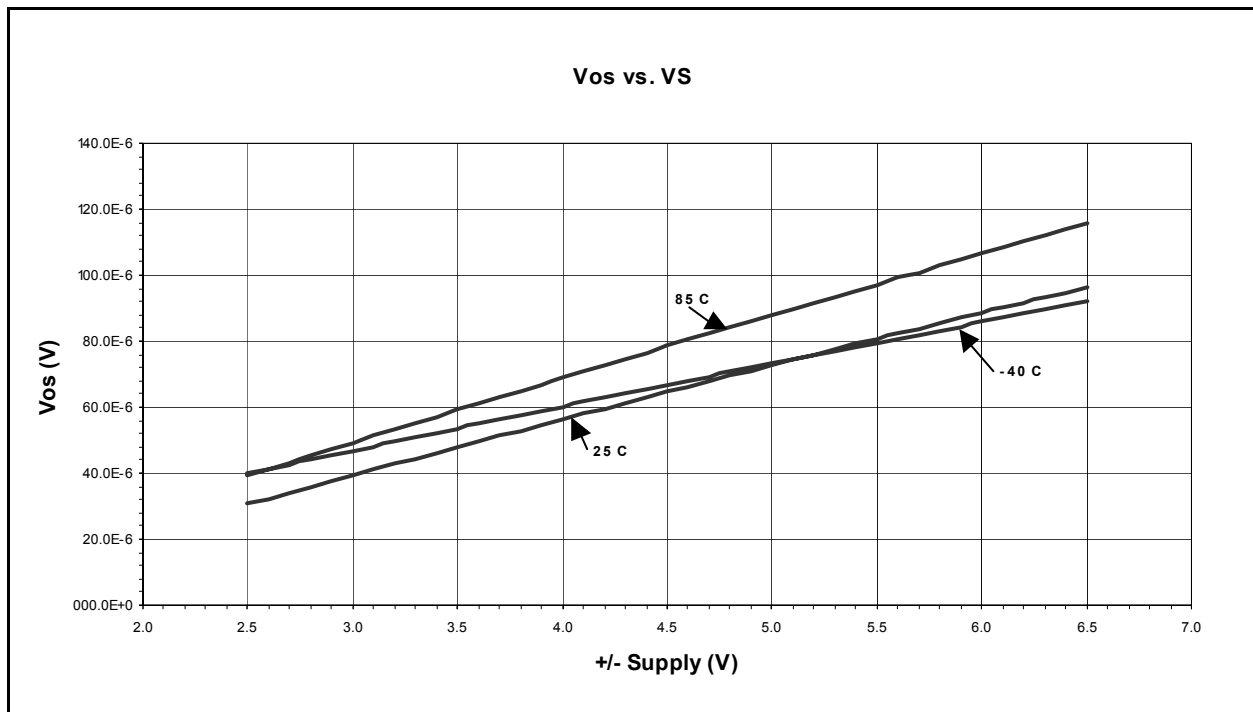


Figure 19 - V_{os} vs. V_S

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

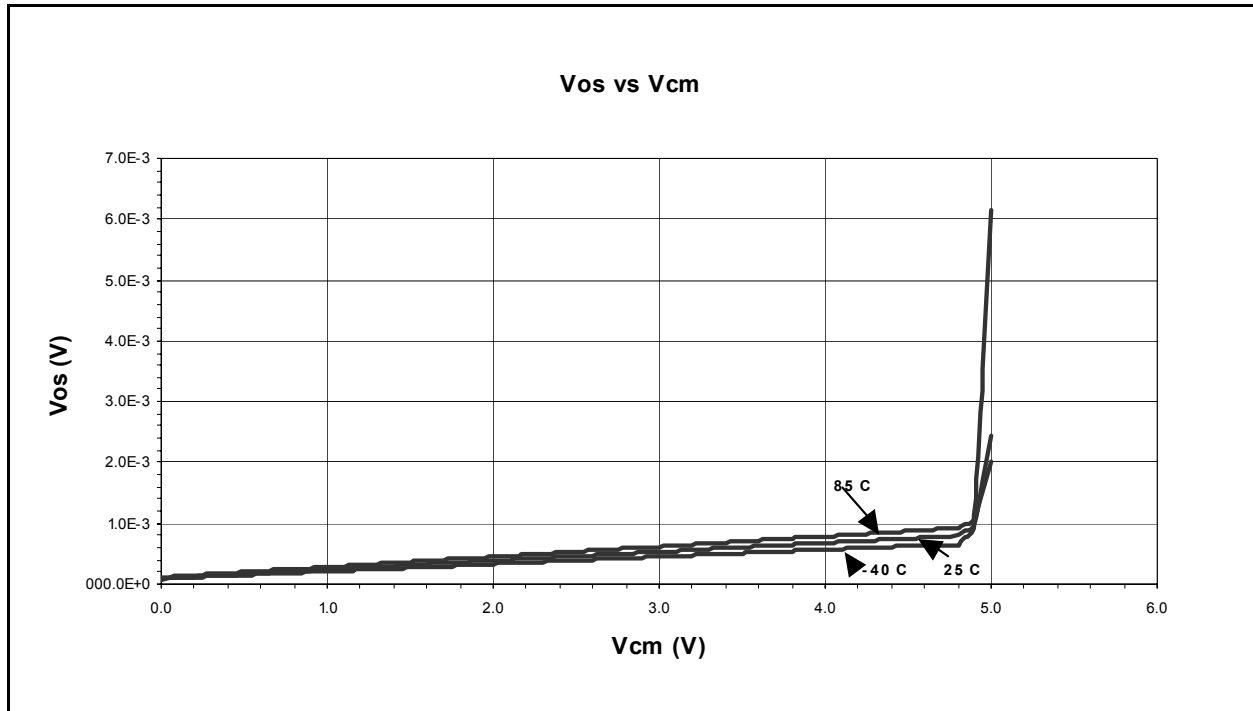


Figure 20 - V_{os} vs. V_{cm}

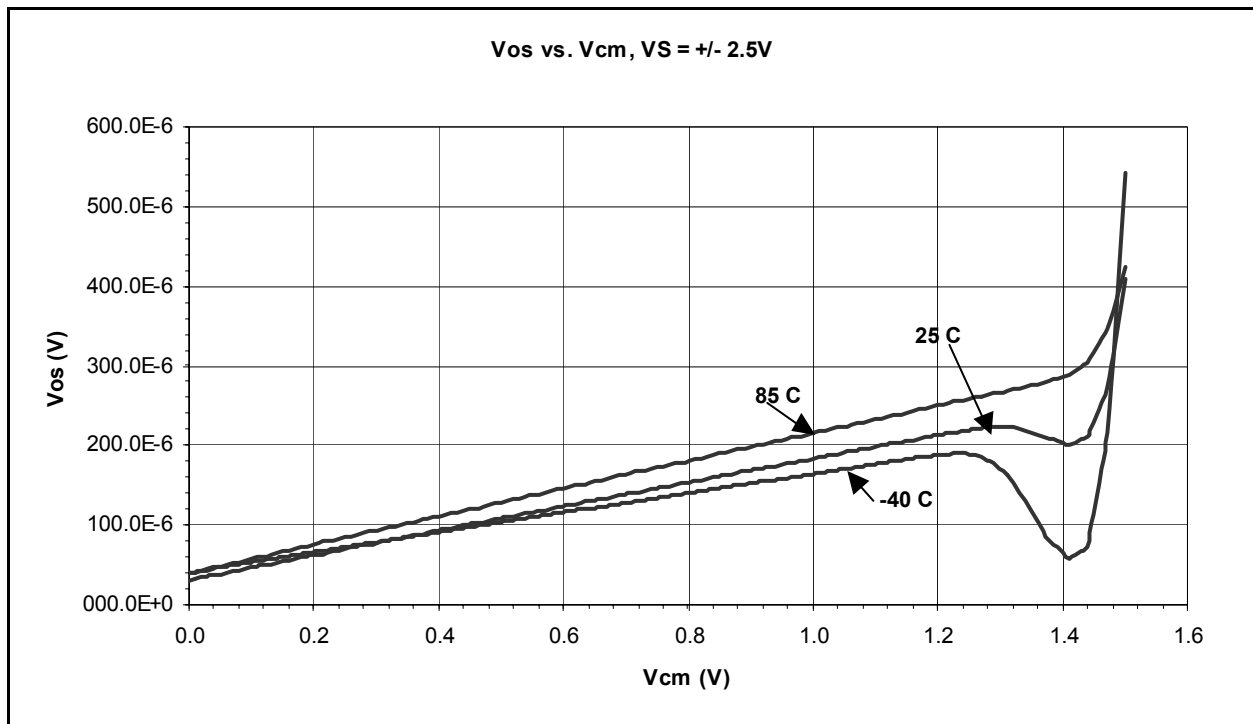


Figure 21 - V_{os} vs. V_{cm} , $V_S = \pm 2.5\text{V}$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

