

TMC2243 CMOS FIR Filter 10 x 10 Bit. 20 MHz

Features

- 20 MHz data input and computation rate
- 10 x 10 bit multiplication with 23-bit extended precision sum of products (overflow, plus 16 output and 6 guard bits)
- Up to 3 zero and 3 non-zero stages per device
- Two's complement arithmetic
- 16-bit Sum-In and Sum-Out ports for cascading
- Internal 1/2 LSB rounding
- · All inputs and outputs are registered
- One coefficient update per clock cycle
- Low power consumption CMOS process
- Single +5V supply

 Available in 68-pin ceramic pin grid array and 69-pin plastic PGA packages

Applications

- · FIR filters
- Adaptive filters
- Multi-bit correlation
- · One and two dimension video filtering
- Radar signal processors
- · One and two dimension convolution
- Arithmetic element for systolic array processors

Description

The TMC2243 is a video speed three stage 10 x 10 bit FIR (Finite Impulse Response) filter integrated circuit composed of three registered multiplier-adders concatenated into a one-dimensional systolic array. Utilizing two's complement representation, the TMC2243 accepts one 10-bit data point, updates one 10-bit coefficient, and produces one 16-bit rounded, filtered output point every 50 nanoseconds.

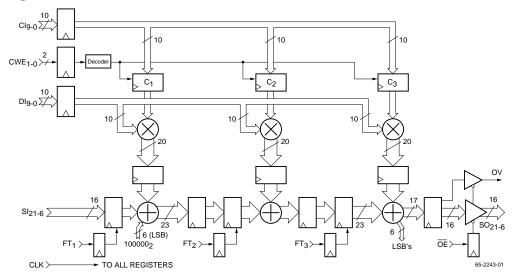
The TMC2243 has features which facilitate longer FIR filters: a 16-bit SUM-In port and user programmable pipeline registers. Enabling these registers allows the insertion of a zero-coefficient stage before each regular filter stage for up to six stages per TMC2243. Larger FIR filters can be built by

cascading Sum-In and Sum-Out.

Coefficients are stored in 3 registers and are addressed via the 2-bit Write Enable control, allowing one coefficient to be changed per clock cycle. All Data, Sum-In, Sum-Out and instruction inputs are registered on the rising edge of clock.

The 16 bits below the MSB of the 23-bit internal summation path are available at the SUM-In and Sum-Out ports. Six bits of cumulative word growth are provided internally. Data Overflow is indicated by an output flag.

Block Diagram



Functional Description

General Information

The TMC2243 consists of three identical arithmetic cells, each of which contains a 10 x 10 two's complement multiplier and a 23-bit adder. Each cell receives the current data (DI) from the Data input register, multiplies it by a locally stored Coefficient (CI_i), and adds it to the Sum $SI_{(i-1)}$ received from the previous cell. The result,

$$SI_i = DI \times CI_i + SI_{(i-1)},$$

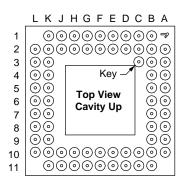
then goes to the next cell via two serial pipeline registers. When only one pipeline register is enabled, stages (i-l) and i are sequential. When both registers are enabled, there is a stage with a zero coefficient between them.

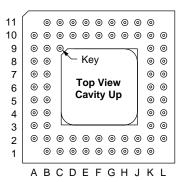
The input arithmetic cell receives $SI_{(i-1)}$ via the 16-bit Sum-In port (registered when $FT_1 = LOW$), filling the six lower bits with 100 000 (1/2 LSB) for internal rounding. The output cell outputs 16 of the MSBs (V₂₁ through V₆) of SO_i through a register to the Sum-Out port. The Overflow flag is set when the final output exceeds 16 bits and resets with the output of the next nonoverflowing result. Sum-Out and the Overflow Flag can be forced to high-impedance with the Output Enable control. See Figure 1.

