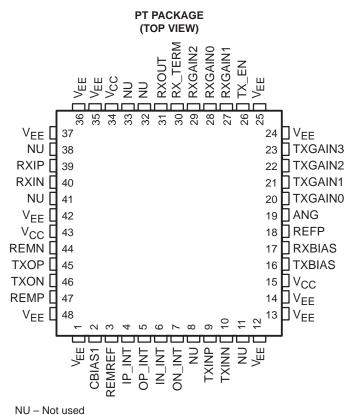
- Single-Chip EtherLoop Transceiver
- Programmable Transmit (TX) and Receive (RX) Gain Via Digital Interface
- Low Overall Power Consumption
- Power-Down Mode Minimizes Server
 Modem Power Consumption in Multiplexed
 Applications
- Low Noise
- Low Distortion

- All Terminals Protected to Survive, Without Damage, a Simulated Static Discharge of 1 kV From a 100-pF Capacitor Applied Through a 1.5-kΩ Resistor With Respect to Chip Ground (V_{EE})
- Single-Rail 5-V Power Supply
- Operating Temperature –40°C to 85°C
 Ambient
 - Allows Operation in Central Office and Distributed-Server Modem Applications
- 48-Pin Thin Plastic Quad Flatpack



description

The TNETEL1400 is an Etherloop transceiver. EtherLoop technology enables simultaneous voice and Ethernet communication over local-loop plain old telephone service (POTS) wiring. The TNETEL1400 supports data rates of up to 6 Mbit/s and POTS wire lengths of up to 21,000 feet. Figure 1 shows a typical system with an EtherLoop modem located at each end of the POTS line. Each EtherLoop modem has a 10Base-T Ethernet interface and is responsible for buffering Ethernet data before sending it over the POTS wire. The server-end (SE) EtherLoop modem is located in a central switching office and can communicate with several client-end (CE) EtherLoop modems, based on a round-robin arbitration scheme. The CE EtherLoop modem typically is located at a remote site.

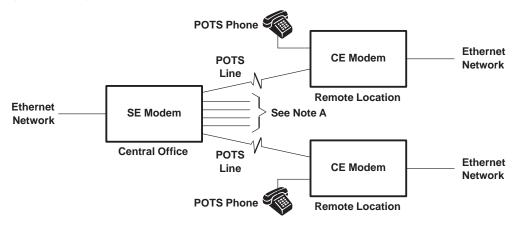


Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

EtherLoop is a trademark of Elastic Networks.



description (continued)



NOTE A: Flexible multiplexing scheme allows one SE modem to interface with many CE modems.

Figure 1. Typical EtherLoop System

Figure 2 shows a block diagram of a typical CE EtherLoop modem. Ethernet data destined for the POTS wire is received via 10Base-T interface and presented to the EtherLoop processor. The EtherLoop processor performs Ethernet frame processing and buffer management. The EtherLoop processor sends buffered Ethernet frames to the TNETEL1200 EtherLoop modem. The TNETEL1200 performs data modulation before passing the modulated digital data to a digital-to-analog (DAC) converter. The resulting analog signal passes to the TNETEL1400 transceiver, which acts as the line interface. The modem uses a half-duplex communication protocol over the POTS wire, and data received from the POTS wire follows the reverse path back to the Ethernet framer.

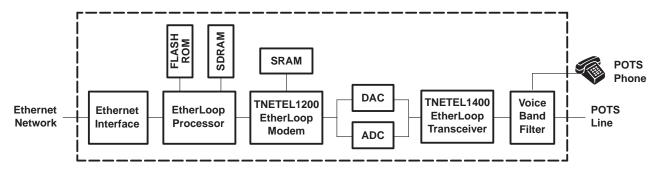


Figure 2. Typical CE EtherLoop Modem

Figure 3 shows a block diagram of a typical SE EtherLoop modem. Data flow follows the same path as in the CE EtherLoop modem. In the SE application, the EtherLoop processor also performs round-robin arbitration between each of the attached TNETEL1400 devices.

