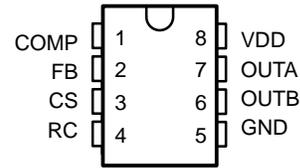
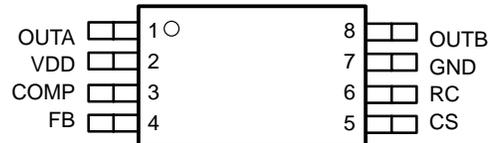


- Dual Output Drive Stages in Push-Pull Configuration
- Current Sense Discharge Transistor to Improve Dynamic Response
- 130- $\mu$ A Typical Starting Current
- 1-mA Typical Run Current
- Operation to 1 MHz
- Internal Soft Start
- On-Chip Error Amplifier With 2-MHz Gain Bandwidth Product
- On Chip VDD Clamping
- Output Drive Stages Capable of 500-mA Peak-Source Current, 1-A Peak-Sink Current

**D OR N PACKAGE (TOP VIEW)**



**PW PACKAGE (TOP VIEW)**

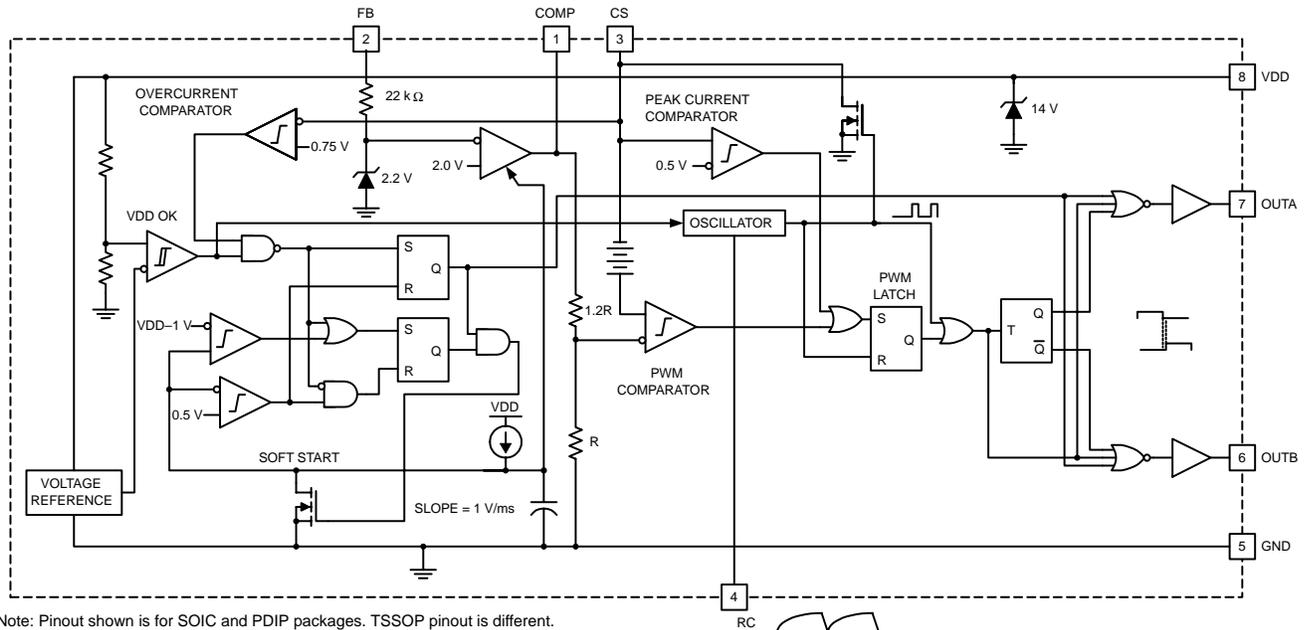


**description**

The UCC3808A is a family of BiCMOS push-pull, high-speed, low-power, pulse-width modulators. The UCC3808A contains all of the control and drive circuitry required for off-line or dc-to-dc fixed frequency current-mode switching power supplies with minimal external parts count.

