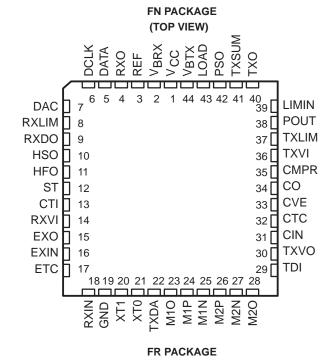
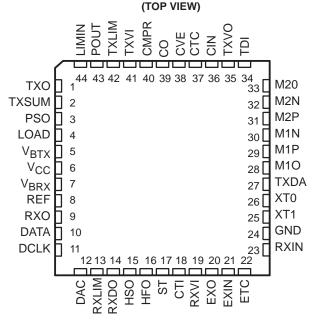
- Switchable AMPS/TACS and NMT Operation
- Integrated RX and TX Voice Filters
- Integrated RX and TX Data Filters
- Narrow-Band RX SAT Filter
- Pre-Emphasis and De-Emphasis Filtering
- CCITT-Compatible Compandor
- Adjustable TX and RX Limiters
- Loudspeaker Driver
- Microphone Preamplifiers
- Digitally Controlled Gains and Signal Muting
- Low Power
- Standby Mode
- Simple 3-Wire Digital Interface
- Low External Component Count
- Single 5-V Supply
- 44-Pin PLCC Package and Thin Quad Flat Pack

## description

Implemented in advanced LinBiCMOS™ technology, the TCM8000 audio processor provides a highly integrated solution for voiceband signal processing in FM cellular mobile and hand-portable telephones. The device incorporates the necessary voice and data filters as well as ancillary functions, such as microphone preamplifiers, a loudspeaker driver, and a CCITT-compatible compandor circuit. A simple 3-wire serial interface provides digital control over signal path switching, muting, and gain adjustment. The filter responses can be switched to suit Advanced





Mobile Phone System/Total Access Communication System (AMPS/TACS) and Nordic Mobile Telephone (NMT) system requirements. Switched-capacitor techniques are used to implement the filtering functions, and appropriate antialiasing and smoothing filters are incorporated in the device.

In the active mode, the TCM8000 uses less than 14 mA of supply current and can be set into a standby configuration, which reduces the supply current to less than 4 mA.

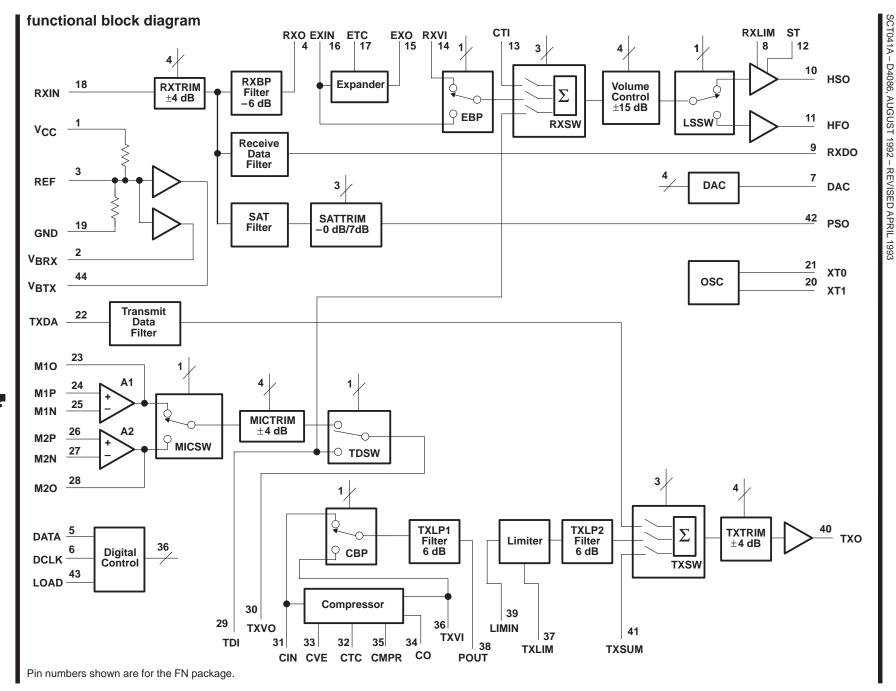
The TCM8000 is characterized for operation from  $-25^{\circ}$ C to  $80^{\circ}$ C.



Caution. These devices have limited built-in protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

LinBiCMOS is a trademark of Texas Instruments Incorporated.