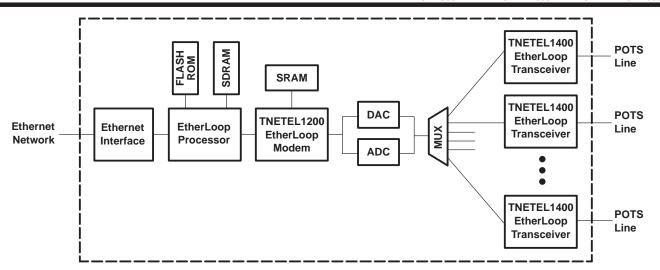
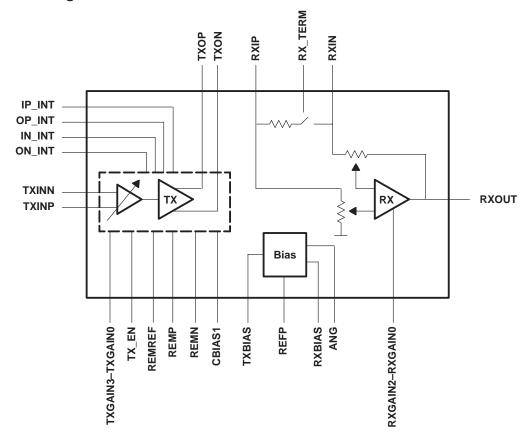


Figure 3. Typical SE EtherLoop Modem

summary of TNETEL1400 EtherLoop transceiver

- Drives POTS line with signal generated by DAC
- Interfaces signal received from POTS line to ADC

functional block diagram





APPLICATION INFORMATION[†]

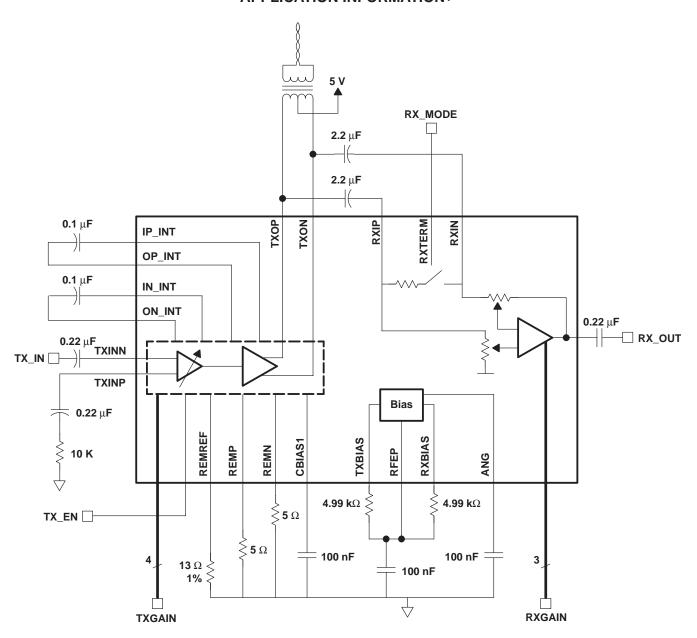


Figure 4. EtherLoop Front-End Application (CE)



[†] All bias resistors should be 1% tolerance. The resistors on REMP, REMN, and REMREF also should be 1% and placed as close as possible to their respective pins.

Terminal Functions

transmit (TX)

TERM	TERMINAL		DESCRIPTION		
NAME	NO.	1/0†	DESCRIPTION		
CBIAS1	2	I/O	Transmit voltage bias decoupling		
IN_INT	6	I	Transmit interstage ac coupling pin 1 (negative side)		
IP_INT	4	I	Transmit interstage ac coupling pin 2 (positive side)		
ON_INT	7	0	Transmit interstage ac coupling pin 2 (negative side)		
OP_INT	5	0	Transmit interstage ac coupling pin 1 (positive side)		
REMREF	3	0	Transmit temperature-compensating bias reference		
			Transmit enable		
TX_EN	26	- 1	1 = Transmitter enabled		
			0 = Transmitter disabled		
TXBIAS	16		Transmit current bias		
TXGAIN3 TXGAIN2 TXGAIN1 TXGAIN0	23 (MSB) 22 21 20 (LSB)	I	Transmit preattenuation select (0 to –30 dB in –3-dB steps) 0000 = 0 dB 0001 = -3 dB 1010 = -30 dB 1011 = TX OFF 1110 = TX OFF 1111 = TX OFF		
TXINN	10	I	Transmit input negative. TXINN can be coupled to ground for SE input).		
TXINP	9	I	Transmit input positive. TXINP can be coupled to ground for SE input).		
TXON	46	0	Transmitter output negative		
TXOP	45	0	Transmitter output positive		