The UCC3808A dual output drive stages are arranged in a push-pull configuration. Both outputs switch at half the oscillator frequency using a toggle flip-flop. The dead time between the two outputs is typically 60 ns to 200 ns depending on the values of the timing capacitor and resistors, thus limiting each output stage duty cycle to less than 50%.

**block diagram**



Note: Pinout shown is for SOIC and PDIP packages. TSSOP pinout is different.

UDG-00097



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

# UCC2808A-1, UCC2808A-2, UCC3808A-1, UCC3808A-2 LOW POWER CURRENT MODE PUSH-PULL PWM

SLUS456D – APRIL 1999 - REVISED AUGUST 2002

## description (continued)

The UCC3808A family offers a variety of package options, temperature range options, and choice of undervoltage lockout levels. The family has UVLO thresholds and hysteresis options for off-line and battery powered systems. Thresholds are shown in the table below.

The UCC3808A is an enhanced version of the UCC3808 family. The significant difference is that the A versions feature an internal discharge transistor from the CS pin to ground, which is activated each clock cycle during the oscillator dead time. The feature discharges any filter capacitance on the CS pin during each cycle and helps minimize filter capacitor values and current sense delay.

### ORDERING INFORMATION

T <sub>A</sub> = T <sub>J</sub>	Packaged Devices			
	UVLO Option	SOIC (D)	PDIP (N)	TSSOP (PW)
-40°C to 85°C	12.5 V/8.3 V	UCC2808AD-1	UCC2808AN-1	UCC2808APW-1
	4.3 V/4.1 V	UCC2808AD-2	UCC2808AN-2	UCC2808APW-2
0°C to 70°C	12.5 V/8.3 V	UCC3808AD-1	UCC3808AN-1	UCC3808APW-1
	4.3 V/4.1 V	UCC3808AD-2	UCC3808AN-2	UCC3808APW-2

† D (SOIC-8) and PW (TSSOP-8) packages are available taped and reeled. Add TR suffix to device type (e.g. UCC3808ADTR-1) to order quantities of 2500 devices per reel for SOIC-8 and 2000 devices per reel for TSSOP-8.

## absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage (I <sub>DD</sub> ≤ 10 mA)	15 V
Supply current	20 mA
OUTA/OUTB source current (peak)	-0.5 A
OUTA/OUTB sink current (peak)	1.0 A
Analog inputs (FB, CS)	-0.3 V to V <sub>DD</sub> 0.3 V, not to exceed 6 V
Power dissipation at T <sub>A</sub> = 25°C (N package)	1 W
Power dissipation at T <sub>A</sub> = 25°C (D package)	650 mW
Power dissipation at T <sub>A</sub> = 25°C (PW package)	400 mW
Storage temperature, T <sub>stg</sub>	-65°C to 150°C
Junction temperature, T <sub>J</sub>	-55°C to 150°C
Lead temperature (soldering, 10 sec.)	300°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

† Currents are positive into, negative out of the specified terminal. Consult Packaging Section of the *Power Supply Control Data Book (TI Literature Number SLUD003)* for thermal limitations and considerations of packages.

## electrical characteristics, T<sub>A</sub> = 0°C to 70°C for the UCC3808A-x, -40°C to 85°C for the UCC2808A-x, V<sub>DD</sub> = 10 V (see Note 6), 1-μF capacitor from V<sub>DD</sub> to GND, R = 22 kΩ, C = 330 pF T<sub>A</sub> = T<sub>J</sub>, (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
<b>Oscillator Section</b>					
Oscillator frequency		175	194	213	kHz
Oscillator amplitude/V <sub>DD</sub>	See Note 1	0.44	0.5	0.56	V/V

NOTES: 1. Measured at RC. Signal amplitude tracks V<sub>DD</sub>.  
6. For UCCx808A-1, set V<sub>DD</sub> above the start threshold before setting at 10 V.