The two-bit Write Enable control specifies the loading of the three coefficient registers (one per arithmetic cell) with data appearing at the Coefficient Input port.

Pin Assignments

68 Pin Grid Array – G8 Package 69 Pin Plastic Pin Grid Array – H8 Package





Pin	Name	Pin	Name	Pin	Name	Pin	Name
A2	S06	В9	SO20	F10	SI11	K4	CI2
А3	S07	B10	OV	F11	SI ₁₀	K5	CI4
A4	S09	B11	GND	G1	DI6	K6	VDD
A 5	S011	C1	FT ₀	G2	DI7	K7	CI7
A6	S013	C2	OE	G10	SI ₁₃	K8	Cl9
A7	S015	C10	VDD	G11	SI12	К9	SI ₂₀
8A	S017	C11	VDD	H1	DI4	K10	GND
A9	S019	D1	FT2	H2	DI5	K11	SI18
A10	S0 ₂₁	D2	FT1	H10	SI ₁₅	L2	DI ₀
B1	VDD	D10	SI7	H11	SI ₁₄	L3	CI ₁
B2	GND	D11	SI ₆	J1	DI ₂	L4	CI3
В3	S08	E1	CWE ₁	J2	DI3	L5	CI5
B4	S010	E2	CWE0	J10	SI17	L6	CI6
B5	S012	E10	SI9	J11	SI16	L7	CI8
B6	S014	E11	SI8	K1	GND	L8	SI21
B7	S016	F1	DI8	K2	DI1	L9	SI19
B8	S018	F2	DI9	К3	CI0	L10	CLK

Note: Pin C3 is a mechanical orientation pin on the H8 package at manufacturer's option.

65-2243-02

Pin Name	Pin Number	Description
Power	1	
V _{DD}	B1, K6, C10, C11	Supply Voltage. The TMC2243 operates from a single +5V supply.
GND	B2, K1, K10, B11	Ground.
Inputs		
DI9-0	F2, F1, G2, G1, H2, H1, J2, J1, K2, L2	Data Input. Dlg through Dl ₀ is the 10-bit registered Data Input; Dlg is the MSB (sign bit) and Dl ₀ is the LSB. Data is in two's complement representation, and is clocked into the data register on each rising edge of clock. See Figure 1.
Sl ₂₁₋₆	L8, K9, L9, K11, J10, J11, H10, H11, G10, G11, F10, F11, E10, E11, D10, D11	Sum Input. SI ₂₁ through SI ₆ is the 16-bit Sum-In port. SI ₂₁ is the MSB (sign bit). Sum-In is truncated to bit SI ₆ (plus the 1/2 LSB rounding bit in SI ₅) and is in two's complement representation. See Figure 1. The Sum-In port is registered, on the rising edge of clock, only when FT ₁ =LOW. Unique input setup requirements must be observed when operating in the feedthrough mode (FT ₁ = HIGH) See text.
Cl9-0	K8, L7, K7, L6, L5, K5, L4, K4, L3, K3	Coefficient Input. Cl9 through Cl ₀ is the 10-bit registered Coefficient Input; Clg is the MSB (sign bit) and Cl ₀ is the LSB. Each coefficient and its write enable address (CWE ₁₋₃) are registered on the same clock. The coefficient is then latched into the indicated register (C ₁₋₃) at the rising edge of the next clock. The contents of this bus are ignored if a coefficient register is not selected (CWE = 00). The format of Cl ₉₋₀ is identical to that of Dl ₉₋₀ .
Outputs		
SO ₂₁₋₆	A10, B9, A9, B8, A8, B7, A7, B6, A6, B5, A5, B4, A4, B3, A3, A2	Sum Output. SO ₂₁ through SO ₆ is the three-state 16-bit registered Sum-Out port; SO ₂₁ is the MSB (sign bit). For maximum precision, the internal products and accumulations are 23 bits but Sum-Out is internally truncated to 16 bits, and excludes the overflow bit and the 6 LSBs. The format is identical to that of SI ₂₁₋₆ . See Figure 1.
Clock		
CLK	L10	Master Clock. The TMC2243 has a single clock input. The rising edge of CLK strobes all enabled registers. All timing specifications are referenced to the rising edge of clock.
Controls		
CWE ₁₋₀	E1, E2	Coefficient Write Enable. The two bits of the registered Coefficient Write Enable control indicate which of the coefficient registers is to receive a new coefficient at the beginning of the next clock cycle.
		CWE ₁₋₀ Coefficient Register Selected
		0 0 Holds all coefficients unchanged. 0 1 C ₁ 1 0 C ₂ 1 1 C ₃
FT ₃₋₁	D1, D2, C1	Feedthrough. These registered Feed Through controls select clocked ($FT_i = LOW$) or feedthrough ($FT_i = HIGH$) operation for each of the pipeline registers. Setting $FT_i = LOW$ inserts a zero coefficient stage, or additional register, before the ith non-zero stage.
ŌĒ	C2	Output Enable. Output Enable is a registered three-state enable control which forces the Sum-Out port and Overflow to the high-impedance state when HIGH. These outputs are enabled when $\overline{\text{OE}}$ is LOW.
Flags	,	
OV	B10	Overflow. The Overflow Flag is a registered three-state output which goes HIGH whenever the summation result exceeds 16 bits and is reset to LOW on the next nonoverflowing result.