[†] I = input, O = output

Terminal Functions (Continued)

receive (RX)

TERM NAME	INAL NO.	1/0†	DESCRIPTION
RX_TERM	30	I	Receive passive termination $ RX_TERM = 1: 110 \ \Omega \ switched \ IN \\ RX_TERM = 0: 110 \ \Omega \ switched \ OUT $
RXBIAS	17	- 1	Receive current bias
RXGAIN2 RXGAIN1 RXGAIN0	29 (MSB) 27 28 (LSB)	I	Receive gain select 000 = RX OFF • • 011 = RX OFF 100 = 0 dB 101 = 12 dB 110 = 24 dB 111 = 30 dB
RXIN	40	Ι	Receiver input negative/TX feedback
RXIP	39	Ī	Receiver input positive/TX feedback
RXOUT	31	0	Receiver output (single ended)

 $[\]dagger I = input, O = output$

miscellaneous

TERMII	NAL	t	DESCRIPTION
NAME	NO.	1/0†	DESCRIPTION
ANG	19	0	Analog ground (2.5 V) reference
REFP	18	0	4-V bandgap reference
REMN	44	I/O	Negative external emitter resistor
REMP	47	I/O	Positive external emitter resistor

[†] I = input, O = output

power supply

TERMINAL		DESCRIPTION		
NAME	NO.	DESCRIPTION		
NU	8, 11, 32, 33, 38, 41	Not used		
Vcc	15, 34, 43	5-V power		
VEE	1, 12, 13, 14, 24, 25, 35, 36, 37, 42, 48	Ground		



absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply-voltage range, V _{CC}	4.3 V to V _{CC} to 0.7 V
Input-voltage range: Analog inputs	. -0.7 V to V_{CC} to 0.7 V
Output-voltage range, V _O	
Storage temperature range, T _{stq}	55°C to 25°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	MAX	UNIT
Vcc	Supply voltage	4.75	5.25	V
VIH	High-level input voltage	2.1		V
V _{IL}	Low-level input voltage		1	V
IOH	High-level input current			mA
loL	Low-level input current			mA
TA	Operating free-air temperature range	-40	85	°C

electrical characteristics over recommended operating conditions

PARAMETER		TEST CONDITIONS		TYP [†]	MAX	UNIT
ICC Supply current		Normal operation			50	μA
Icc	Supply current	Power-down mode	1.5		3	μΑ
VREF	4-V reference voltage		3.88	4	4.12	V
VANG	2.5-V reference voltage		2.38	2.5	2.62	V
IREFP	4-V reference current	Source			1	mA
IANG	2.5-V reference current	Source/sink			100	mA

 $[\]overline{\dagger}$ All typical values are at V_{CC} = 5 V, T_A = 25°C (unless otherwise noted).

transmitter (see Figure 5)