# UCC2808A-1, UCC2808A-2, UCC3808A-1, UCC3808A-2

## LOW POWER CURRENT MODE PUSH-PULL PWM

SLUS456D – APRIL 1999 - REVISED AUGUST 2002

electrical characteristics,  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$  for the UCC3808A-x,  $-40^\circ\text{C}$  to  $85^\circ\text{C}$  for the UCC2808A-x,  $V_{DD} = 10\text{ V}$  (see Note 6),  $1\text{-}\mu\text{F}$  capacitor from VDD to GND,  $R = 22\text{ k}\Omega$ ,  $C = 330\text{ pF}$ ,  $T_A = T_J$ , (unless otherwise noted)

Error Amplifier Section					
Input voltage	COMP = 2 V	1.95	2	2.05	V
Input bias current		-1		1	$\mu\text{A}$
Open loop voltage gain		60	80		dB
COMP sink current	FB = 2.2 V, COMP = 1 V	0.3	2.5		mA
COMP source current	FB = 1.3 V, COMP = 3.5 V	-0.2	-0.5		mA
PWM Section					
Maximum duty cycle	Measured at OUTA or OUTB	48	49	50	%
Minimum duty cycle	COMP = 0 V			0	%
Current Sense Section					
Gain	See Note 2	1.9	2.2	2.5	V/V
Maximum input signal	COMP = 5 V See Note 3	0.45	0.5	0.55	V
CS to output delay	COMP = 3.5 V, CS from 0 mV to 600 mV		100	200	ns
CS source current		-200			nA
CS sink current	CS = 0.5 V, RC = 5.5 V See Note 7	5	10		mA
Over current threshold		0.7	0.75	0.8	V
COMP to CS offset	CS = 0 V	0.35	0.8	1.2	V
Output Section					
OUT low level	I = 100 mA		0.5	1	V
OUT high level	I = -50 mA, VDD - OUT		0.5	1	V
Rise time	$C_L = 1\text{ nF}$		25	60	ns
Fall time	$C_L = 1\text{ nF}$		25	60	ns
Undervoltage Lockout Section					
Start threshold	UCCx808A-1 See Note 6	11.5	12.5	13.5	V
	UCCx808A-2	4.1	4.3	4.5	V
Minimum operating voltage after start	UCCx808A-1	7.6	8.3	9	V
	UCCx808A-2	3.9	4.1	4.3	V
Hysteresis	UCCx808A-1	3.5	4.2	5.1	V
	UCCx808A-2	0.1	0.2	0.3	V
Soft Start Section					
COMP rise time	FB = 1.8 V, Rise from 0.5 V to 4 V		3.5	20	ms
Overall Section					
Startup current	VDD < start threshold		130	260	$\mu\text{A}$
Operating supply current	FB = 0 V, CS = 0 V See Note 5 and 6		1	2	mA
VDD zener shunt voltage	IDD = 10 mA See Note 4	13	14	15	V

NOTES: 2. Gain is defined by:  $A = \frac{\Delta V_{COMP}}{\Delta V_{CS}}$ ,  $0 \leq V_{CS} \leq 0.4\text{ V}$ .

3. Parameter measured at trip point of latch with FB at 0 V.
4. Start threshold and zener shunt threshold track one another.
5. Does not include current in the external oscillator network.
6. For UCCx808A-1, set VDD above the start threshold before setting at 10 V.
7. The internal current sink on the CS pin is designed to discharge an external filter capacitor. It is not intended to be a dc sink path.

# UCC2808A-1, UCC2808A-2, UCC3808A-1, UCC3808A-2

## LOW POWER CURRENT MODE PUSH-PULL PWM

SLUS456D – APRIL 1999 - REVISED AUGUST 2002

---

### pin assignments

**COMP:** COMP is the output of the error amplifier and the input of the PWM comparator. The error amplifier in the UCC3808A is a true low-output impedance, 2-MHz operational amplifier. As such, the COMP pin can both source and sink current. However, the error amplifier is internally current limited, so that zero duty cycle can be externally forced by pulling COMP to GND.

The UCC3808A family features built-in full-cycle soft start. Soft start is implemented as a clamp on the maximum COMP voltage.