# **Terminal Functions**

NAME   NO.   FR   FR	
CIN 31 36 I Compressor analog input  CMPR 35 40 I Compressor rectifier analog input  CO 34 39 O Compressor analog output. CO should be ac coupled to TXVI and to CMPR.  CTC 32 27 O Compressor time constant control, analog output. A 220-nF capacitor should be connected bet GND.  CTI 13 18 I Call tone analog input. Signals on CTI are summed into the receive voice path prior to the circuit. CTI can be muted under control of the serial interface.  CVE 33 38 I Compressor virtual earth, analog input  DAC 7 12 O Digital-to-analog converter output, programmable between GND and V <sub>CC</sub> /2 in sixteen steps DATA 5 10 I Serial interface digital data input  DCLK 6 11 I Serial interface digital clock input  ETC 17 22 O Expander time constant control, analog output. A 220-nF capacitor should be connected bet GND.  EXIN 16 21 I Expander analog input  EXO 15 20 O Expander analog output. EXO should be ac coupled to RXVI.  GND 19 24 Ground	
CMPR 35 40 I Compressor rectifier analog input  CO 34 39 O Compressor analog output. CO should be ac coupled to TXVI and to CMPR.  CTC 32 27 O Compressor time constant control, analog output. A 220-nF capacitor should be connected bet GND.  CTI 13 18 I Call tone analog input. Signals on CTI are summed into the receive voice path prior to the circuit. CTI can be muted under control of the serial interface.  CVE 33 38 I Compressor virtual earth, analog input  DAC 7 12 O Digital-to-analog converter output, programmable between GND and V <sub>CC</sub> /2 in sixteen steps DATA 5 10 I Serial interface digital data input  DCLK 6 11 I Serial interface digital clock input  ETC 17 22 O Expander time constant control, analog output. A 220-nF capacitor should be connected bet GND.  EXIN 16 21 I Expander analog input  EXO 15 20 O Expander analog output. EXO should be ac coupled to RXVI.  GND 19 24 Ground	
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GND.	
EXO 15 20 O Expander analog output. EXO should be ac coupled to RXVI.  GND 19 24 Ground	ween ETC and
GND 19 24 Ground	
HFO 11 16 O Normal drive receiver voice analog output	
HSO 10 15 O High drive receiver voice analog output	
LIMIN 39 44 I Transmit limiter analog input	
LOAD 43 4 I Serial interface load, digital input. Data shifted into DATA under control of DCLK is transfe registers when LOAD goes high.	rred to internal
M1N 25 30 I Preamplifier 1 negative analog input	
M1O 23 28 O Preamplifier 1 analog output	
M1P 24 29 I Preamplifier 1 positive analog input	
M2N 27 32 I Preamplifier 2 negative analog input	
M2O 28 33 O Preamplifier 2 analog output	
M2P 26 31 I Preamplifier 2 positive analog input	
POUT 38 43 O Preemphasis filter analog output. POUT should be ac coupled to LIMIN.	
PSO 42 3 O Analog output of SAT filter	
REF 3 8 Unbuffered midsupply voltage, nominal output V <sub>CC</sub> /2	
RXDO 9 14 O Receive data filter analog output	
RXIN 18 23 I Receive analog input	
RXLIM 8 13 I/O Receive limiter adjust voltage, controls maximum signal output on HSO. The dc input is betw V <sub>CC</sub> /2.	ween GND and
RXO 4 9 O Receive voice filter analog output	
RXVI 14 19 I Receive voice analog input to volume control circuit. RXVI can be muted under control of the	serial interface.
ST 12 17 I Side tone analog input. Signals on this terminal are summed into the HSO output signal.	
TDI 29 34 I Tone of DTMF analog input. Signals on this terminal can be used in place of the main voice s path and summed into the RX voice path, under control of the serial interface.	signal in the TX
TXDA 22 27 I Transmit data analog input	
TXLIM 37 42 I Transmit-limiter level adjust, controls maximum signal level at TXO. The dc input is between GN	ND and V <sub>CC</sub> /2.
TXO 40 1 O Transmit section analog output	
TXSUM 41 2 I Transmit summing analog input. The signal on TXSUM is summed into the main TX output after TXSUM can be muted under control of the serial interface.	all filter stages.
TXVI 36 41 I Transmit voice analog input	
TXVO 30 35 O Transmit voice analog output. TXVO is the output from the preamplifier and microphone gain a lt should be ac coupled into the compressor input CIN.	adjust sections.
V <sub>BRX</sub> 2 7 Buffered midsupply voltage to receive sections, nominal output V <sub>CC</sub> /2	



### **Terminal Functions (continued)**

	PIN			
NAME	N	D. I/O		DESCRIPTION
NAME	FN FR			
V <sub>BTX</sub>	44	5		Buffered midsupply for all transmit sections, nominal output V <sub>CC</sub> /2
VCC	1	6		Most positive supply voltage, 5 V ±5%
XT0, XT1	21,20	26,25		Crystal oscillator terminals for connection to 3.58-MHz crystal resonator

### detailed description - transmit path

### input conditioning

A pair of uncommitted operational amplifiers is provided at the main audio signal input. These components can be configured using external resistors to adjust the gain as required to suit particular microphones and external preamplifiers. The TCM8000 has been designed to provide 100 mVrms at the outputs of the preamplifier at pins M1O and M2O. The switch MICSW allows the selection of either amplifier input for hands-free or handset microphones. A digitally controlled gain/attenuation block, MICTRIM, allows adjustment of the signal level into the compressor in sixteen 0.5-dB steps. A digitally controlled switch, TDSW, allows alternative signals from other sources such as audio or DTMF tones to be injected into the transmit signal path from TDI. The output of TDSW is accoupled at TXVO through an external capacitor into the compressor.