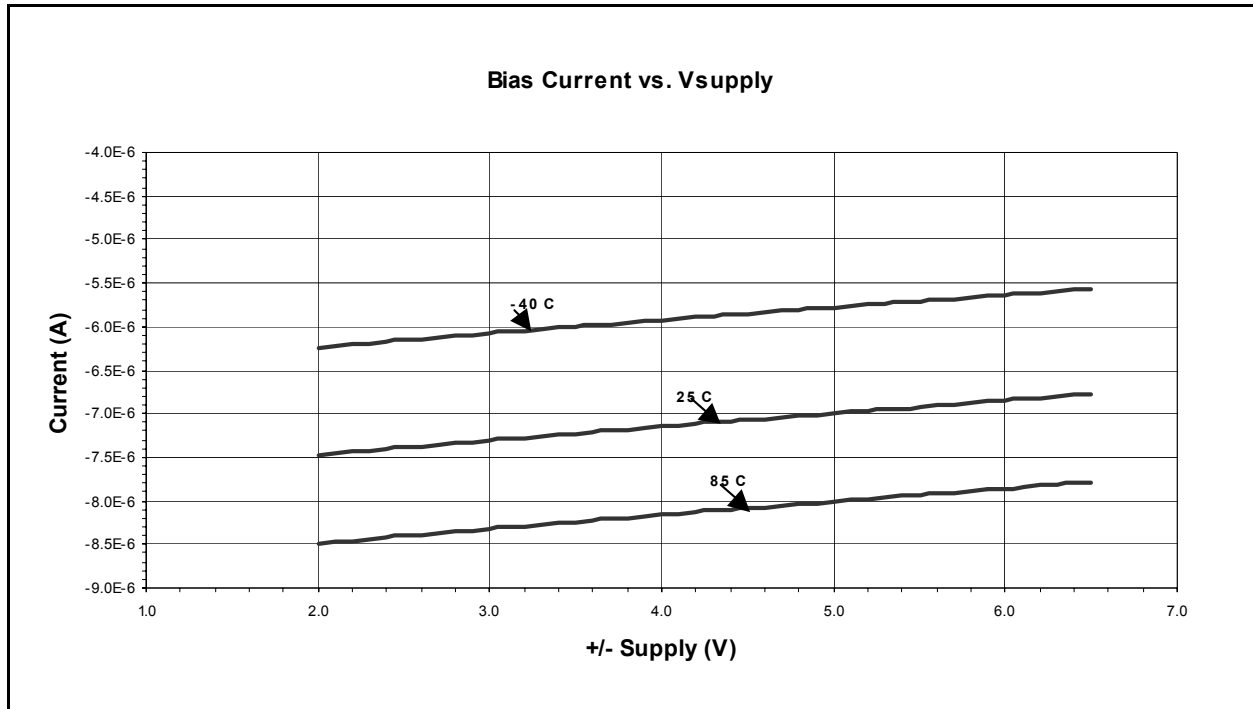


Figure 22 - Bias Current vs. Vsupply

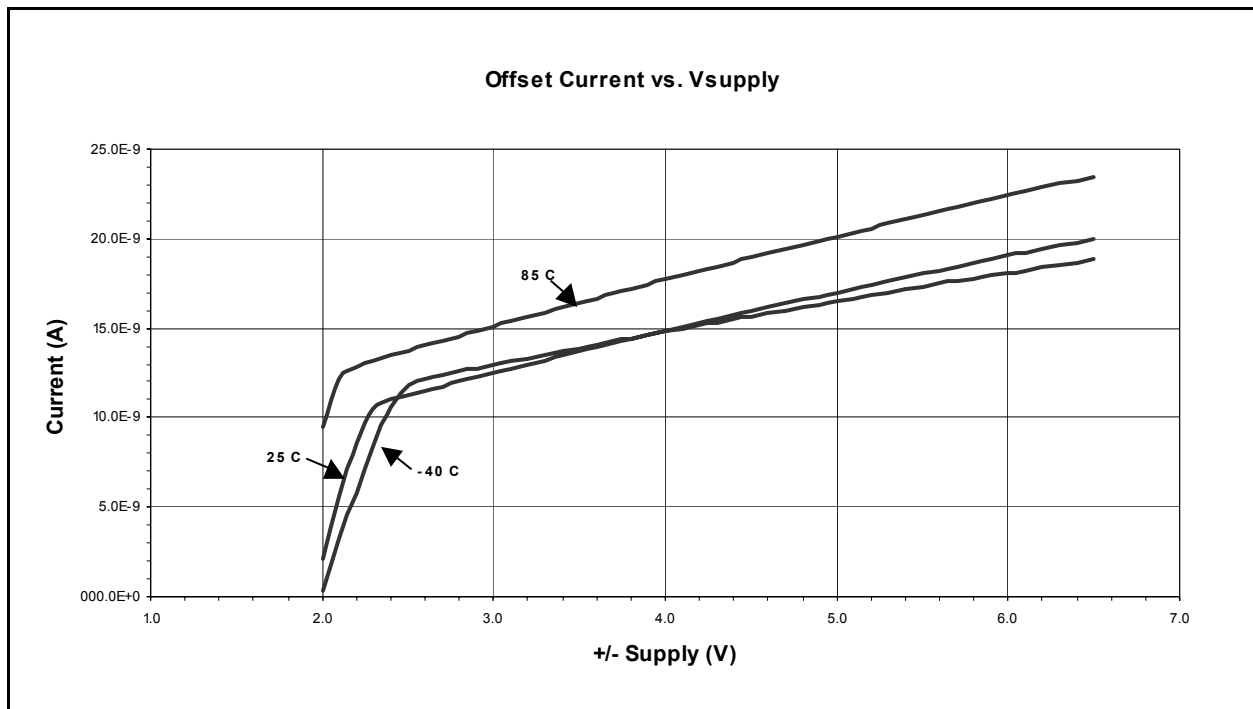


Figure 23 - Offset Current vs. Vsupply

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

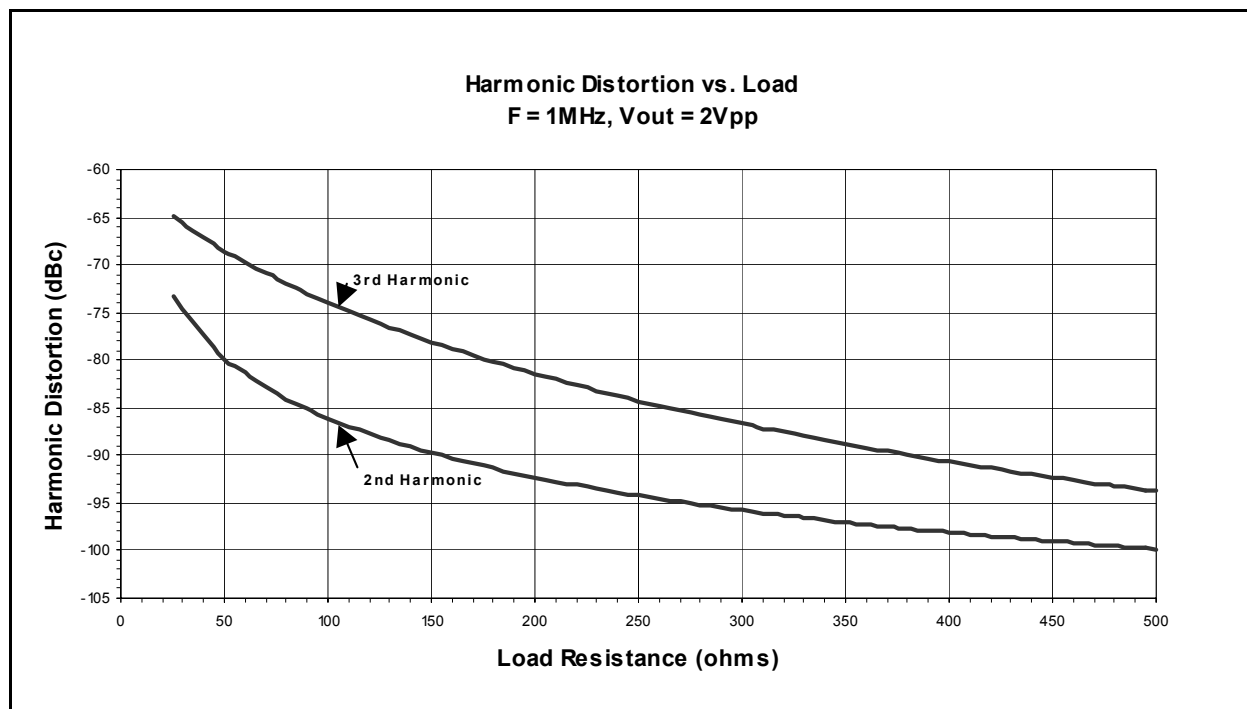


Figure 24 - Harmonic Distortion vs. Load F = 1MHZ, Vout = 2Vpp

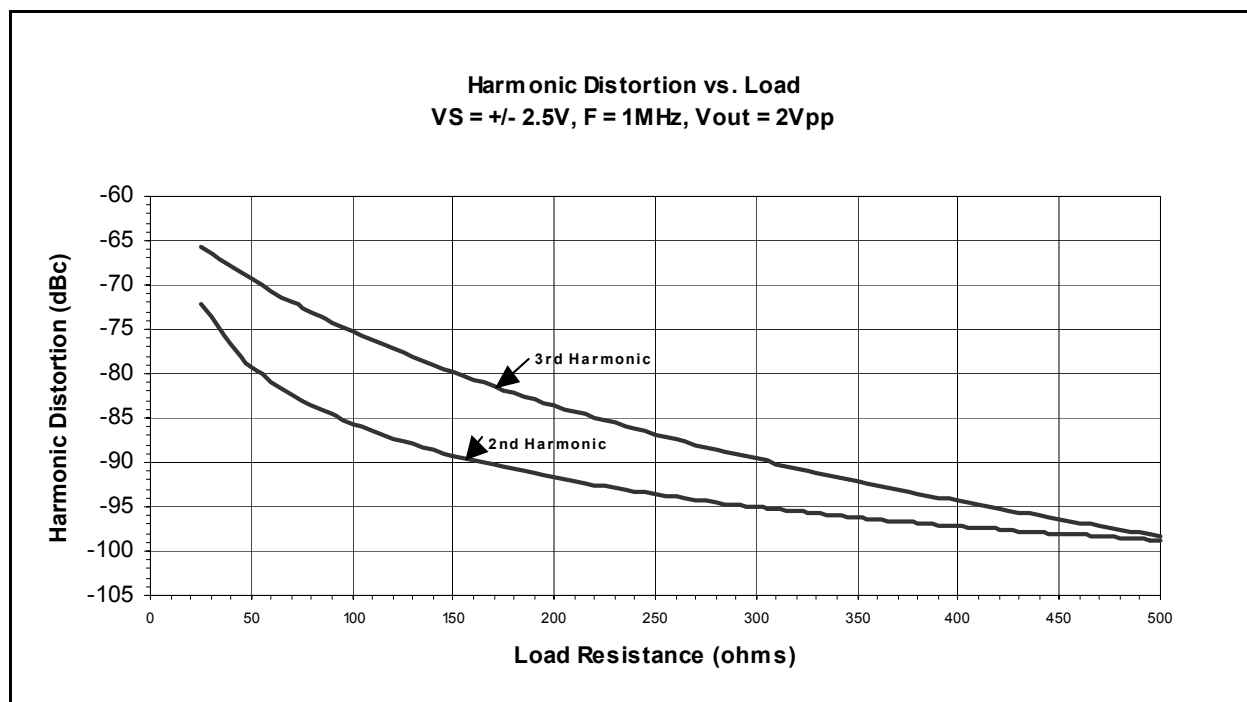


Figure 25 - Harmonic Distortion vs. Load $V_S = \pm 2.5\text{V}$, F = 1MHZ, Vout = 2Vpp

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

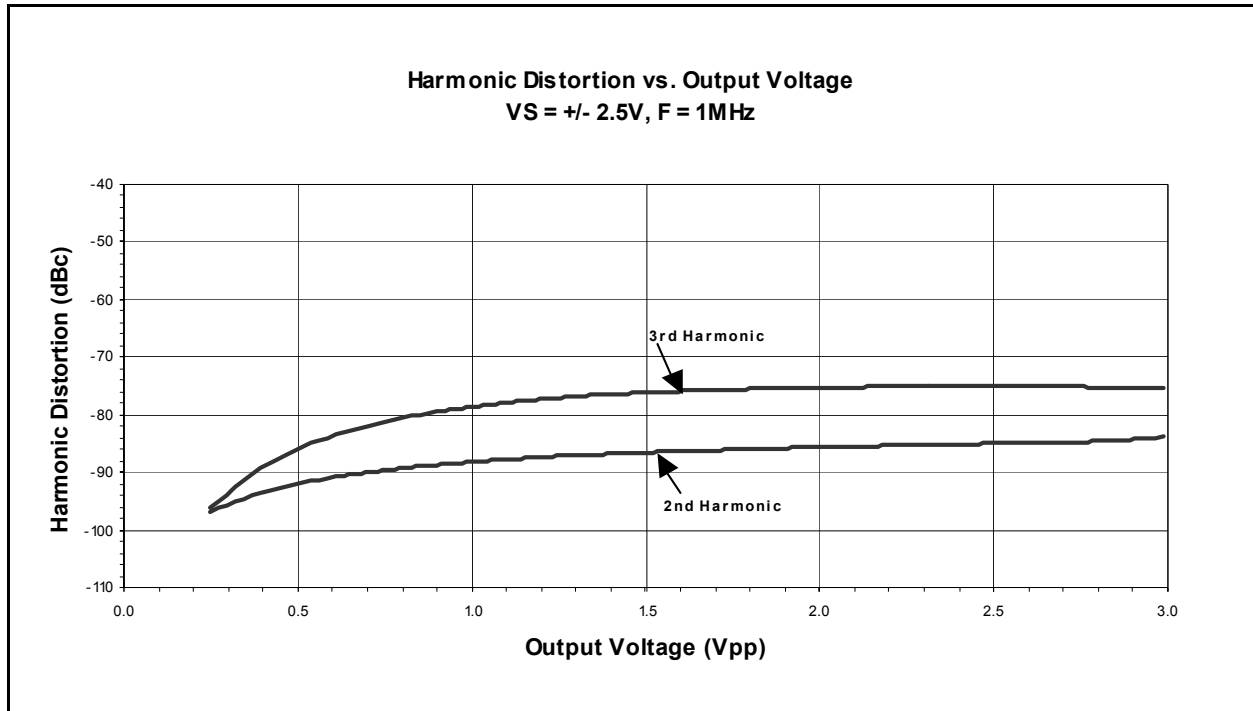


Figure 26 - Harmonic Distortion vs. Output Voltage $V_S = \pm 2.5\text{V}$, $F = 1\text{MHz}$