PAR	RAMETER	TEST CONDITIONS		MIN	TYP [†]	MAX	UNIT
TX maximum	output level	V_{CC} = 5 V, R_L = 110 Ω, RX_TERM = 0, TX_EN = 1, V_{TXINP} = 1.2 Vpp sinusoid at f = 500 kHz (see Note 1)	TXGAIN = 0000, RXGAIN = 0XX	21	22	23	dBm
			TXGAIN = 0001	-3.3	-3	-2.7	
			TXGAIN = 0010	-6.3	-6	- 5.7	
			TXGAIN = 0011	-9.3	-9	-8.7	
TX attentuato	r accuracy	$V_{CC} = 5 \text{ V}, R_{L} = 110 \Omega,$	TXGAIN = 0100	-12.3	-12	-11.7	
	aximum output)	RX_TERM = 0, TX_EN = 1, RXGAIN = 0XX, V(TXINP) = 1.2 Vpp sinusoid at f = 500 kHz	TXGAIN = 0101	-15.3	-15	-14.7	dB
TX attentuato	,	[output is measured at IN_INT and IP_INT	TXGAIN = 0110	-18.3	-18	-17.7	uБ
(relative to maximum output)		(ac coupled)]	TXGAIN = 0111	-21.3	-21	-20.7	
			TXGAIN = 1000	-24.3	-24	-23.7	
			TXGAIN = 1001	-27.3	-27	-26.7	
			TXGAIN = 1010	-30.3	-30	-29.7	
TX output level variation over frequency		30 kHz < f < 2.5 MHz monotonically decreasing for f > 3 MHz, $R_L = 110~\Omega$, RX_TERM = 0, TX_EN = 1, V(TXINP) = 1.2~Vpp sinusoid at f = 500 kHz with transformer connected as in Figure 1 (see Note 1)	TXGAIN = 0000, RXGAIN = 0XX			1	dB
TX output distortion	Odd harmonics	$V_{CC} = 5 \text{ V}, R_{L} = 110 \Omega,$ $RX_{TERM} = 0, TX_{EN} = 1,$	TXGAIN = 0000,	-35			4D.c
(all gain settings)	Even harmonics	V _(TXINP) = 1.2 V _{PP} sinusoid at f = 500 kHz (see Note 1)	RXGAIN = 0XX	-50			dBc
TX output signal-to-nois (all gain settir		V_{CC} = 5 V, R_L = 110 Ω, RX_TERM = 0, TX_EN = 1, $V_{(TXINP)}$ = 1.2 Vpp sinusoid at f = 500 kHz (see Note 1)	TXGAIN = 0000, RXGAIN = 0XX	50			dB
TX maximum output-level variation with V _{CC}		V_{CC} = 5 V ± 0.25 V, R _L = 110 Ω, RX_TERM = 0, TX_EN = 1, V(TXINP) = 1.2 Vpp sinusoid at f = 500 kHz (see Note 1)	TXGAIN = 0000, RXGAIN = 0XX			1	dB/V
Z _{in(TXIN)}	TX input impedance	TXGAIN = 0000, RXGAIN = 0XX (see Note 1)			1400	TYP+ 30%	Ω
TX input impe	edance variation Zin(TXIN)	TXGAIN = 0000, RXGAIN = 0XX		-30%		30%	

[†] All typical values are at V_{CC} = 5 V, T_A = 25°C (unless otherwise noted).

NOTE 1: While the RX circuit is disabled during transmission, it is still connected and, therefore, must withstand the signal levels placed at its input terminals.



transmitter (see Figure 5) (continued)

PARAMETER	TEST CONDITIONS			TYP [†]	MAX	UNIT
I _{rms} (TXOP)- TX output I _{rms} (TXON) current balance	V_{CC} = 5 V, R_L = 110 Ω , RX_TERM = 0, TX_EN = 1, $V_{(TXINP)}$ = 1.2 V_{PP} sinusoid at f = 500 kHz (see Note 1)	TXGAIN = 0000, RXGAIN = 0XX	– 5		5	mA
TX output stability	Source impedance \leq 50 Ω , Supply impedance \leq 10 Ω , Z $_{loads}$: voltage standing-wave ratio (VSWR) 4:1 and open circuit	TXGAIN = 0000, RXGAIN = 0XX				
TX supply current	$V_{OUT} = 0$,	TXGAIN = 0000		35	45	mA
1 × supply current	$V_{OUT} = MAX$,	TXGAIN = 0000			120	IIIA
TX output return loss	TXGAIN = 0000, RXGAIN = 0XX		18			dB
TX power-up time	TXGAIN = 0000, RXGAIN = 0XX (see Note 2)				100	μs

[†] All typical values are at V_{CC} = 5 V, T_A = 25°C (unless otherwise noted).