**CS:** The input to the PWM, peak current, and overcurrent comparators. The overcurrent comparator is only intended for fault sensing. Exceeding the overcurrent threshold will cause a soft start cycle. An internal MOSFET discharges the current sense filter capacitor to improve dynamic performance of the power converter.

**FB:** The inverting input to the error amplifier. For best stability, keep FB lead length as short as possible and FB stray capacitance as small as possible.

**GND:** Reference ground and power ground for all functions. Due to high currents, and high frequency operation of the UCC3808A, a low impedance circuit board ground plane is highly recommended.

**OUTA and OUTB:** Alternating high current output stages. Both stages are capable of driving the gate of a power MOSFET. Each stage is capable of 500-mA peak-source current, and 1-A peak-sink current.

The output stages switch at half the oscillator frequency, in a push-pull configuration. When the voltage on the RC pin is rising, one of the two outputs is high, but during fall time, both outputs are off. This *dead time* between the two outputs, along with a slower output rise time than fall time, insures that the two outputs can not be on at the same time. This dead time is typically 60 ns to 200 ns and depends upon the values of the timing capacitor and resistor.

The high-current-output drivers consist of MOSFET output devices, which switch from VDD to GND. Each output stage also provides a very low impedance to overshoot and undershoot. This means that in many cases, external-schottky-clamp diodes are not required.

**RC:** The oscillator programming pin. The UCC3808A's oscillator tracks VDD and GND internally, so that variations in power supply rails minimally affect frequency stability. Figure 1 shows the oscillator block diagram.

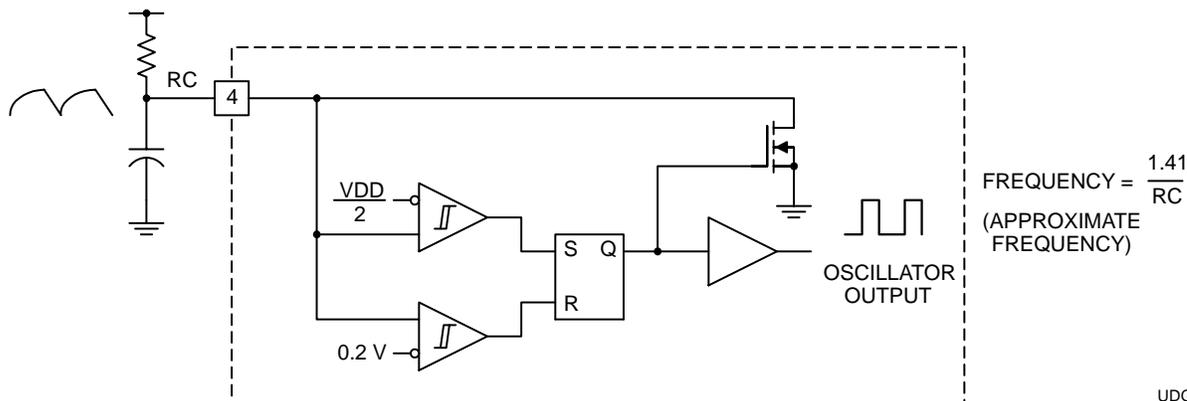
Only two components are required to program the oscillator: a resistor (tied to the VDD and RC), and a capacitor (tied to the RC and GND). The approximate oscillator frequency is determined by the simple formula:

$$f_{\text{OSCILLATOR}} = \frac{1.41}{RC}$$

where frequency is in Hz, resistance in Ohms, and capacitance in Farads. The recommended range of timing resistors is between 10 kΩ and 200 kΩ and range of timing capacitors is between 100 pF and 1000 pF. Timing resistors less than 10 kΩ should be avoided.

For best performance, keep the timing capacitor lead to GND as short as possible, the timing resistor lead from VDD as short as possible, and the leads between timing components and RC as short as possible. Separate ground and VDD traces to the external timing network are encouraged.

**pin assignments (continued)**



UDG-00095

**Figure 1. Block Diagram for Oscillator**

NOTE A: The oscillator generates a sawtooth waveform on RC. During the RC rise time, the output stages alternate on time, but both stages are off during the RC fall time. The output stages switch a 1/2 the oscillator frequency, with ensured duty cycle of < 50% for both outputs.