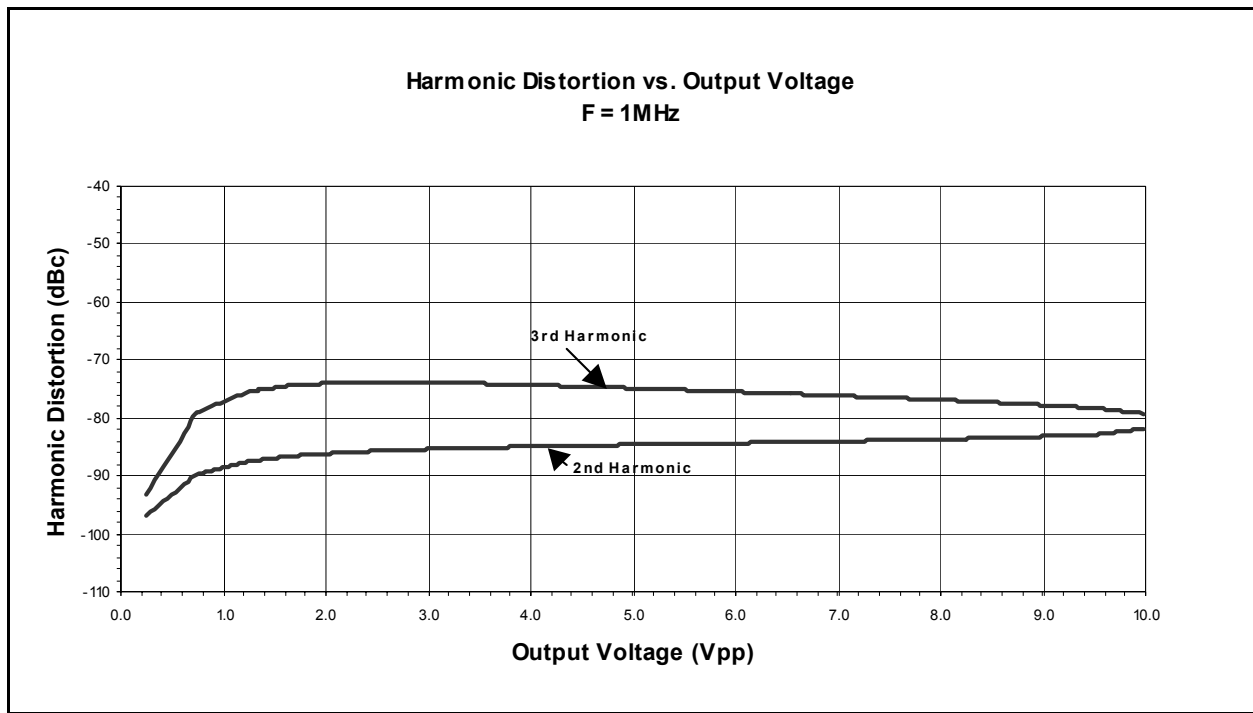


Figure 27 - Harmonic Distortion vs. Output Voltage $F = 1\text{MHz}$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

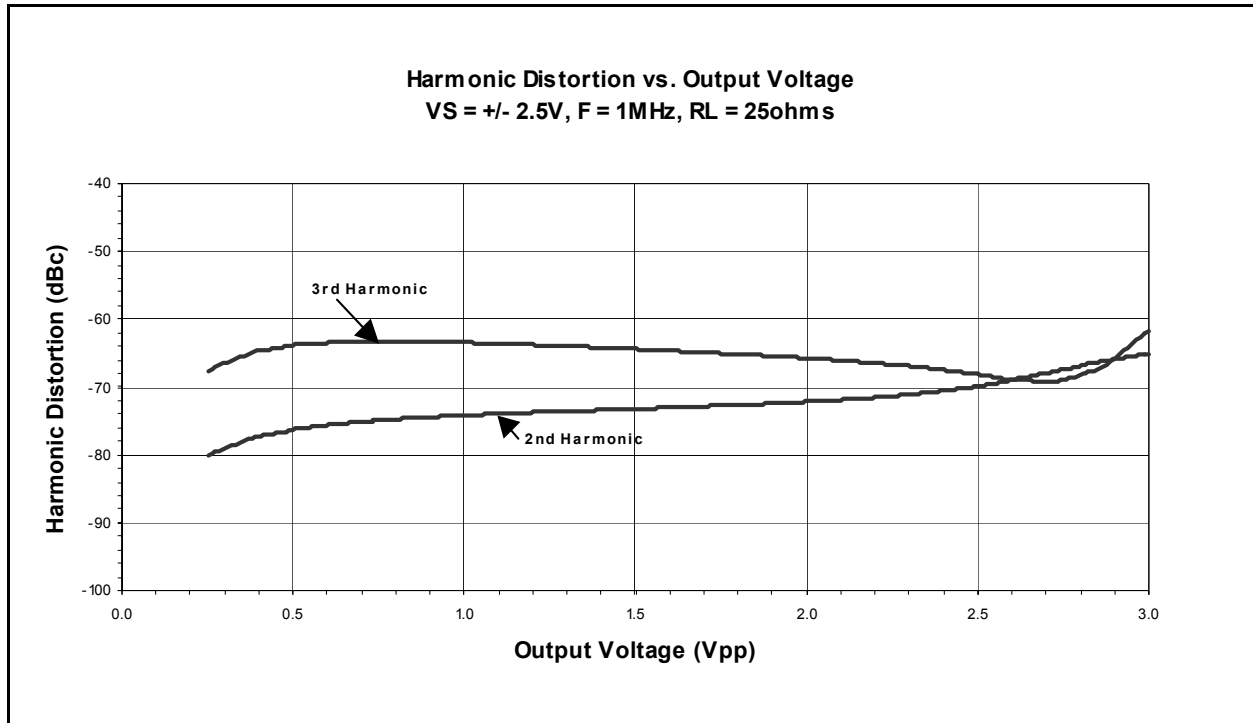


Figure 28 - Harmonic Distortion vs. Output Voltage $V_S = \pm 2.5\text{V}$, $F = 1\text{MHz}$, $R_L = 25\Omega$

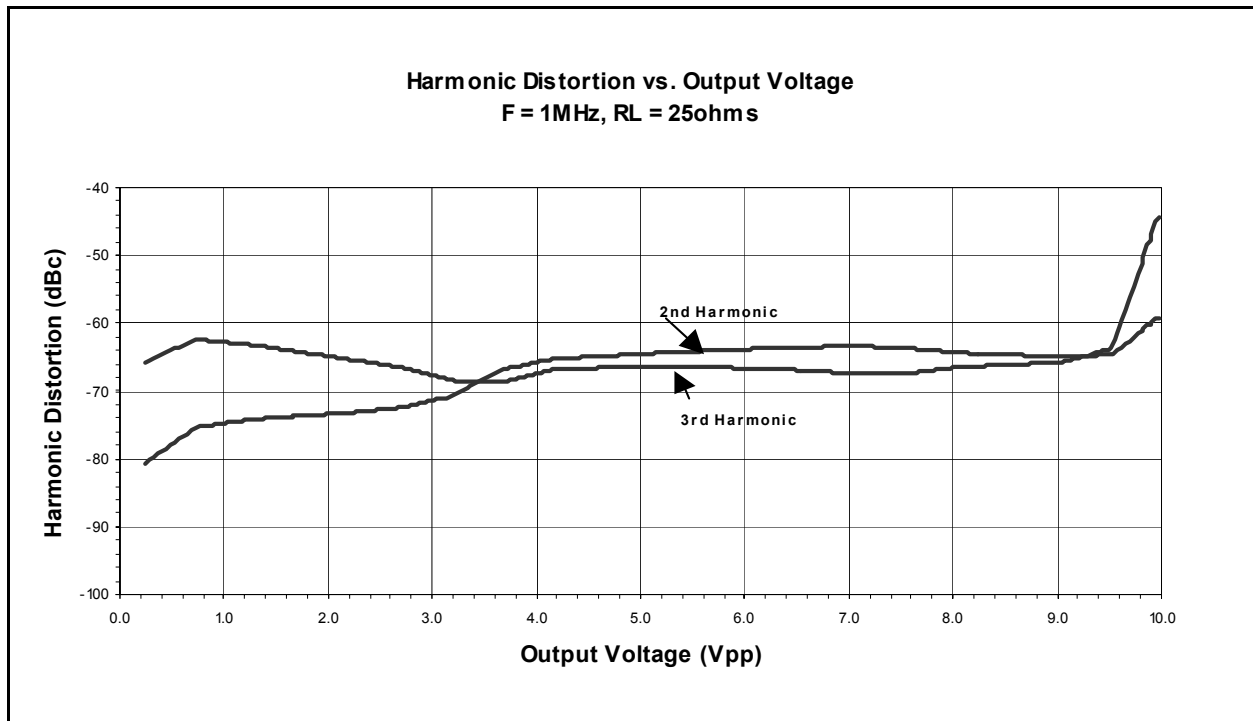


Figure 29 - Harmonic Distortion vs. Output Voltage $F = 1\text{MHz}$, $R_L = 25\Omega$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