- NOTES: 1. While the RX circuit is disabled during transmission, it is still connected and, therefore, must withstand the signal levels placed at its input terminals.
 - 2. The power-up/power-down time is the time it takes for the signal path to completely settle and meet all the transmission specifications after TXGAIN and RXGAIN are set to power-up condition or switched from one gain setting to another. This time consists of slewing and exponential settling of bias and AC coupling capacitors and, therefore, the values of these components must be as shown in the application diagram, Figure 4.

receiver (see Figures 6 and 7)

PARAMETER	TEST CONDITIONS		MIN	TYP†	MAX	UNIT
	$R_L = 2 k\Omega$, $C_L = 20 pF$, 30 kHz < f < 2.75 MHz,	RXGAIN = 11			691	
RX idle channel noise	RX_TERM = 1, TX_EN = 0,	RXGAIN = 10			478	μV
	V(RXIP - RXIN) = 0.04 Vpp sinusoid at f = 500 kHz	RXGAIN = 01			266	RMS
	(see Note 3)	RXGAIN = 00			160	
		RXGAIN = 11	24	30	31	
DV gain accuracy	$\label{eq:CC} \begin{array}{l} \text{V}_{CC} = 5 \text{ V, R}_{L} = 2 \text{k}\Omega, \text{C}_{L} = 20 \text{ pF,} \\ 30 \text{ kHz} < \text{f} < 2.75 \text{ MHz, RX_TERM} = 1, \text{ TX_EN} = 0, \\ \text{V}(\text{RXIP} - \text{RXIN}) = 0.04 \text{ Vpp sinusoid at f} = 500 \text{ kHz} \\ \end{array}$	RXGAIN = 10	23	24	25	dB
RX gain accuracy		RXGAIN = 01	11	12	13	
		RXGAIN = 00	-1	0	1	
RX gain over frequency (WRT gain at 500 kHz)	$R_L = 2 \text{ k}\Omega, C_L = 20 \text{ pF, } 30 \text{ kHz} < f < 2.75 \text{ MHz,} \\ 30 \text{ kHz} < f < 2.5 \text{ MHz monotically decreasing} \\ \text{for } f > 3 \text{ MHz, } \text{TX_EN} = 0, \\ \text{V(RXIP} - \text{RXIN)} = 0.04 \text{ Vpp sinusoid at } f = 500 \text{ kHz} \\ \end{cases}$	RX_TERM = 1, RXGAIN = 111	-1		1	dB
RX power-supply rejection (WRT V _{CC} only)	$R_L = 2 \text{ k}\Omega$, $C_L = 20 \text{ pF}$, dc < f < 3 MHz, $TX_EN = 0$, $V(RXIP - RXIN) = 0.04 \text{ Vpp}$ sinusoid at f = 500 kHz	RX_TERM = 1, RXGAIN = 111			0.03	V/V
RX common-mode rejection	$R_L = 2 \text{ k}\Omega$, $C_L = 20 \text{ pF}$, $TX_EN = 0$, $V(RXIN) = 1.5 \text{ Vpp}$, $V(RXIP - RXIN) = 0.04 \text{ Vpp}$ sinusoid at $f = 500 \text{ kHz}$	RX_TERM = 1, RXGAIN = 111	30			dB
RX IIP3 intercept	$R_L = 2 \text{ k}\Omega$, $C_L = 20 \text{ pF}$, TX_EN = 0, V(RXIP - RXIN) = 0.04 Vpp sinusoid at f = 500 kHz (see Note 4)	RX_TERM = 1, RXGAIN = 111	17			dBm

[†] All typical values are at $V_{CC} = 5 \text{ V}$, $T_A = 25^{\circ}\text{C}$ (unless otherwise noted).

- NOTES: 3. Idle channel noise is the noise (V_{rms}) measured at RXOUT with no signal at RXIN. This voltage is integrated over the 30-KHz to 2.75-MHz band. This specification is in place of the original noise-figure specification, and is correlated to NF with laboratory measurements.
 - 4. The two tones used for this test are at 1.39 MHz and 1.58 MHz, and the in-band IIP3 products are at 1.2 MHz and 1.77 MHz. The IIP3 intercept point is the output power level, where the power of the harmonics equals that of the signal frequencies. This point is an intersection of two straight lines extrapolated from two low-power measurements.