**VDD:** The power input connection for this device. Although quiescent VDD current is very low, total supply current will be higher, depending on OUTA and OUTB current, and the programmed oscillator frequency. Total VDD current is the sum of quiescent VDD current and the average OUT current. Knowing the operating frequency and the MOSFET gate charge (Qg), average OUT current can be calculated from:

$$I_{OUT} = Q_g \times F, \text{ where } F \text{ is frequency}$$

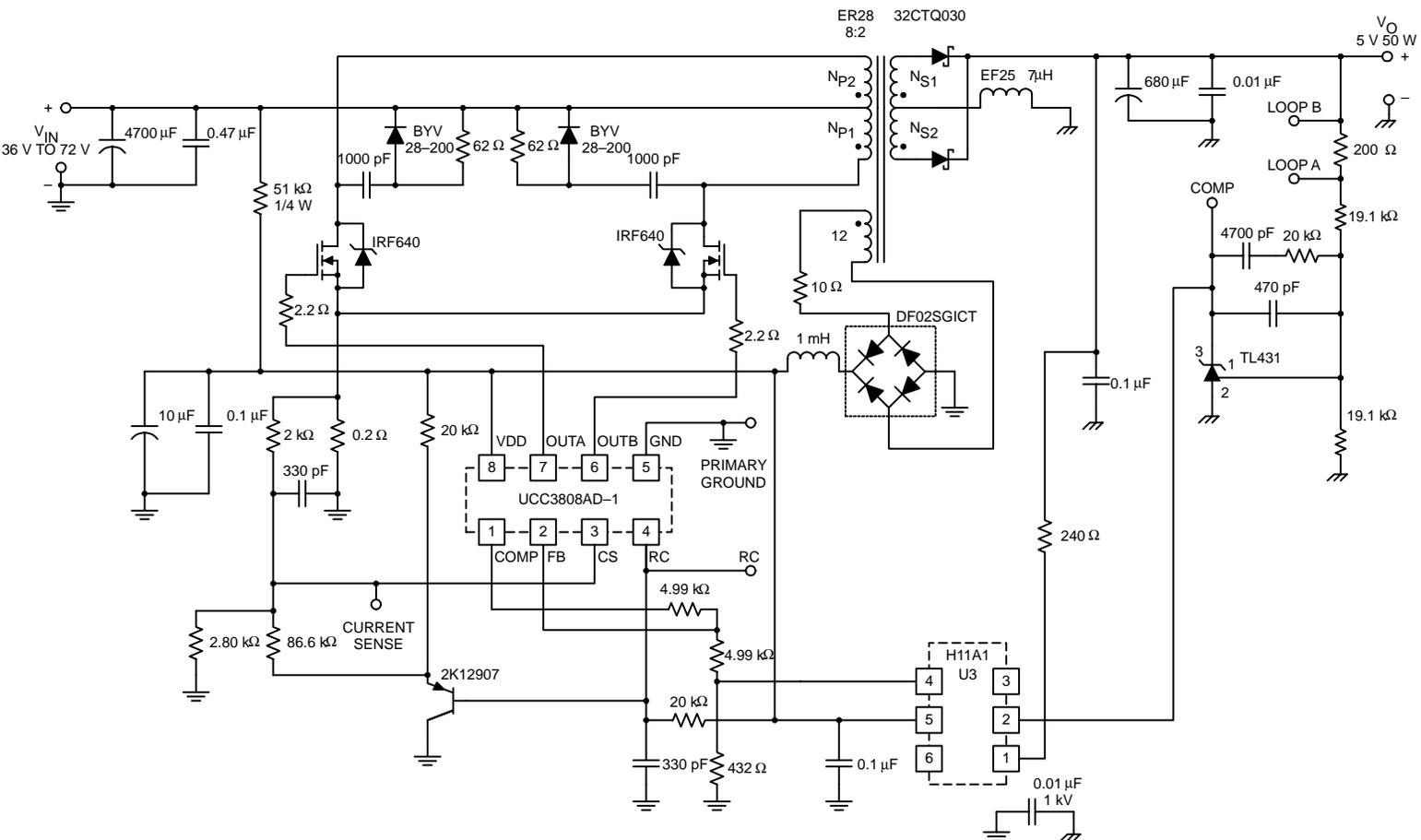
To prevent noise problems, bypass VDD to GND with a ceramic capacitor as close to the chip as possible along with an electrolytic capacitor. A 1- $\mu$ F decoupling capacitor is recommended.