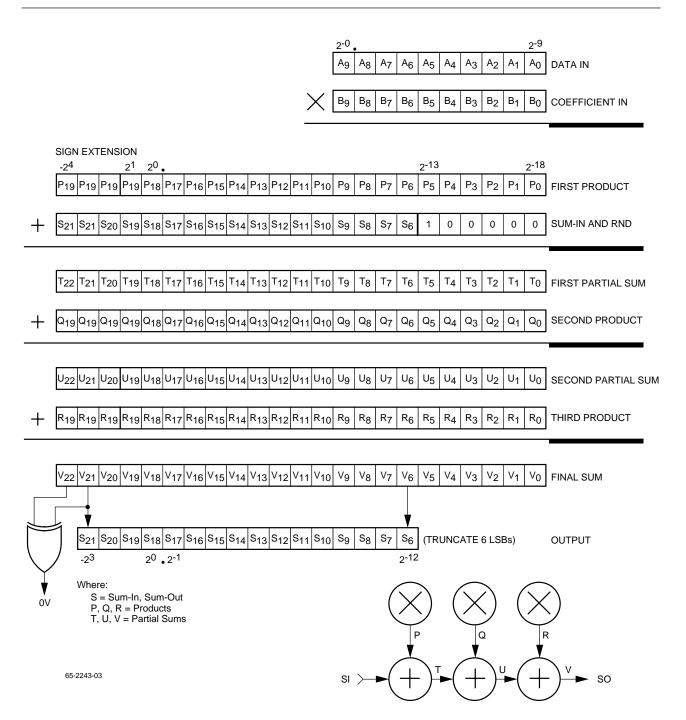
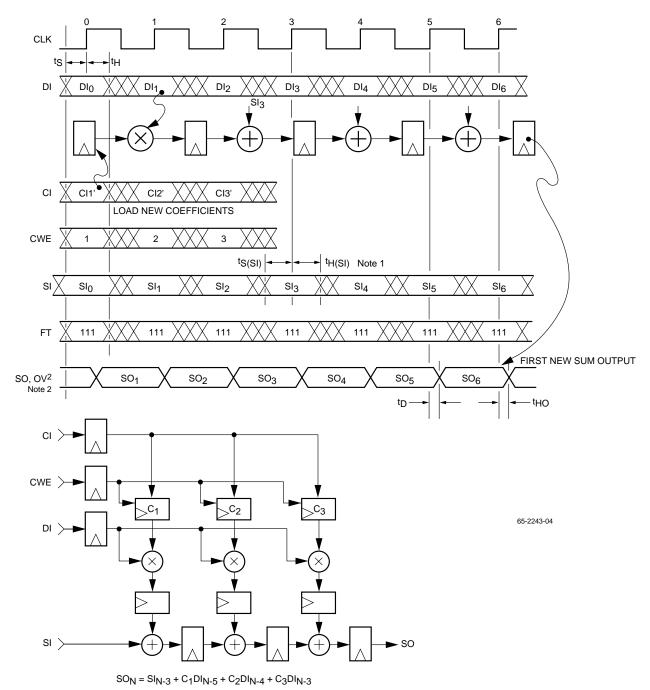