### compressor

The compressor provides a 2:1 dynamic range compression of its input signal, converting an input range of 60 dB to 30 dB at its output. The test circuit diagram (Figure 2) shows the external components required to set time constants and to limit gain under idle channel conditions. The envelope time constant is determined by the value of a capacitor attached to CTC with 220 nF providing attack time of 3 ms approximately and decay times of approximately 13 ms. The compressor can be bypassed under control of the serial interface. The output from this stage, CO, is externally ac coupled into the transmit filter at TXVI and into the compressor rectifier input at CMPR.

#### TX filtering and limiter

The transmit band limiter is located between a pair of band-pass filters. Filter TXF1 is located before the limiter and provides 6 dB per octave preemphasis; filter TXF2, located after the limiter, features a phase-equalized response to control overshoot. The frequency responses of TXF1 and TXF2 are switchable between 3 kHz and 3.4 kHz cut off to meet the requirements of AMPS/TACS and NMT systems, respectively. Frequency response templates and typical responses are shown in Figures 3 through 6.

The limiter is provided to meet maximum deviation specifications, and the limit level is controlled by the dc voltage applied to TXLIM. The output signal from POUT is ac coupled into the limiter input to eliminate asymmetrical limiting that could arise as a result of dc offsets.

### output conditioning

The voice output signal from TXLP2 is passed to a three-input summing circuit, whose other inputs come from the TX data path and from TXSUM. Each input to the summing circuit can be enabled or muted under control of the serial bus. The overall deviation level is controlled by a digitally controlled gain/attenuation block that provides sixteen 0.5-dB steps of output level adjustment. The transmit supervisory audio tone (SAT) is applied to TXSUM. TXSUM can also be used as an input for transmit data, allowing implementation of alternative data filtering and conditioning to that implemented on the TCM8000.

### transmit data

Transmit data, which is voice-band FSK coded for NMT and wide-band Manchester coded for AMPS and TACS (8K-baud TACS, 10K-baud AMPS), is applied to TXDA. The signal is filtered by the TX DATA filter, whose response can be switched from a band-pass to a low-pass configuration to meet the requirements of NMT and AMPS/TACS, respectively. Frequency templates and typical responses are shown in Figures 7 and 8.



### detailed description - receive path

### input conditioning

The received demodulated signal is applied to RXIN and passes to a gain-adjust block that provides  $\pm 4$  dB of adjustment in sixteen 0.5-dB steps. At this point, the receive signal path is split into the voice path, the data path, and the supervisory audio-tone (SAT) path.

### receive voice filter

The received voice signal is processed in the band-pass RXBP filter, whose response has been designed to meet the requirements of AMPS/TACS and NMT systems. The response incorporates the required 6-dB/octave de-emphasis. Frequency response templates and a typical response are shown in Figures 9 and 10.

## expander

The expander implements a 2:1 dynamic range expansion of the signal at its input, EXIN. It is designed to produce 0-dB gain at 100 mVrms. Envelope attack and decay times are determined by a capacitor attached to ETC, with a value of 220 nF producing attack time of 3 ms and decay times of approximately 13 ms. The expander can be bypassed under control of the serial interface.

### receive summing circuit and volume control

A 3-input summing circuit is provided, which takes as input the voice signal from the expander, audio tones from the TDI input, and a call-tone input applied to CTI. Each input can be enabled or muted under control of the serial interface. The output from this block is passed to the volume control, which provides  $\pm 15$  dB of output level control in sixteen 2-dB steps.

## output buffers

Two output buffer circuits are provided. The buffer that drives HFO is capable of driving loads down to 10-k $\Omega$  impedance and is intended for use with an external speaker driver. The HSO buffer features a high-drive bipolar output stage that allows impedances of  $500~\Omega$  to be driven directly with low distortion. The HSO output path also incorporates a summing circuit to facilitate side-tone injection from the ST input and a limiter circuit, which allows maximum sound pressure levels to be determined. The limit level is controlled by the voltage applied to RXLIM. Selection between the HFO and HSO outputs is achieved via the serial interface.

### receive data path

A switchable response filter is provided to condition receive data before it is output on RXDO. Frequency response templates and typical responses are shown in Figures 11 and 12 for AMPS/TACS and NMT, respectively.

### receive SAT path

Received supervisory audio tones (SAT) are separated from the voice signal in a switchable response narrow-band filter. This filter has a center frequency of 6 kHz in the AMPS/TACS mode and 4 kHz in the NMT mode. The filter output is passed to a level-adjust circuit, which provides eight 1-dB steps of adjustment. The processed SAT signal is output on PSO.

### miscellaneous

### midsupply

The analog midsupply, effectively signal ground, is derived from a resistive divider connected between  $V_{CC}$  and GND to produce  $V_{CC}/2$  at REF. This voltage is buffered by a pair of low output impedance amplifiers to produce midrail supplies for the receive and transmit sections at  $B_{BRX}$  and  $V_{BTX}$ , respectively, with separate midrails used to minimize coupling between the receive and transmit circuits. REF,  $V_{BTX}$ , and  $V_{BRX}$  should be decoupled to ground with capacitors physically mounted as close to the device as possible.  $V_{BTX}$  and  $V_{BRX}$  directly control the 0-dB point of the compressor and expander, respectively, so it is important that a stable and well-regulated supply is provided to  $V_{CC}$ .