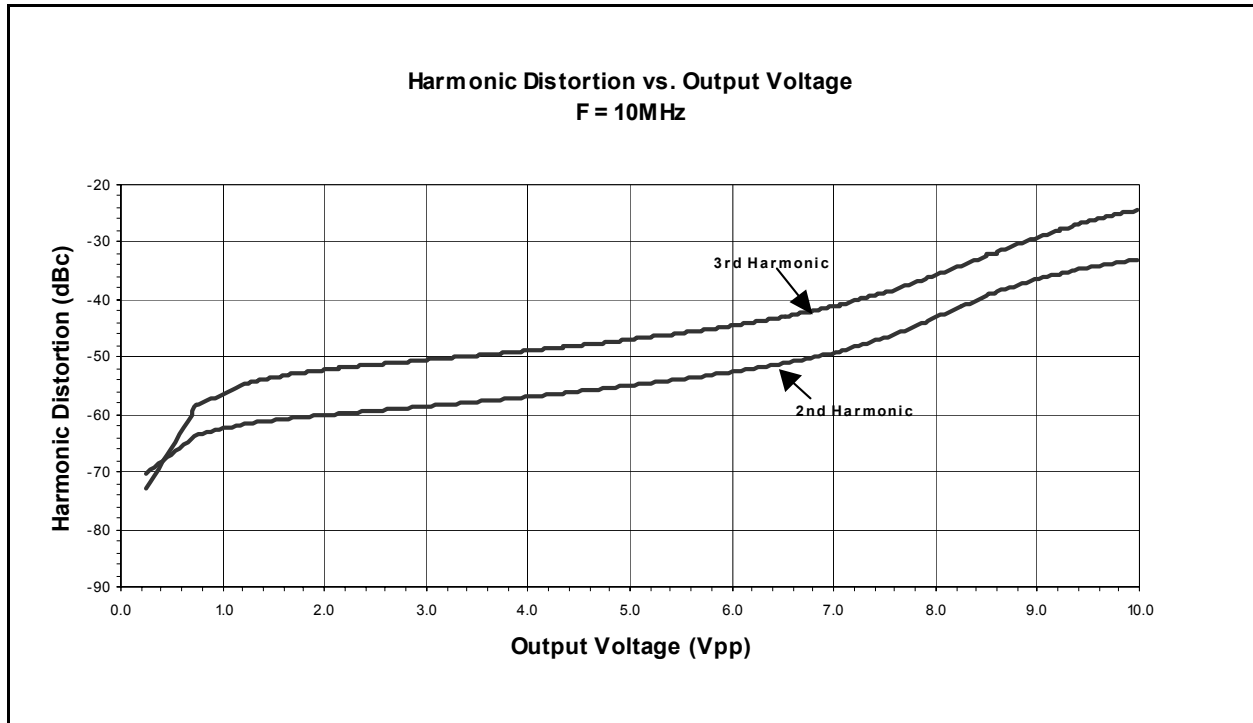


Figure 30 - Harmonic Distortion vs. Output Voltage F = 10MHz

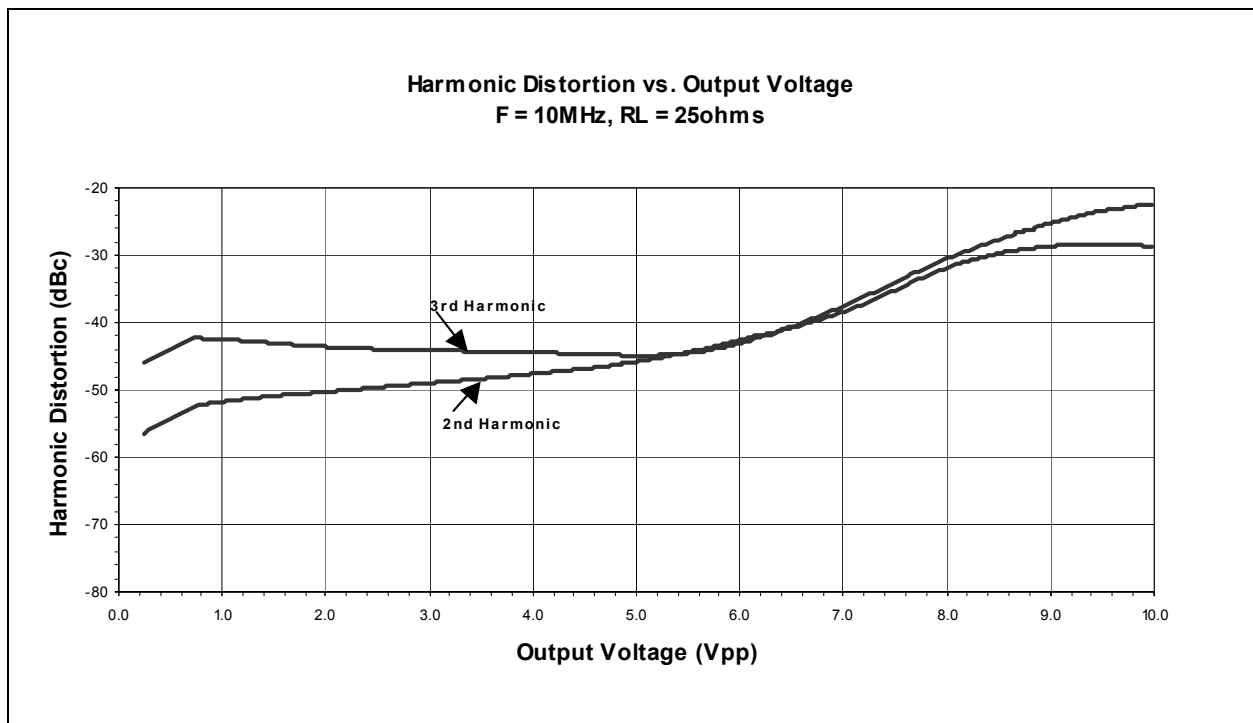


Figure 31 - Harmonic Distortion vs. Output Voltage F = 10MHz, $R_L = 25\Omega$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

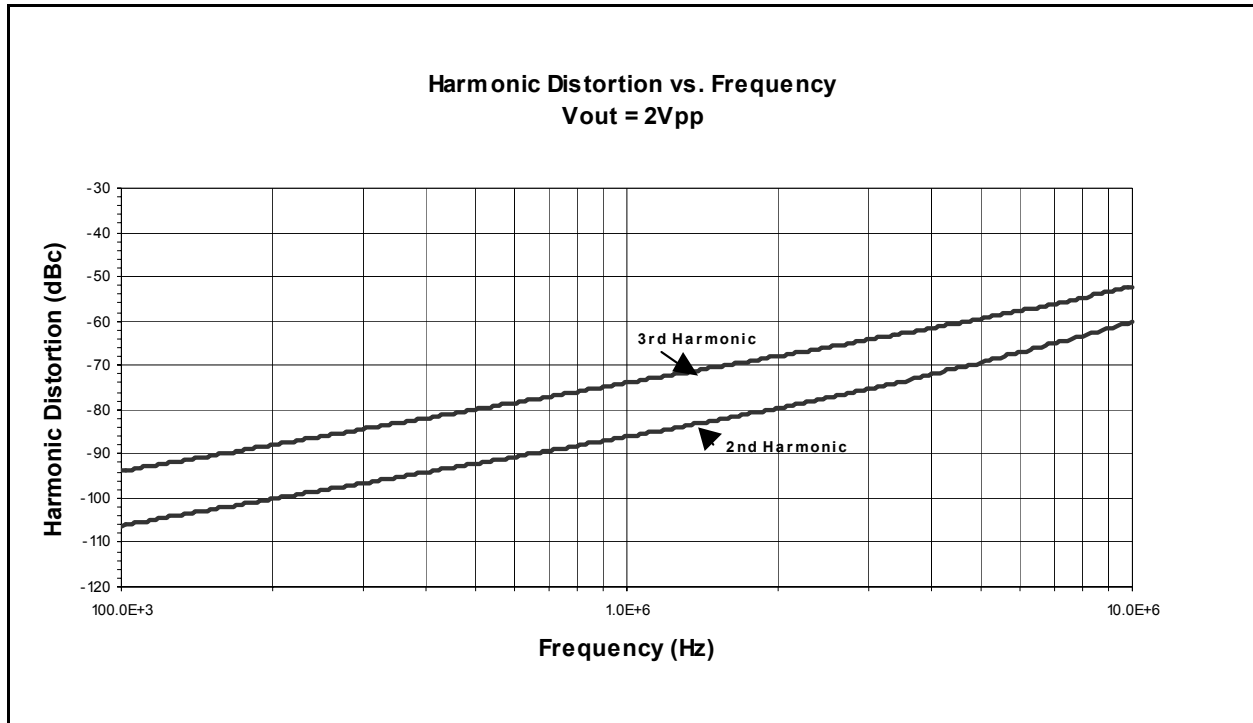


Figure 32 - Harmonic Distortion vs. Frequency $V_{out} = 2\text{Vpp}$

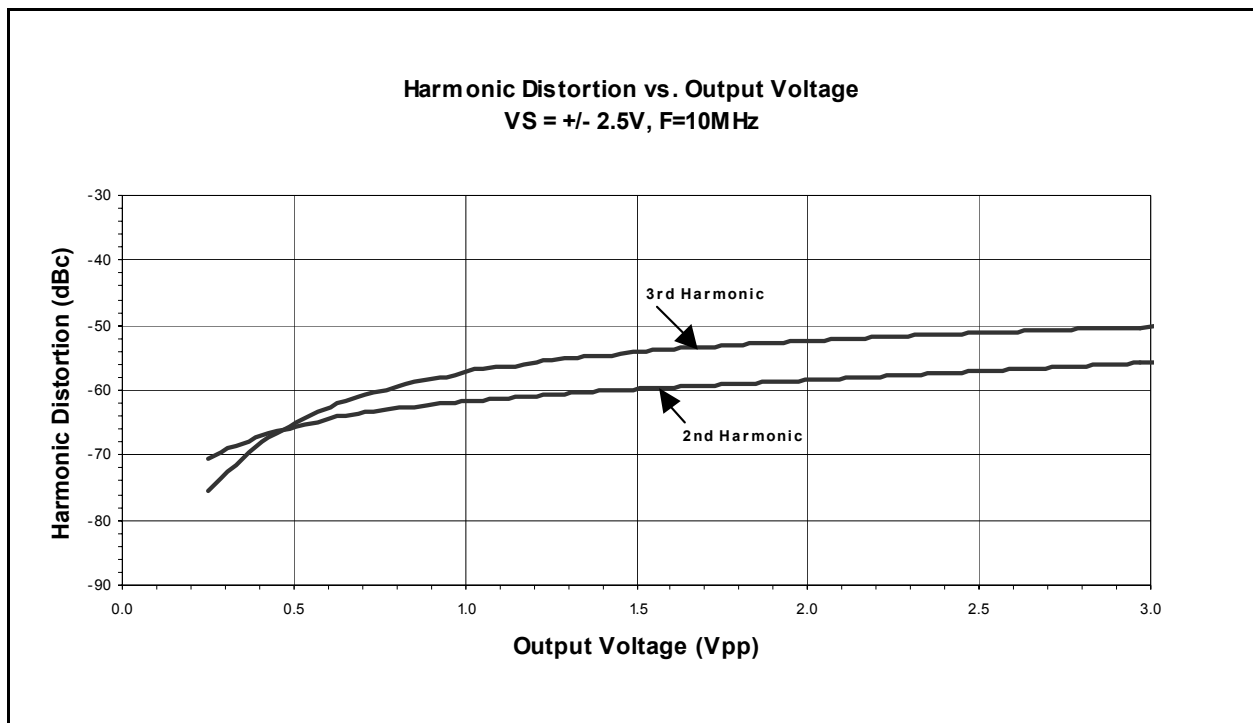


Figure 33 - Harmonic Distortion vs. Output Voltage $V_S = \pm 2.5\text{V}$, $F = 10\text{MHz}$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_s = 6\text{V}$. Unless otherwise specified.