Figure 1. Data Formats and Internal Busing

Because the Sum-In and Sum-Out are truncated by 6 bits relative to the external accumulation pipeline, the TMC2243 rounds internally by adding 2^{-13} to each emerging sum of

products, effecting half-LSB rounding relative to the output format. The chip internally utilizes all lower-order bits, to 2^{-18} .

Basic Operation



- 1. Setup and Hold requirements for the Sum Input are similar to the other registered inputs when $FT_1 = LOW$. See text.
- 2. Sum Out and Overflow timing are shown with $\overline{OE} = LOW$.

Figure 2. Timing Diagram Demonstrating Basic Operation with FT₁₋₃ = HIGH (no zero stages)

The basic equation describing the function of the TMC2243 operating in a fixed state is:

$$\begin{split} SO(N) &= SI(N-6+FT1+FT2+FT3) \\ &+ C_1 \times DI(N-7+FT2+FT3) \\ &+ C_2 \times DI(N-5+FT3) \\ &+ C_3 \times DI(N-3) \end{split}$$

Careful observation of the clock delays shown is basic to construction of a filter algorithm. The operating sequence for the common application with FT1-3 = HIGH (no zero stages) is shown in Figure 2. The simplified block diagram demonstrates the clock stages in this configuration. When FT1 = HIGH, the input feedthrough register is bypassed, and care must be taken to observe the setup requirements on the input of the first adder. Due to the absence of the input register buffer, note that the adder operates on data stable just prior to the arrival of the next clock, and not that setup at the rising edge of the current clock. When FT1 = LOW the input register latches the input data, and the Sum Input follows

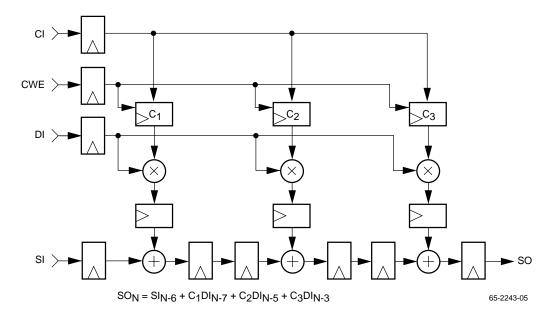
setup and hold requirements similar to the other registered inputs of the TMC2243. When $FT_1 = HIGH$, $t_S(SI)$ is guaranteed to allow 20MHz pipelined operation, assuming that input setup is observed, including cascaded operation. See the AC Characteristics table and Figure 9 in the Applications Discussion section.

Figure 3 shows the effects of the feedthrough registers on filter operation, with two different configurations. The inputs are those presented at the corresponding rising edge of clock, excepting the delayed setup requirements of the Sum Input when $FT_1 = HIGH$. The outputs are those available up to and including the corresponding edge of clock. Applications utilizing the TMC2243's ability to modify coefficients dynamically are demonstrated in Figure 4, showing the operation of a typical adaptive filter. Note that the Sum Output will be zero in the first few clock cycles of all examples only if the Coefficient Registers are initialized to zero beforehand.