### detailed description - miscellaneous

### crystal oscillator

A low-power crystal oscillator is provided to generate the master clock signal for the switched-capacitor filters. Sampling frequencies and internal divide ratios have been selected to allow a low cost 3.58-MHz crystal to be used. No external components except the crystal are required.

### DAC

A four-bit digital-to-analog (DAC) converter provides sixteen levels from GND to  $V_{BRX}$  in linear steps. The DAC can be used to set limit levels by controlling TXLIM or RXLIM. Its output is unbuffered and should only be used to drive high-impedance loads.

### standby mode

The standby mode disables all signal paths with the exception of the receive data path. In the standby mode, current consumption is reduced to less than 4 mA.

### digital interface

The TCM8000 is configured using a three-wire digital interface. Eight-bit words, comprised of four address and four data bits, are applied in serial to DATA and clocked into the device on the rising edge of DCLK. The data is transferred to internal registers by pulsing the LOAD signal high (see Figure 1). All internal registers are reset low on power up.

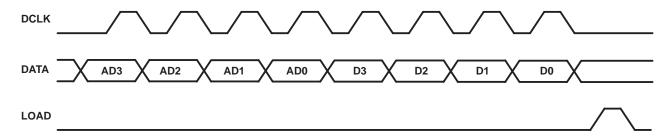


Figure 1. Serial Interface Timing Diagram



# **Table 1. Register Definitions**

AD <sub>3</sub>	AD <sub>2</sub>	AD <sub>1</sub>	AD <sub>0</sub>	D <sub>3</sub>	D <sub>2</sub>	D <sub>1</sub>	D <sub>0</sub>	FUNCTION
0	0	0	0	VC <sub>3</sub>	VC <sub>2</sub>	VC <sub>1</sub>	VC <sub>0</sub>	Volume control VC <sub>3</sub> -VC <sub>0</sub> set the output level on HFO and HSO in 2-dB steps from –16 dB <0000> to 14 dB <1111> relative to nominal, <1000>
0	0	0	1	TX3	TX <sub>2</sub>	TX <sub>1</sub>	TX <sub>0</sub>	TXTRIM gain adjust  TX <sub>3</sub> -TX <sub>0</sub> set the main transmit output level in 0.5-dB steps from -4 dB <0000> to 3.5 dB <1111> relative to nominal, <1000>
0	0	1	0	SBY	RTM	RVM	СТМ	SBY: Standby/active control, when set to 1, all sections of the device are active; when set to 0, all sections apart from from the receive data path are disabled.  RTM: Receive DTMF/tone mute, 1 = mute, 0 = enabled  RVM: Receive voice mute, 1 = mute, 0 = active  CTM: Call tone mute, 1 = mute, 0 = active
0	0	1	1	HSS	TSM	TVM	TDM	HSS: Handset select, 1 selects HSO RX output and M1 preamplifier, 0 selects HFO output and M2 preamplifier TSM: TXSUM mute, 1 = mute, 0 = active TVM: Transmit voice mute, 1 = mute, 0 = active TDM: Transmit data-path mute, 1 = mute, 0 = active
0	1	0	0	-	NAT	TTS	CEN	NAT: NMT-AMPS/TACS select, 1 = NMT, 0 = AMPS/TACS TTS: Transmit DTMF/tone select, 1 = DTMF/tone, 0 = TX voice CEN: Compandor enable, 1 = compressor/expander enabled, 0 = bypassed (0-dB gain)
0	1	0	1	RX3	RX <sub>2</sub>	RX <sub>1</sub>	RX <sub>0</sub>	RXTRIM gain adjust RX <sub>3</sub> -RX <sub>0</sub> set the main transmit output level in 0.5-dB steps from -4 dB <0000> to 3.5 dB <1111> relative to nominal, <1000>
0	1	1	0	STM	ST <sub>2</sub>	ST <sub>1</sub>	ST <sub>0</sub>	SATTRIM gain adjust ST <sub>3</sub> -ST <sub>0</sub> set the receive SAT gain in 1-dB steps from -4 dB <000> to 3 dB <111> relative to nominal, <100>, STM: Receive SAT path mute, 1 = mute, 0 = active
0	1	1	1	МТ3	MT <sub>2</sub>	MT <sub>1</sub>	MT <sub>0</sub>	MICTRIM gain adjust MT <sub>3</sub> -MT <sub>0</sub> set the microphone trim gain in 0.5-dB steps from -4 dB <0000> to 3.5 dB <1111> relative to nominal, <1000>
1	Х	Х	Х	DA <sub>3</sub>	DA <sub>2</sub>	DA <sub>1</sub>	DA <sub>0</sub>	DAC control DA3-DA0 set the level on the DAC output from 0 <0000> to VBRX <1111>

# absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

NOTE 1: Voltage value is with respect to GND.