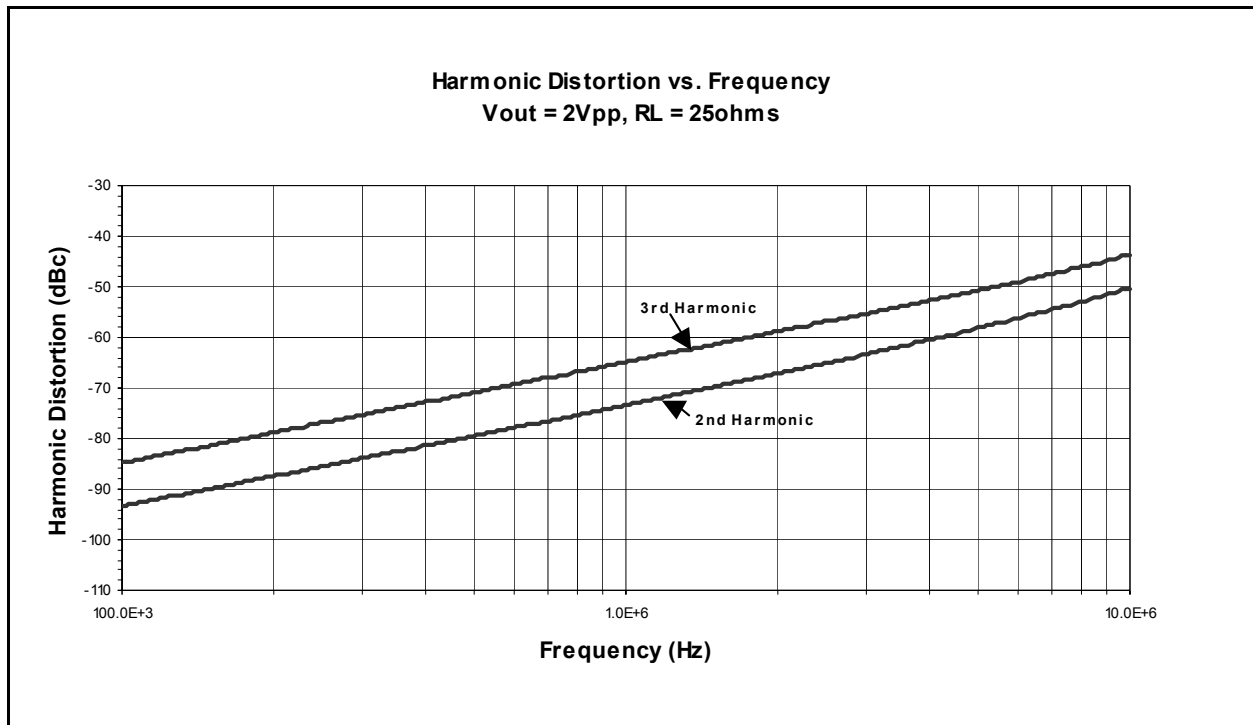


Figure 34 - Harmonic Distortion vs. Frequency $V_{out} = 2\text{V}_{pp}$, $R_L = 25\Omega$

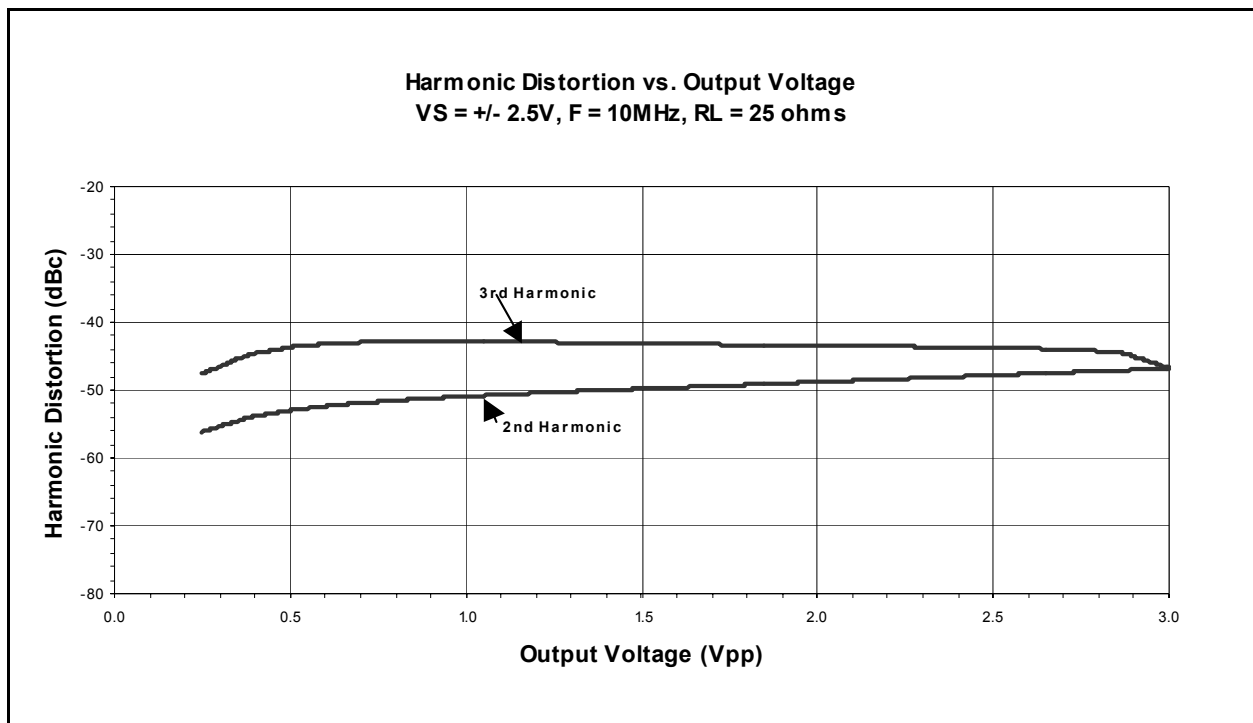


Figure 35 - Harmonic Distortion vs. Output Voltage $V_S = \pm 2.5\text{V}$, $F = 10\text{MHz}$, $R_L = 25\Omega$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

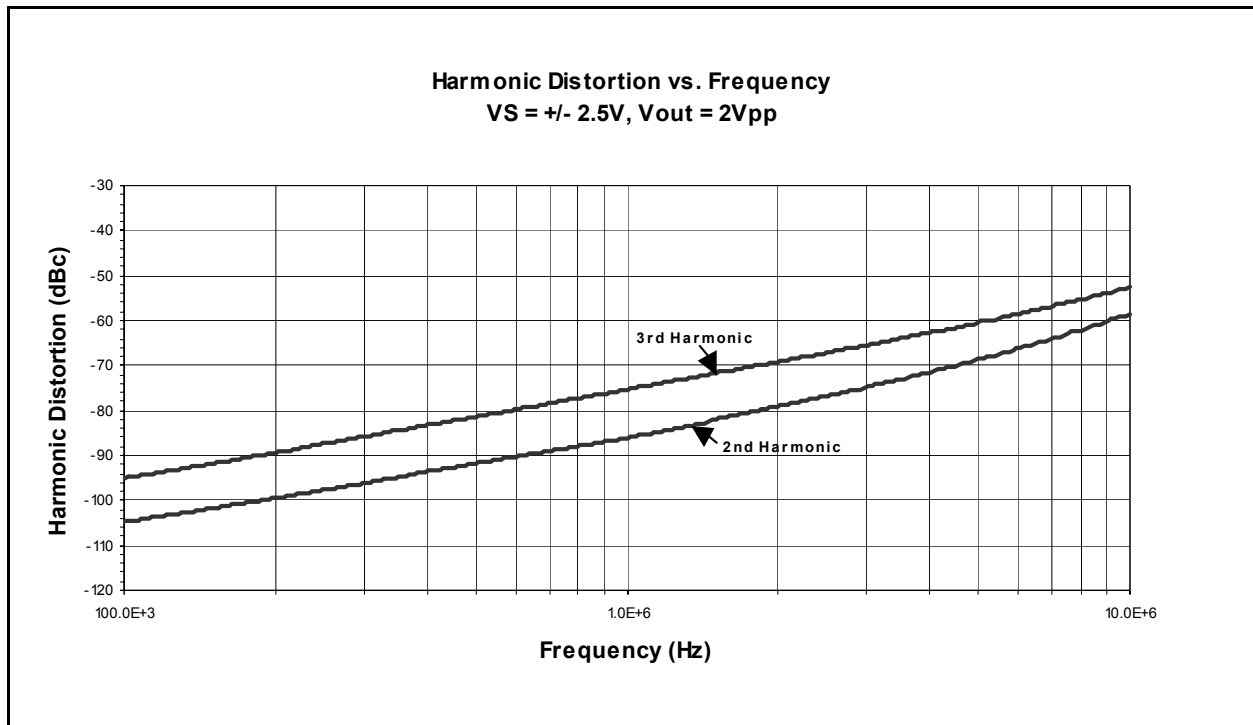


Figure 36 - Harmonic Distortion vs. Frequency $V_S = \pm 2.5\text{V}$, $V_{out} = 2\text{Vpp}$

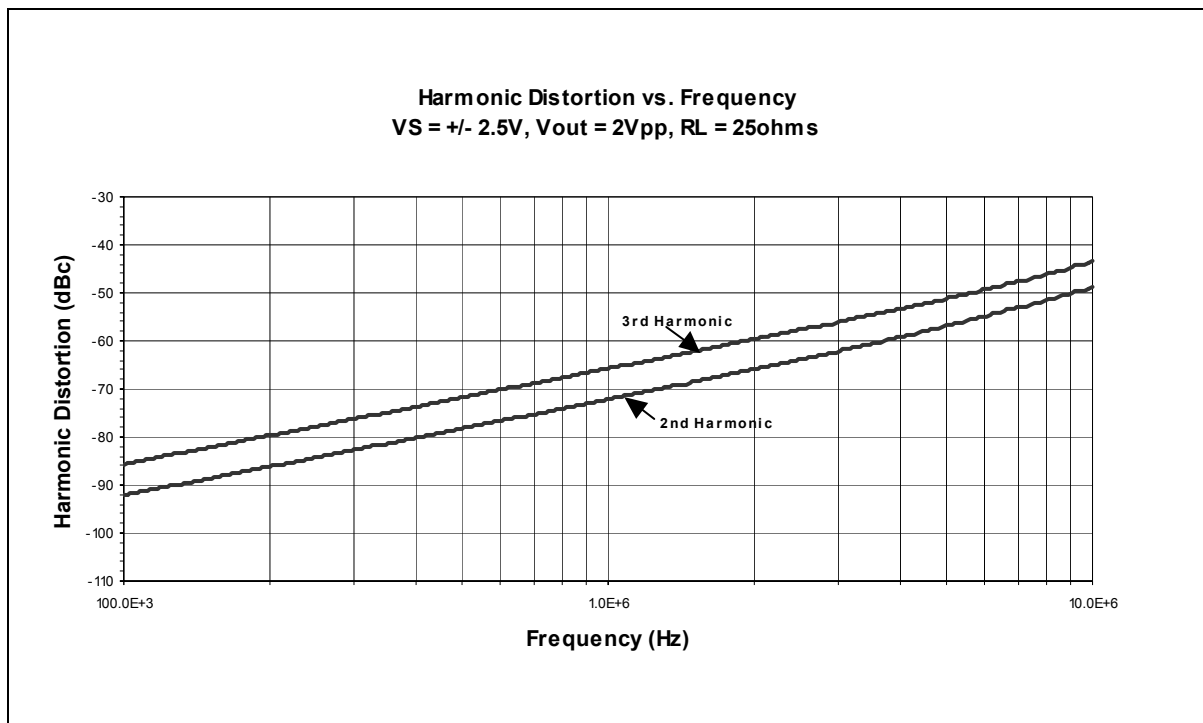


Figure 37 - Harmonic Distortion vs. Frequency $V_S = \pm 2.5\text{V}$, $V_{out} = 2\text{Vpp}$, $R_L = 25\Omega$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