**APPLICATION INFORMATION**

A 200-kHz push-pull application circuit with a full-wave rectifier is shown in Figure 2. The output,  $V_O$ , provides 5 V at 50 W maximum and is electrically isolated from the input. Since the UCC3808A is a peak-current-mode controller the 2N2907 emitter following amplifier (buffers the CT waveform) provides slope compensation which is necessary for duty ratios greater than 50%. Capacitor decoupling is very important with a single ground IC controller, and a 1  $\mu$ F is suggested as close to the IC as possible. The controller supply is a series RC for start-up, paralleled with a bias winding on the output inductor used in steady state operation.

Isolation is provided by an optocoupler with regulation done on the secondary side using the TL431 adjustable precision shunt regulator. Small signal compensation with tight voltage regulation is achieved using this part on the secondary side. Many choices exist for the output inductor depending on cost, volume, and mechanical strength. Several design options are iron powder, molypermalloy (MPP), or a ferrite core with an air gap as shown here. The main power transformer has a Magnetics Inc. ER28 size core made of P material for efficient operation at this frequency and temperature. The input voltage may range from 36 V dc to 72 V dc.

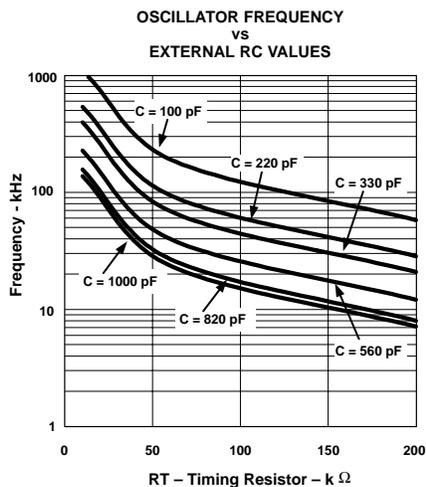