# recommended operating conditions

		MIN	TYP	MAX	UNIT
Vcc	Supply voltage	4.75	5	5.25	V
VIH	High-level input voltage	3.5			V
VIL	Low-level input voltage			0.8	V
TA	Operating free-air temperature	- 25		80	°C

# electrical characteristics over recommended operating temperature range, $V_{CC} = 5 \text{ V}$ , $f_{xtal} = 3.58 \text{ MHz}$

	PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
loo	Cupply current	Standby			4	mA
ICC Su	Supply current	Operating			14	IIIA
lн	High-level input current	V <sub>I</sub> = 5 V		10	μΑ	
IլL	Low-level input current	V <sub>I</sub> = 0 V		10	μΑ	
fclock	Serial clock frequency, DCLK			200	kHz	
7.	Input impedance, RXIN, RXVI, CTI, ST, TXVI, LIMIN,	TXSUM, CIN, TXDA,	f = 1 kHz	70		kΩ
Zi	EXIN, CMPR, TDI		f = 1 kHz	20		K22
	Midsupply reference voltage, REF		2.4	2.6	V	
	Buffered midsupply reference voltage, VBTX, VBRX		2.4	2.6	V	
	Input current at M1P, M1N, M2P, M2N, RXLIM, TXLIM	Л			1	μΑ

# transmit path specifications, V<sub>CC</sub> = 5 V, f<sub>xtal</sub> = 3.58 MHz

# M1/M2 to TXVO, MICTRIM, A1 and A2 configured as unity gain inverting amplifiers, input = 100 mVrms at 1 kHz at M1 or M2 (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain, MICTRIM = <1000>		±1	dB
Step size	0.4	0.6	dB
Positive range	3	4	dB
Negative range	-4.5	-3.5	dB



<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# transmit path specifications, $V_{CC} = 5 \text{ V}$ , $f_{xtal} = 3.58 \text{ MHz}$ (continued)

M1/M2 to TXO frequency response, NMT mode, TXLIM = 2.5 V, input = 20 mVrms, TXTRIM = <1000>, MICTRIM = <1000>, compressor bypassed (see Figures 3 and 4 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 1 kHz	11.5	13.5	dB
	f = 100 Hz		-19	
	f = 300 Hz	-13.1	-9.5	
	f = 500 Hz	-7	-5	
	f = 2 kHz	5	7	
Relative gain	f = 3 kHz	7.7	10.5	dB
	f = 3.4 kHz	6.6	11.6	
	f = 3.94 kHz		-40	
	f = 4.06 kHz		-40	
	f = 10 kHz		-16.5	

M1/M2 to TXO frequency response, AMPS/TACS mode, TXLIM = 2.5 V, input = 20 mVrms, TXTRIM = <1000>, MICTRIM = <1000>, compressor bypassed (see Figures 5 and 6 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 1 kHz	11.5	13.5	dB
	f = 100 Hz		-19	
	f = 300 Hz	-13.1	-9.5	
	f = 500 Hz	-7	-5	dB
	f = 2 kHz	5	7	
Relative gain	f = 2.75 kHz	6.8	9.8	
relative gain	f = 3 kHz	5.5	10.5	
	f = 3.1 kHz		10.8	
	f = 5.9 kHz		-25	
	f = 6.1 kHz		-25	
	f = 10 kHz		-19	

### TXSUM to TXO gain adjust, input = 100 mVrms at 1 kHz (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain, TXTRIM = <1000>		±1	dB
Step size	0.4	0.6	dB
Positive range	3	4	dB
Negative range	-4.5	-3.5	dB

# compressor gain characteristics, M1/M2 to TXO, f = 1 kHz, MICTRIM = <1000> (see Figure 2 for external components)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain (reference)	M1/M2 = 100 mVrms	11	13	dB
	M1/M2 = 100 mV+ 3 dB	1	2	
Relative gain	M1/M2 = 100  mV - 25  dB	-13	-12	dB
	M1/M2 = 100  mV - 50  dB	-25.5	-24.5	

TX limiter, peak-to-peak output at TXO relative to nominal output, M1/M2 = 1 Vrms at 1 kHz, MICTRIM = <1000>, compressor enabled, nominal output measured at TXO for 100 mVrms input at M1/M2

PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
	TXLIM = 500 mV	1.25	
	TXLIM = 550 mV	1.35	
Peak-to-peak output	TXLIM = 600 mV	1.45	V/V
	TXLIM = 650 mV	1.55	
	TXLIM = 800 mV	1.85	



transmit path specifications,  $V_{CC} = 5 \text{ V}$ ,  $f_{xtal} = 3.58 \text{ MHz}$  (continued)

total harmonic distortion at TXO vs TXLIM, M1/M2 = 100 mVrms at 1 kHz, MICTRIM = <1000>, TXTRIM = <1000>, compressor enabled

PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
	TXLIM = 500 mV	10%	
Total harmonic distortion	TXLIM = 550 mV	4%	
	TXLIM = 600 mV	1%	
	TXLIM = 800 mV	1%	

# noise at TXO, M1/M2 = V<sub>BTX</sub>, MICTRIM = <1000>, TXTRIM = <1000>, compressor enabled (see Table 1)

PARAMETER	MIN	MAX	UNIT
Psophometrically-weighted rms noise		4	mV

## transmit path switch attenuation (see Table 1)

PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
Attenuation	MICSW, HSS = 0/1	50	
	TDSW, TTS = 0/1	50	
	TXVOICE, TVM = 1	50	dB
	TXSUM, TSM = 1	50	1
	TXDATA, TDM = 1	50	1