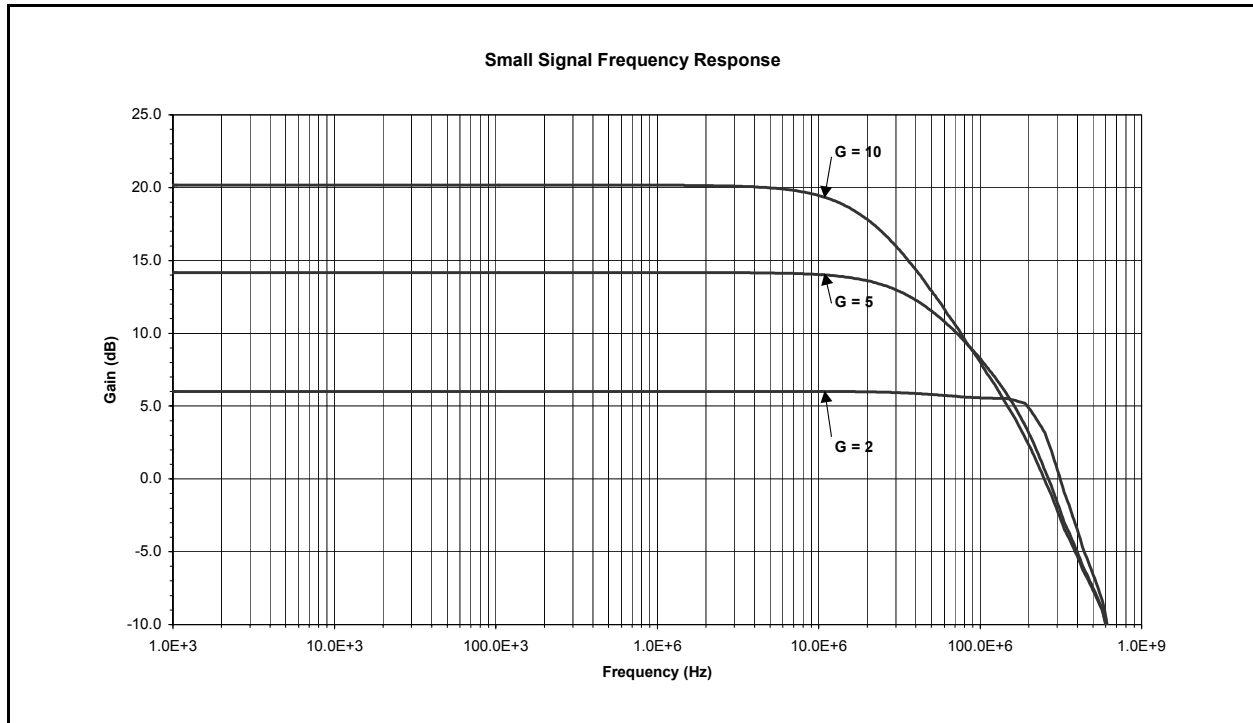


Figure 38 - Small Signal Frequency Response

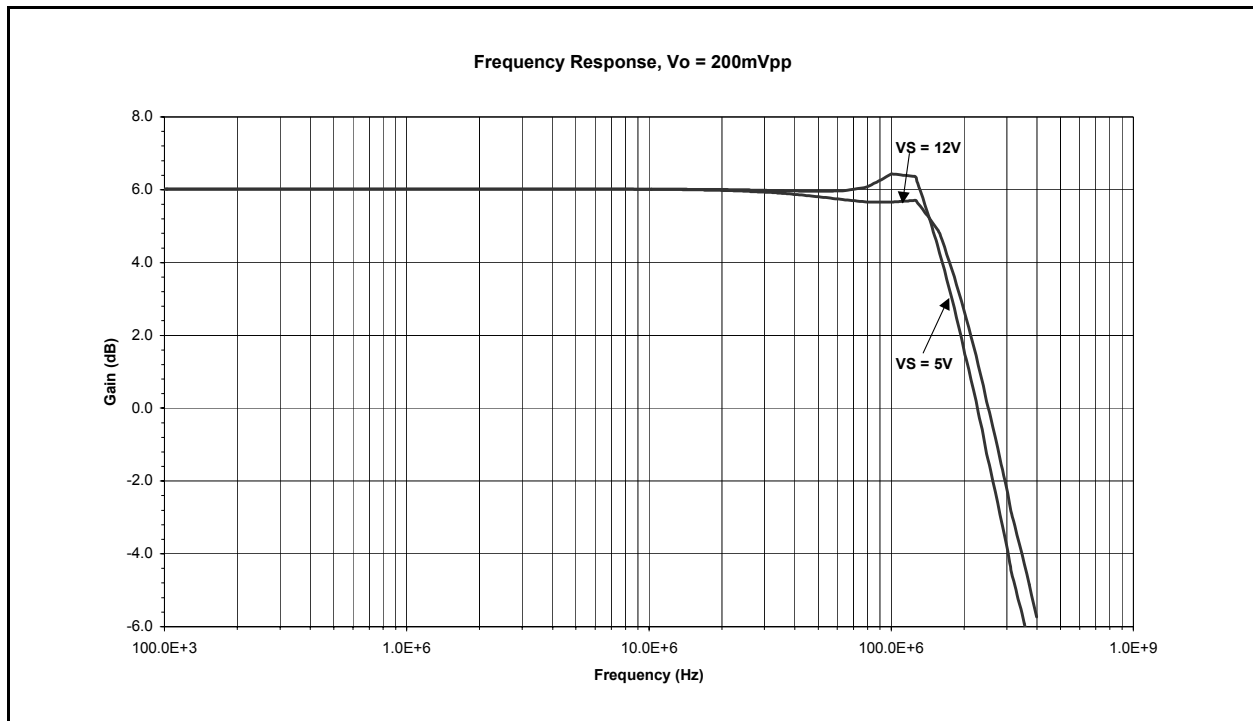


Figure 39 - Frequency Response, $V_O = 200\text{mVpp}$

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

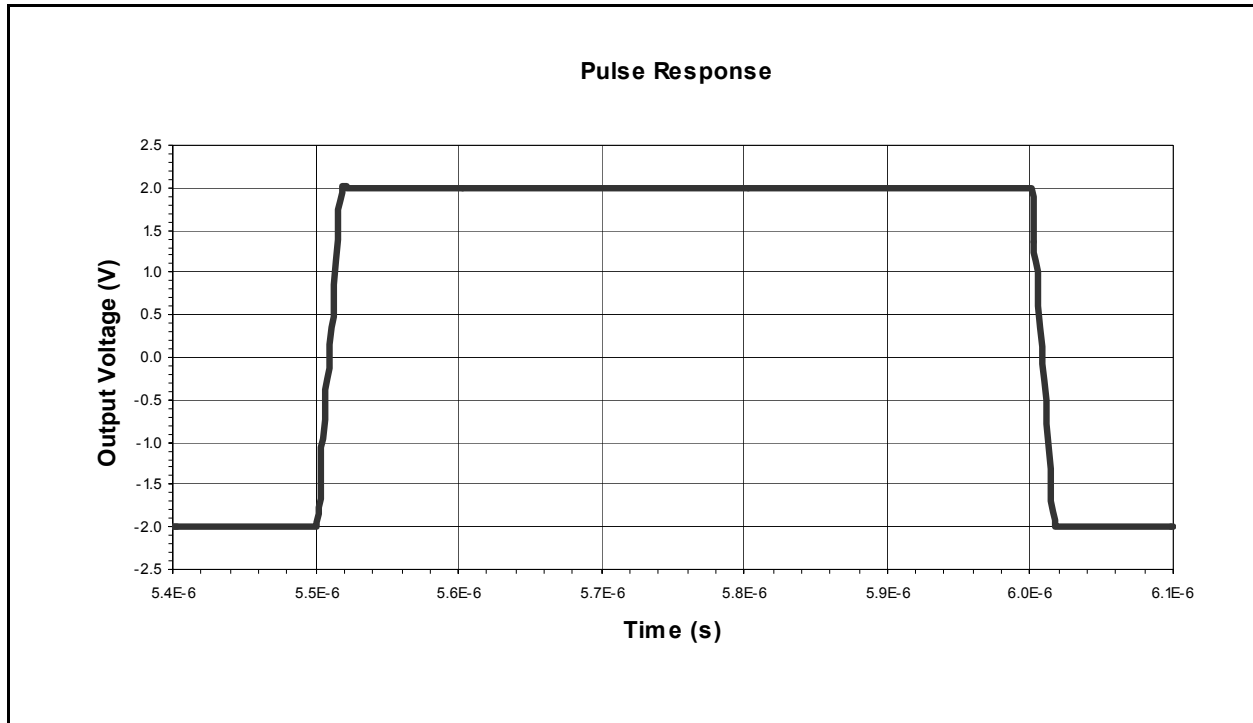


Figure 40 - Pulse Response

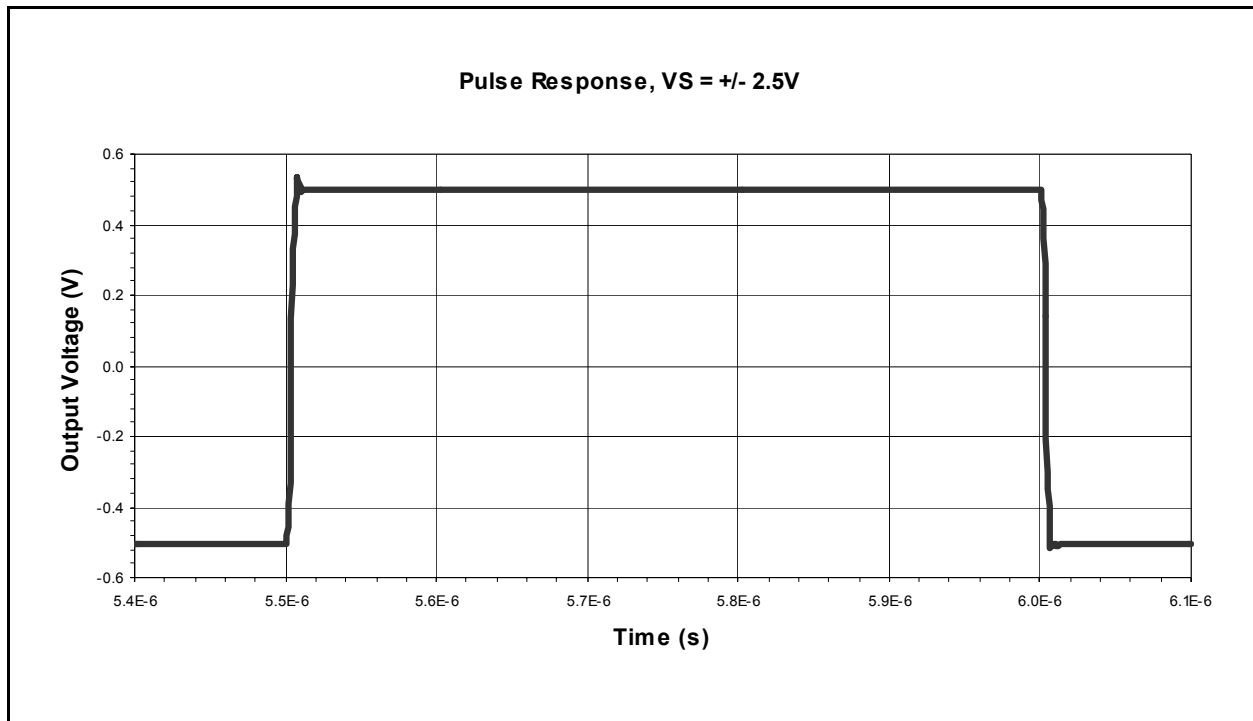


Figure 41 - Pulse Response, $V_S = \pm 2.5\text{V}$

Typical Performance Characteristics At TA = 25°C, RF = RG = 510, gain = +2, RL = 100, Vs = 6V. Unless otherwise specified.