TNETEL1400 **EtherLoop™ TRANSCEIVER**

SPHS004A - FEBRUARY 1999 - REVISED MARCH 1999

receiver (see Figures 6 and 7) (continued)

PARAMETER	TEST CONDITIONS		MIN	TYP [†]	MAX	UNIT
RX output total harmonic distortion	$R_L = 2 \text{ k}\Omega$, $C_L = 20 \text{ pF}$, $TX_EN = 0$, $V(RXOUT) = 2 \text{ Vpp}$, $V(RXIP - RXIN) = 0.04 \text{ Vpp}$ sinusoid at $f = 500 \text{ kHz}$	RX_TERM = 1, RXGAIN = 111			-40	dBc
RX Z _{IN}	$R_L = 2 kΩ$, $C_L = 20 pF$, RXGAIN = 111, TX_EN = 0,	RX_TERM = 1	77	110	143	Ω
KA ZIN	$V(RXIP - RXIN) = 0.04 V_{PP}$ sinusoid at f = 500 kHz	RX_TERM = 0	10			kΩ
RX maximum supply current	$R_L = 2 \text{ k}\Omega$, $C_L = 20 \text{ pF}$, $TX_EN = 0$, $V(RXOUT) = 4 \text{ Vpp}$, $V(RXIP - RXIN) = 0.04 \text{ Vpp}$ sinusoid at $f = 500 \text{ kHz}$	RX_TERM = 1, RXGAIN = 111			20	mA
RX power-up time	R_L = 2 k Ω , C_L = 20 pF, TX_EN = 0, V(RXIP - RXIN) = 0.04 Vpp sinusoid at f = 500 kHz (see Note 2)	RX_TERM = 1, RXGAIN = 111			30	μs
Power-down supply current	$R_L=2~k\Omega,~C_L=20~pF,~RX_TERM=1,~TX_EN=0,\\ V(RXIP-RXIN)=0.04~Vpp~sinusoid~at~f=500~kHz~(see~Note~2)$	TXGAIN = 1111, RXGAIN = 000			3	mA

[†] All typical values are at V_{CC} = 5 V, T_A = 25°C (unless otherwise noted).

NOTE 2. The power-up/power-down time is the time it takes for the signal path to completely settle and meet all the transmission specifications after TXGAIN and RXGAIN are set to power-up condition or switched from one gain setting to another. This time consists of slewing and exponential settling of bias and AC coupling capacitors and, therefore, the values of these components must be as shown in the application diagram, Figure 4.