## TDI to TXO gain, TDI = 100 mVrms at 1 kHz, TXTRIM = <1000> (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain	11	13	dB

# transmit data frequency response, NMT mode, TXDA = 275 mVrms, TXTRIM = <1000>(see Figure 7 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 1.5 kHz	2	4	dB
Relative gain	f = 100 Hz		-22	
	f = 700 Hz	-7.6	-5.6	
	f = 2.1 kHz	1.9	2.9	dB
	f = 2.3 kHz		4.7	uБ
	f = 3.5 kHz		4.7	
	f = 100 kHz		-25	

# transmit data frequency response, AMPS/TACS mode, TXDA = 400 mVrms, TXTRIM = <1000>(see Figure 8 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 1 kHz		±1	dB
Relative gain	f = 8 kHz		±1	
	f = 16 kHz	-6	1	dB
	f = 30 kHz		-18	uБ
	f = 100 kHz		-18	

### digital-to-analog converter, measured at DAC (see Table 1 for codes)

PARAMETER	MIN	MAX	UNIT
Output code <0000>		50	mV
Output code <1111>, relative VBRX	-100	0	mV
DNL, typical step size = 160 mV		0.5	LSB



# receive path specifications, V<sub>CC</sub> = 5 V, f<sub>xtal</sub> = 3.58 MHz

## RXIN to RXO, RXTRIM, RXIN = 100 mVrms at 1 kHz (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain, RXTRIM = <1000>	-7	-5	dB
Step size	0.4	0.6	dB
Positive range	3	4	dB
Negative range	-4.5	-3.5	dB

# RXIN to HFO/HSO frequency response, RXIN = 60 mVrms, RXTRIM = <1000>, VC = <1000>, expander bypassed (see Figures 9 and 10 and Table 1)

PARAI	METER	TEST CONDITIONS	MIN	MAX	UNIT
Gain		f = 1 kHz	-7	-5	dB
	f = 100 Hz		-30		
	f = 240 Hz		13.4		
	f = 300 Hz	7.9	11.5		
	f = 500 Hz	5	7		
Relative gain		f = 2 kHz	-7	-5	dB
		f = 3 kHz	-12	-8.5	
	f = 3.4 kHz	-17	-9.6		
	f = 4 kHz		-40		
	f = 10 kHz		-40		

# CTI to HFO/HSO, volume control, CTI = 100 mVrms at 1 kHz (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain, VC = <1000>	0	2	dB
Step size	1.8	2.2	dB
Positive range	13.5	14.5	dB
Negative range	-16.5	-15.5	dB

### ST/TDI to HFO/HSO gain, ST/TDI = 100 mVrms at 1 kHz, VC = <1000> (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain ST to HSO	-7	-5	dB
Gain to TDI to HSO/HFO		±1	dB

### expander gain characteristics, RXIN to HSO/HFO, f = 1 kHz, RXTRIM = <1000>, VC = <1000> (see Figure 2)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain (reference)	RXIN = 200 mVrms		±1	dB
	RXIN = 200 mV + 3 dB	2	4	
Expanded relative gain	e gain $RXIN = 200 \text{ mV} - 12.5 \text{ dB}$	-13.5	-11.5	dB
·	RXIN = 200 mV - 25 dB	-23	-21	

# RX limiter, peak-to-peak output at HSO vs RXLIM, RXIN = 500 mVrms at 1 kHz, RXTRIM = <1000>, VC = <1000>, expander enabled

PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
Peak-to-peak output	RXLIM = 750 mV	1.6	
	RXLIM = 850 mV	1.8	VPP
	RXLIM = 950 mV	2	



# receive path specifications, $V_{CC} = 5 \text{ V}$ , $f_{xtal} = 3.58 \text{ MHz}$ (continued)

noise and distortion at HFO/HSO, RXTRIM = <1000>, VC = <1000>, expander enabled (see Table 1)

PARAMETER	MIN	MAX	UNIT
Psophometrically-weighted rms noise, RXIN = V <sub>BRX</sub>		1	mV
Total harmonic distortion, RXIN = 200 mV rms at 1 kHz		1%	

### receive path switch attenuation (see Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
	CTI, CTM = 1	50		
Attenuation	TDI, RTM = 1	50		dB
	RXVOICE, RVM = 1	50		ub
	LSSW, HSS = 0/1	50		

# receive data frequency response, NMT mode, RXIN = 100 mVrms, RXTRIM = <1000> (see Figure 11 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 1.5 kHz	8	10	dB
Relative gain	f = 100 Hz		-10	dB
	f = 600 Hz		5.4	
	f = 800 Hz	3.4	5.4	
	f = 900 Hz	3.4	5.4	uБ
	f = 2.2 kHz	-4.3	-2.3	
	f = 10 kHz		-15.5	

# receive data frequency response, AMPS/TACS mode, RXIN = 100 mVrms, RXTRIM = <1000> (see Figure 12 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 1 kHz	5	7	dB
Relative gain	f = 8 kHz		±1	
	f = 16 kHz	-6	1	dB
	f = 40 kHz		-18	ub
	f = 100 kHz		-18	