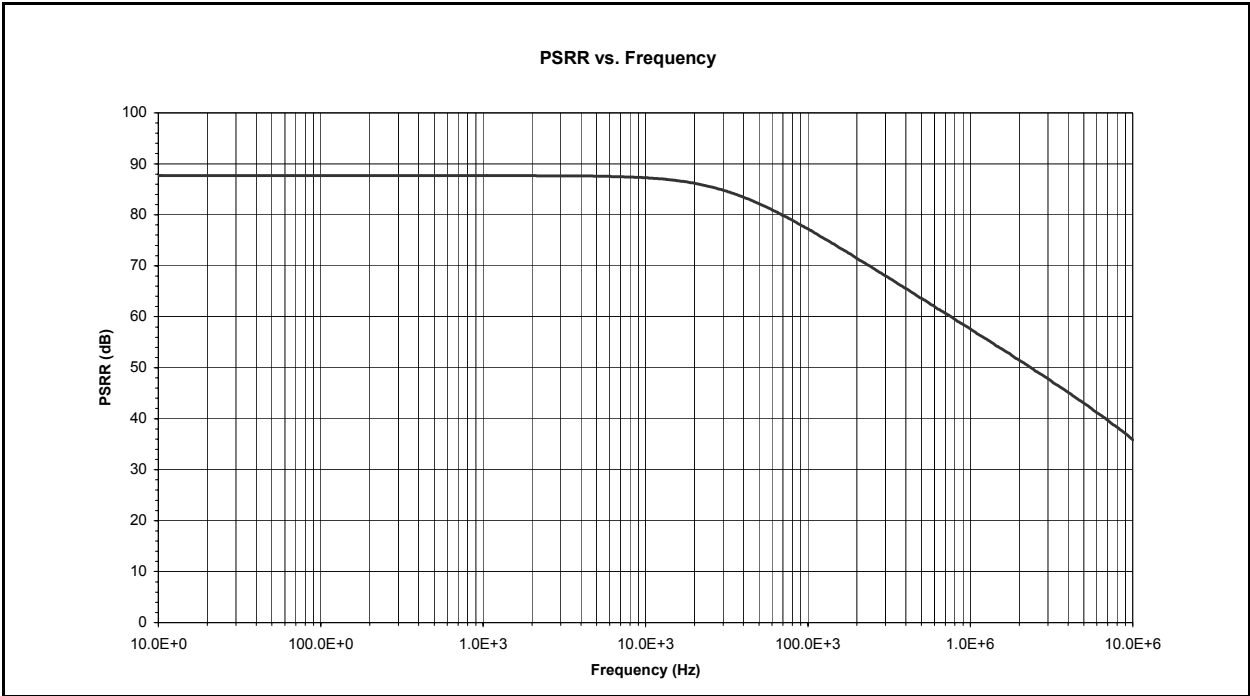


Figure 42 - PSRR vs Frequency

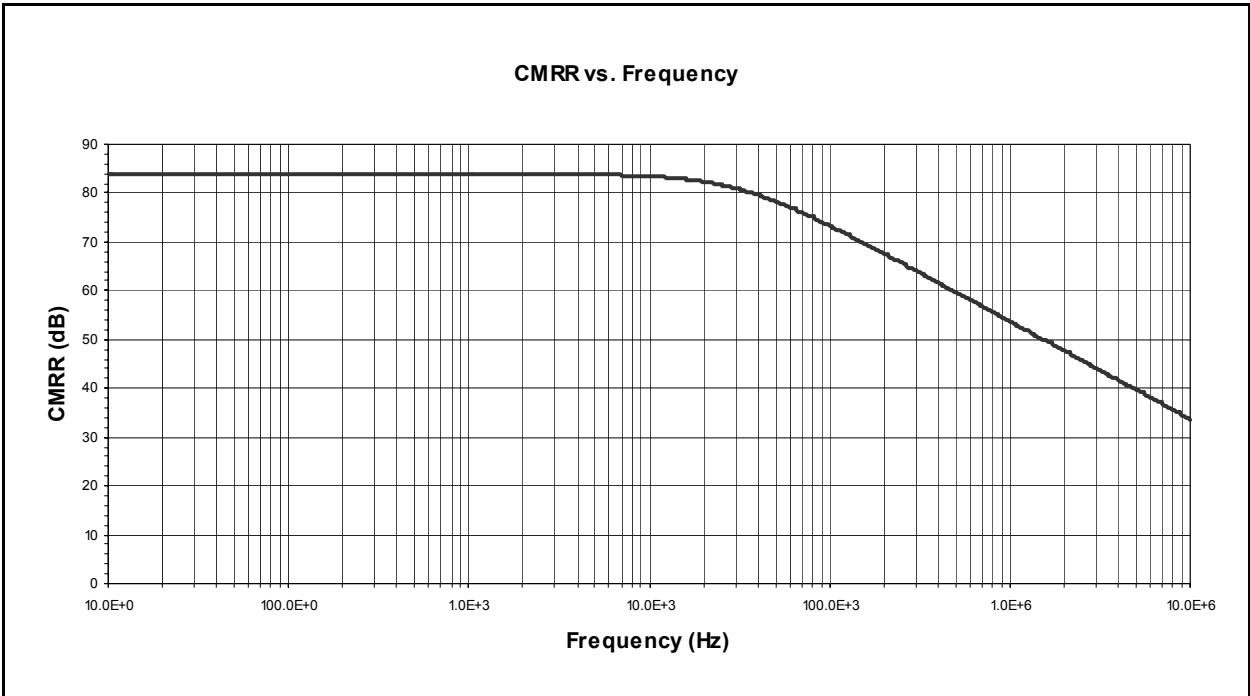


Figure 43 - CMRR vs. Frequency

Typical Performance Characteristics At TA = 25°C, RF = RG = 510, gain = +2, RL = 100, Vs = 6V. Unless otherwise specified.

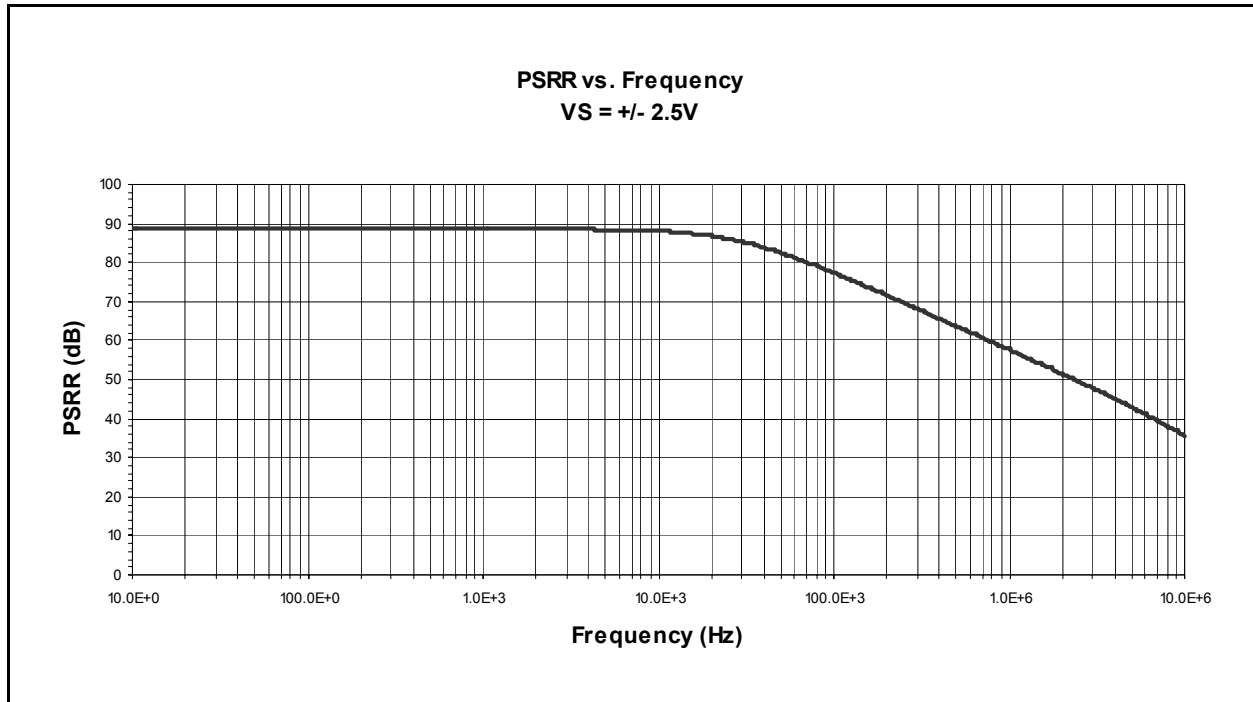


Figure 44 - PSRR vs. Frequency Vs = ±2.5V

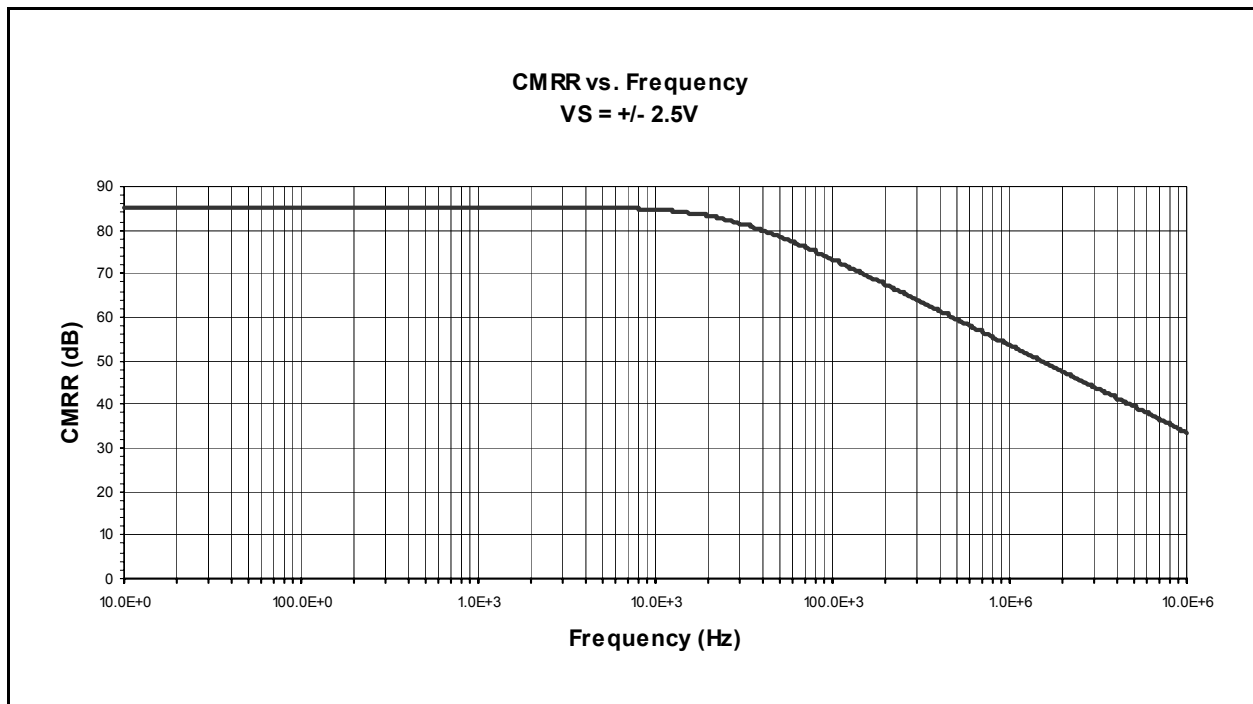


Figure 45 - CMRR vs. Frequency Vs = ±2.5V

Typical Performance Characteristics At $T_A = 25^\circ\text{C}$, $R_F = R_G = 510$, $\text{gain} = +2$, $R_L = 100$, $V_S = 6\text{V}$. Unless otherwise specified.