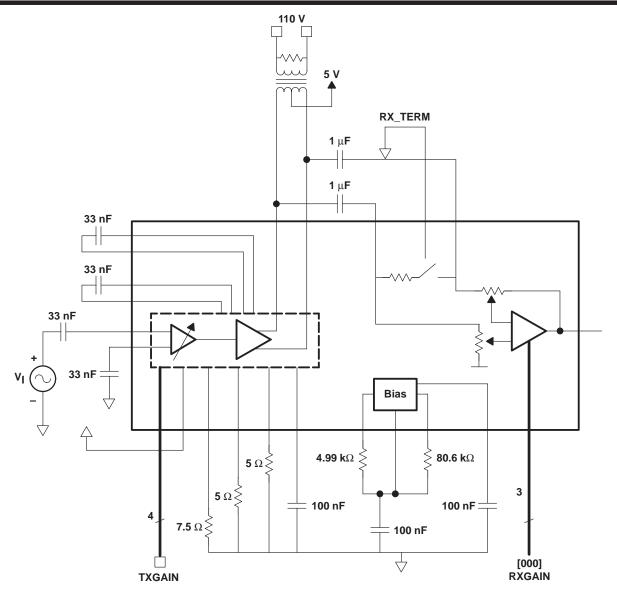


Figure 5. TX Test Circuit

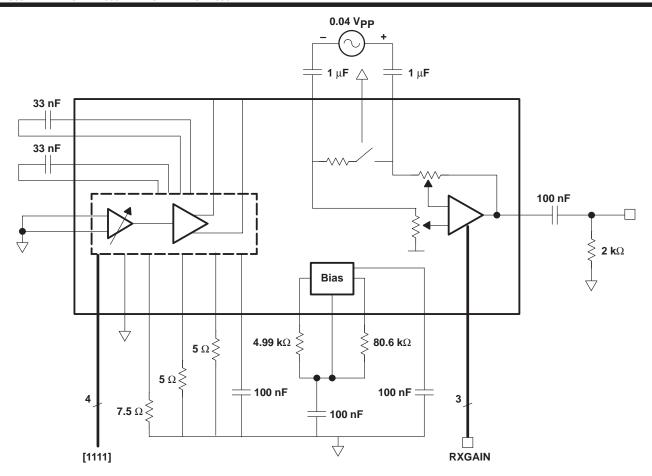


Figure 6. RX Test Circuit

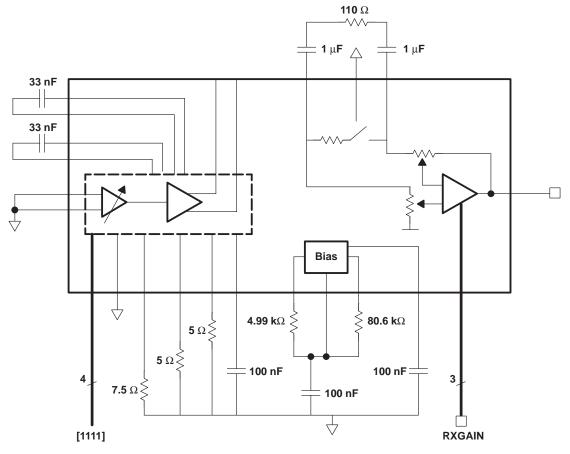
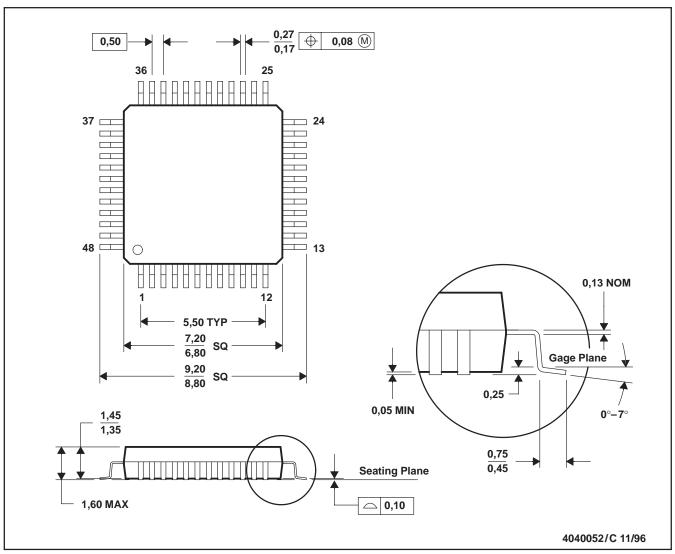


Figure 7. RX Noise Test Circuit

MECHANICAL DATA

PT (S-PQFP-G48)

PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-026
- D. This may also be a thermally enhanced plastic package with leads conected to the die pads.

IMPORTANT NOTICE

Texas Instruments and its subsidiaries (TI) reserve the right to make changes to their products or to discontinue any product or service without notice, and advise customers to obtain the latest version of relevant information to verify, before placing orders, that information being relied on is current and complete. All products are sold subject to the terms and conditions of sale supplied at the time of order acknowledgement, including those pertaining to warranty, patent infringement, and limitation of liability.

TI warrants performance of its semiconductor products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

CERTAIN APPLICATIONS USING SEMICONDUCTOR PRODUCTS MAY INVOLVE POTENTIAL RISKS OF DEATH, PERSONAL INJURY, OR SEVERE PROPERTY OR ENVIRONMENTAL DAMAGE ("CRITICAL APPLICATIONS"). TI SEMICONDUCTOR PRODUCTS ARE NOT DESIGNED, AUTHORIZED, OR WARRANTED TO BE SUITABLE FOR USE IN LIFE-SUPPORT DEVICES OR SYSTEMS OR OTHER CRITICAL APPLICATIONS. INCLUSION OF TI PRODUCTS IN SUCH APPLICATIONS IS UNDERSTOOD TO BE FULLY AT THE CUSTOMER'S RISK.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards must be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance or customer product design. TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used. TI's publication of information regarding any third party's products or services does not constitute TI's approval, warranty or endorsement thereof.

Copyright © 1999, Texas Instruments Incorporated