# receive SAT path frequency response, NMT mode, RXIN = 100 mVrms, RXTRIM = <1000>, SATTRIM = <100> (see Figure 13 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 4 kHz	3	5	dB
	f = 100 Hz		-35	
	f = 2 kHz		-35	
	f = 3.2 kHz		-25	
Relative gain	f = 3.4 kHz		-20	
	f = 3.8 kHz	-5	0.5	
	f = 3.94 kHz		±0.5	dB
	f = 4.06 kHz		±0.5	
	f = 4.2 kHz	-5	0.5	
	f = 5 kHz		-20	
	f = 6 kHz		-35	
	f = 10 kHz		-35	



receive path specifications,  $V_{CC} = 5 \text{ V}$ ,  $f_{xtal} = 3.58 \text{ MHz}$  (continued)

receive SAT path frequency response, AMPS/TACS mode, RXIN = 100 mVrms, RXTRIM = <1000>, SATTRIM = <100> (see Figure 15 and Table 1)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Gain	f = 6 kHz	3	5	dB
	f = 100 Hz		-35	
	f = 3 kHz		-35	
	f = 4.8 kHz		-25	
	f = 5.1 kHz		-20	
Relative gain	f = 5.8 kHz	-5	0.5	
	f = 5.94 kHz		±0.5	dB
	f = 6.06 kHz		±0.5	
	f = 6.2 kHz	-5	0.5	
	f = 7.5 kHz		-20	
	f = 9 kHz		-35	
	f = 10 kHz		-35	

receive SAT path trim, SATTRIM, RXIN = 100 mVrms, f at 4 kHz NMT, f at 6 kHz AMPS/TACS, RXTRIM = <1000> (see Table 1)

PARAMETER	MIN	MAX	UNIT
Gain, SATTRIM = <100>	3	5	dB
Step size	0.8	1.2	dB
Positive range	2.5	3.5	dB
Negative range	-4.5	-3.5	dB

PARAMETER MEASUREMENT INFORMATION

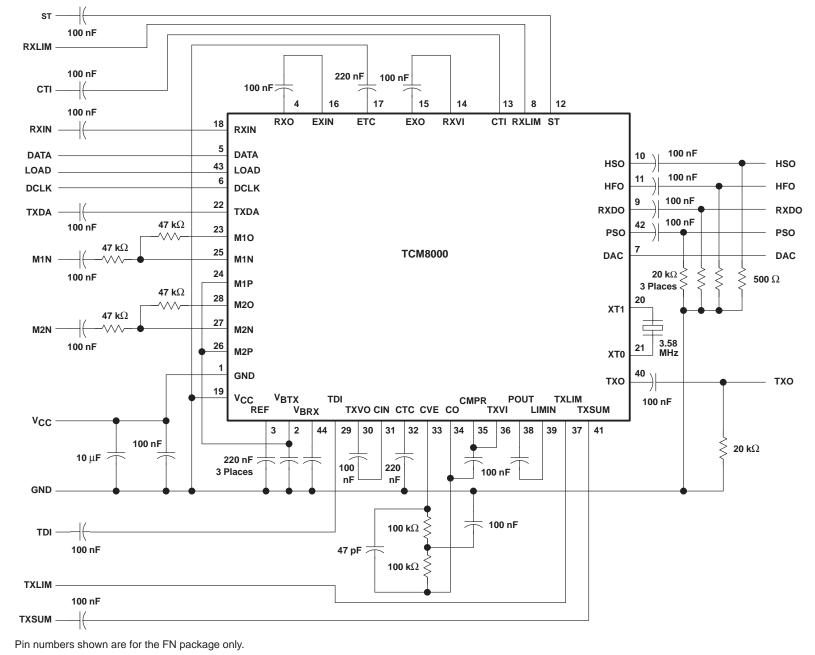


Figure 2. Test Circuit

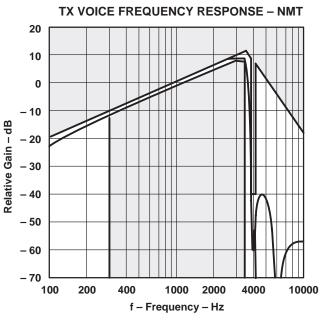


Figure 3

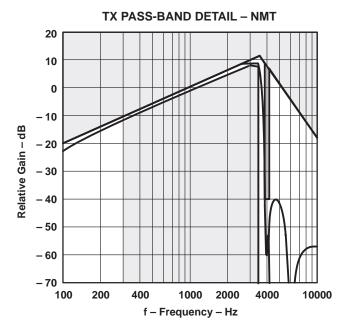
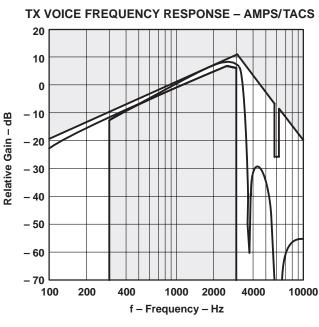


Figure 4





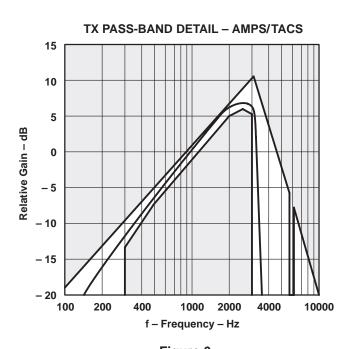


Figure 6

<sup>&</sup>lt;sup>†</sup> The boundaries of the shaded areas represent the frequency response limits met by the TCM8000. The curves in the unshaded areas represent typical filter response of the device.