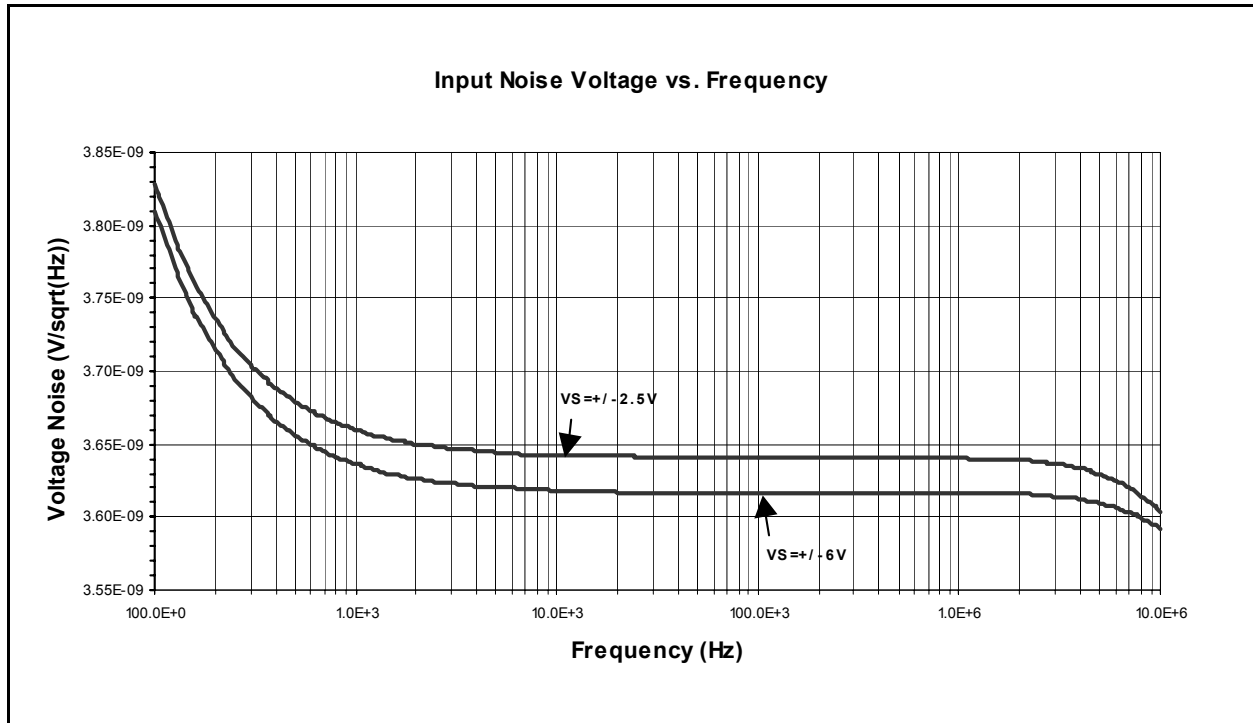


Figure 46 - Noise Voltage vs. Frequency

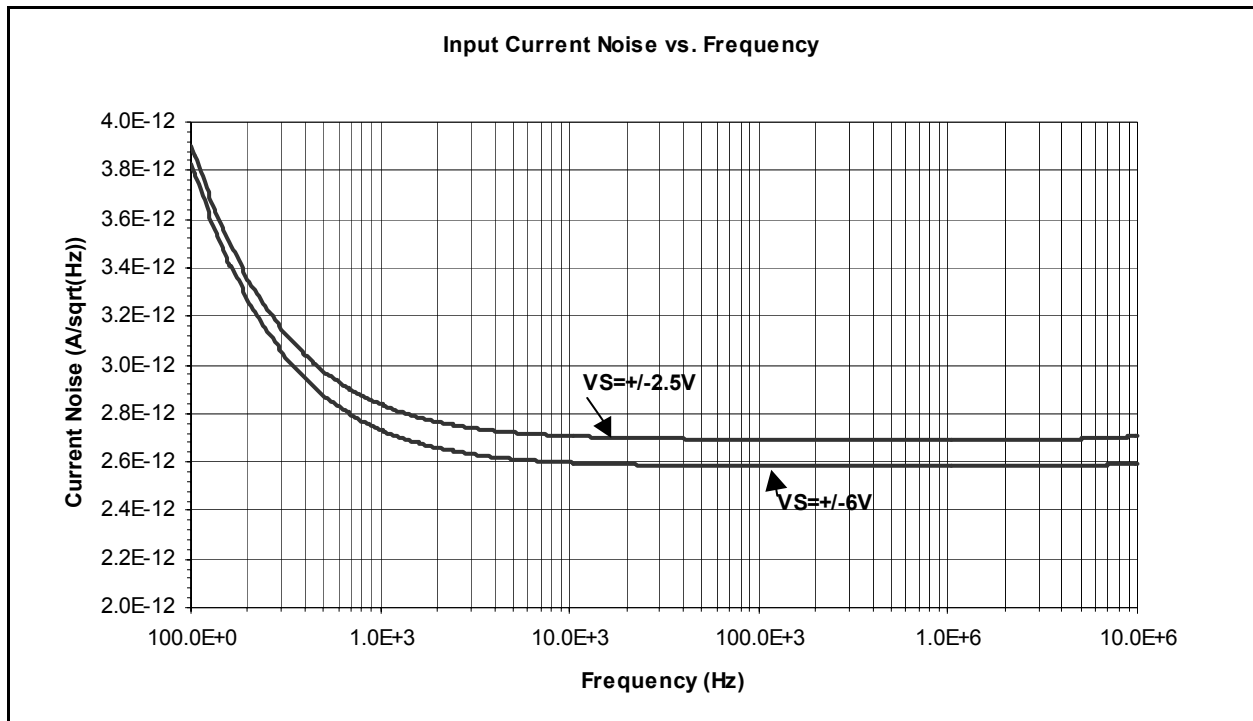
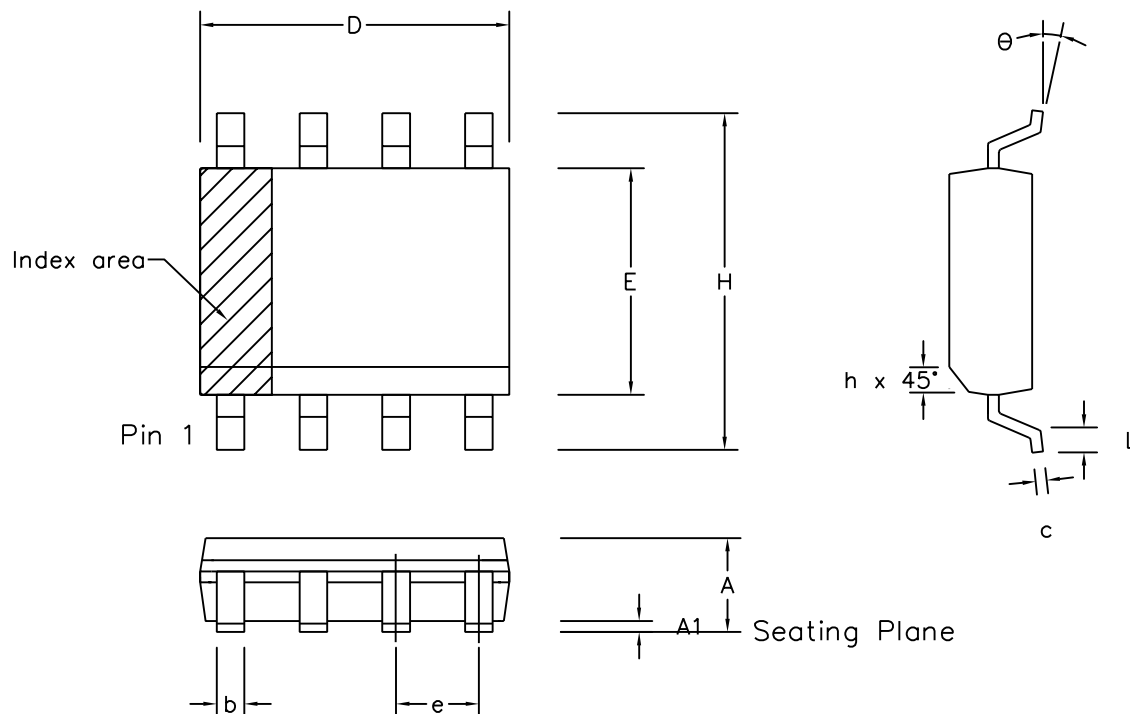


Figure 47 - Current Noise vs. Frequency



	Min mm	Max mm	Min inch	Max inch
A	1.35	1.75	0.053	0.069
A1	0.10	0.25	0.004	0.010
D	4.80	5.00	0.189	0.197
H	5.80	6.20	0.228	0.244
E	3.80	4.00	0.150	0.157
L	0.40	1.27	0.016	0.050
e	1.27 BSC		0.050 BSC	
b	0.33	0.51	0.013	0.020
c	0.19	0.25	0.008	0.010
O	0°	8°	0°	8°
h	0.25	0.50	0.010	0.020
	Pin Features			
N	8		8	
Conforms to JEDEC MS-012AA Iss. C				

Notes:

1. The chamfer on the body is optional. If not present, a visual index feature, e.g. a dot, must be located within the cross-hatched area.
2. Controlling dimensions are in inches.
3. Dimension D do not include mould flash, protusion or gate burrs. These shall not exceed 0.006" per side.
4. Dimension $E1$ do not include inter-lead flash or protusion. These shall not exceed 0.010" per side.
5. Dimension b does not include dambar protusion / intrusion. Allowable dambar protusion shall be 0.004" total in excess of b dimension.

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ISSUE	1	2	3	4	5	Previous package codes MP / S	Package Outline for 8 lead SOIC (0.150" Body width)
ACN	6745	201936	202595	203705	212424		
DATE	5Apr95	27Feb97	12Jun97	9Dec97	22Mar02		
APPRD.							GPD000010





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