APPLICATION INFORMATION



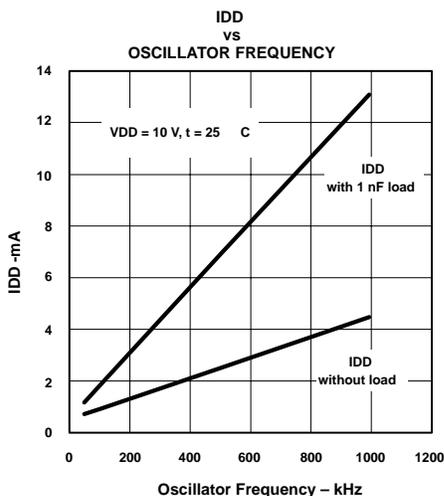
UDG-00096

Figure 2. Typical Application Diagram: 48-V In, 5-V, 50-W Output

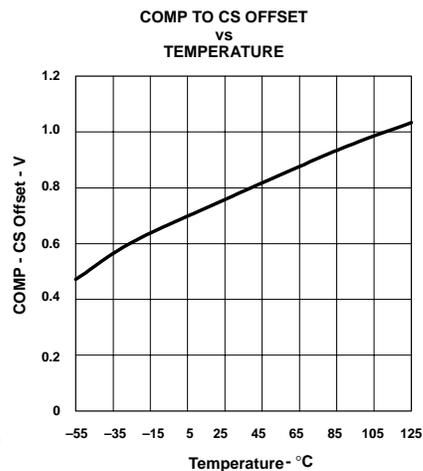
### TYPICAL CHARACTERISTICS



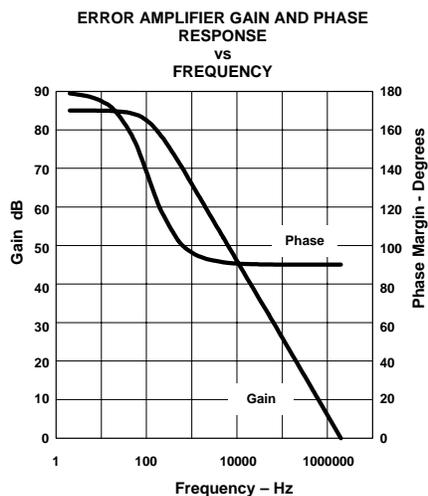
**Figure 3**



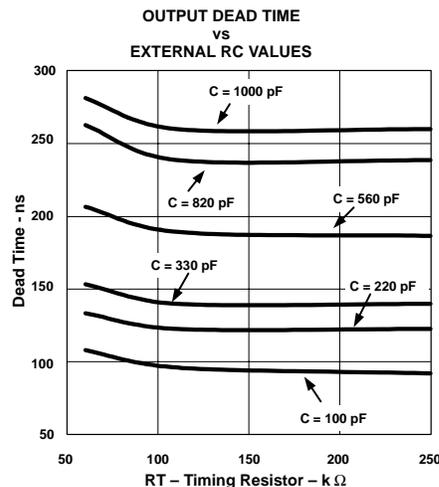
**Figure 4**



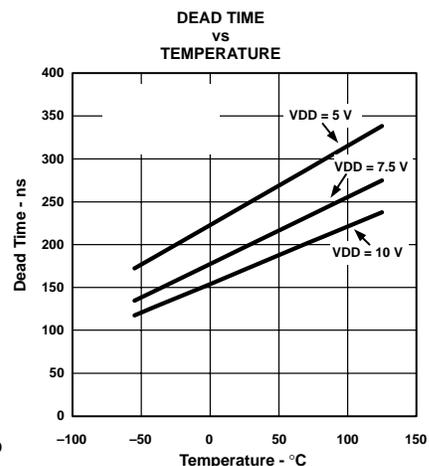
**Figure 5**



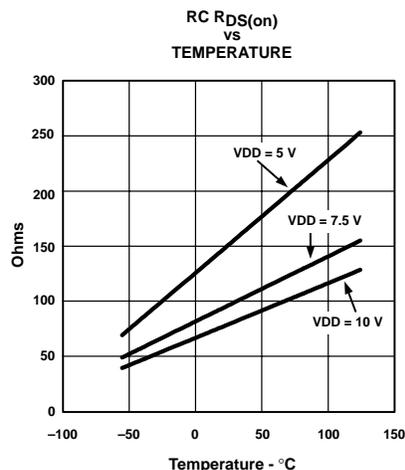
**Figure 6**



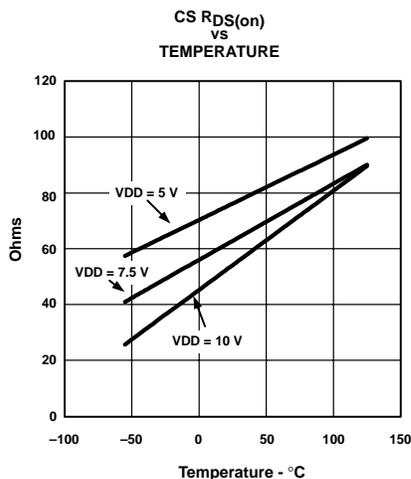
**Figure 7**



**Figure 8**



**Figure 9**



**Figure 10**

## IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All products are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

TI assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using TI components. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any TI patent right, copyright, mask work right, or other TI intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information published by TI regarding third-party products or services does not constitute a license from TI to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. Reproduction of this information with alteration is an unfair and deceptive business practice. TI is not responsible or liable for such altered documentation.

Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

### Mailing Address:

Texas Instruments  
Post Office Box 655303  
Dallas, Texas 75265