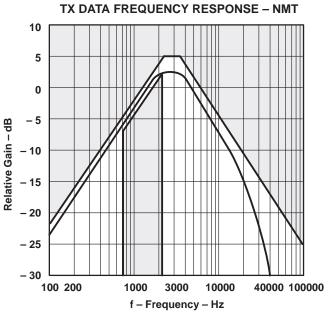


Figure 7

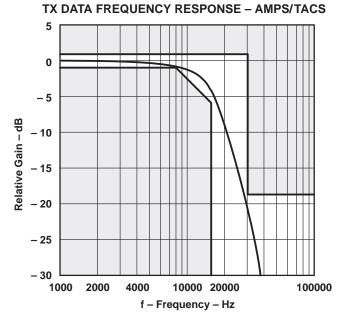


Figure 8

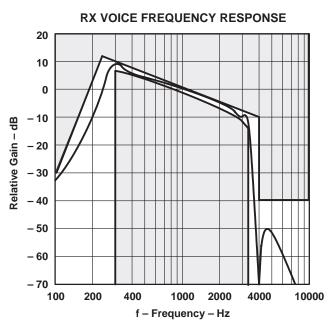
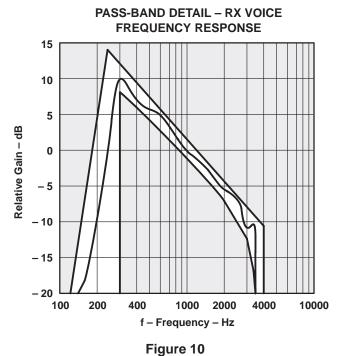


Figure 9



<sup>†</sup> The boundaries of the shaded areas represent the frequency response limits met by the TCM8000. The curves in the unshaded areas represent the filter response of the device.



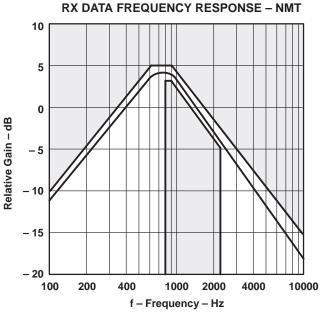


Figure 11

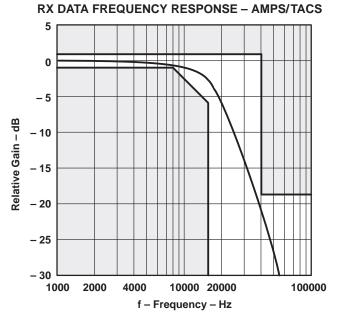


Figure 12

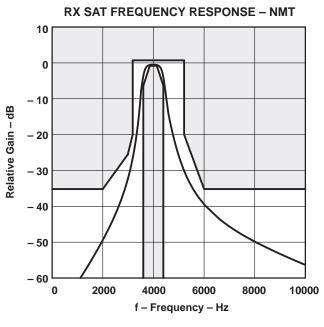
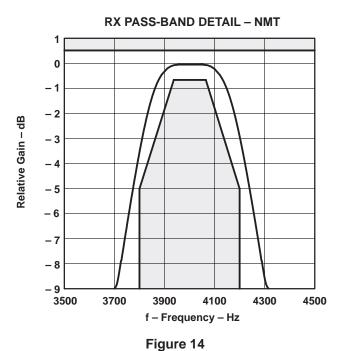
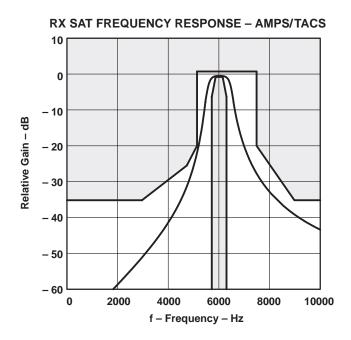


Figure 13



<sup>†</sup> The boundaries of the shaded areas represent the frequency response limits met by the TCM8000. The curves in the unshaded areas represent the filter response of the device.





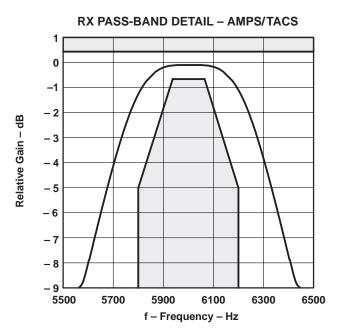


Figure 15 Figure 16

<sup>&</sup>lt;sup>†</sup> The boundaries of the shaded areas represent the frequency response limits met by the TCM8000. The curves in the unshaded areas represent the filter response of the device.

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