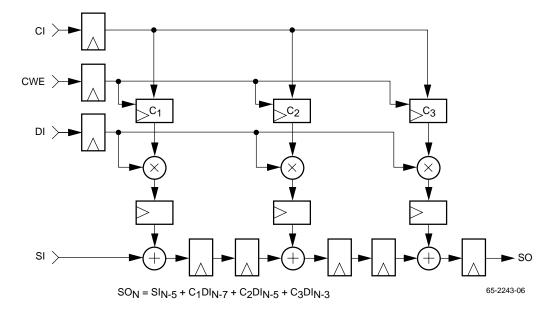
Cycle	SI(A) FT ₁ = LOW	SI(B) FT ₁ = HIGH	DI	CI	CWE	SO
1	0	0	0	K0	01	X
2	0	0	0	K1	10	X
3	0	0	0	K2	11	X
4	SI ₀	0	DI ₀	0	00	X
5	SI ₁	SI ₀	DI ₁	0	00	0
6	SI ₂	SI ₁	D1 ₂	0	00	0
7	SI ₃	SI ₂	DI ₃	0	00	DI_0K_2
8	SI ₄	SI ₃	DI ₄	K ₀ '	01	DI ₁ K ₂
9	SI ₅	SI ₄	DI ₅	0	00	$DI_0K_1 + DI_2K_2$
10	SI ₆	SI ₅	DI ₆	K ₁ '	10	$SI_0 + DI_1K_1 + DI_3K_2$
11	SI7	SI ₆	DI ₇	0	00	$SI_1 + DI_0K_0 + DI_2K_1 + DI_4K_2$
12	SI8	SI ₇	DI ₈	K ₂ '	11	$SI_2 + DI_1K_0 + DI_3K_1 + DI_5K_2$
13	SI9	SI ₈	DI ₉	0	00	$SI_3 + DI_2K_0 + DI_4K_1 + DI_6K_2$
14	0	SI ₉	0	0	00	$SI_4 + DI_3K_0 + DI_5K_1 + DI_7K_2$
15	0	0	0	0	00	$SI_5 + DI_4K_0 + DI_6K_1 + DI_8K_2$
16	0	0	0	0	00	$SI_6 + DI_5K_0' + DI_7K_1' + DI_9K_2'$
17	0	0	0	0	00	SI ₇ + DI ₆ K ₀ ' + DI ₈ K ₁ '
18	0	0	0	0	00	SI ₈ + DI ₇ K ₀ ' + DI ₉ K ₁ '
19	0	0	0	0	00	SI ₉ + DI ₈ K ₀ ′
20	0	0	0	0	00	Dl ₉ K ₀ ′
21	0	0	0	0	00	0

Figure 3. Impulse Response Filter Operation Sequence with FT2,3 = LOW

SI(A) is the sequence of Sum Input data with $FT_{1-3} = LOW$ (three zero stages).



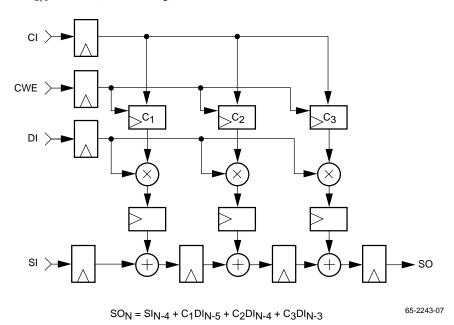
SI(B) is the sequence of Sum Input data with $FT_1 = HIGH$ and $FT_{2,3} = LOW$ (two zero stages).



Cycle	SI	DI	CI	CWE	so
0	SI ₀	DI ₀	A ₁	01	X
1	SI ₁	DI ₁	A ₂	10	X
2	SI ₂	DI ₂	A ₃	11	X
3	SI ₃	DI ₃	B ₁	01	X
4	SI ₄	DI ₄	B ₂	10	X
5	SI ₅	DI ₅	В3	11	X
6	SI ₆	DI ₆	C ₁	01	SI ₂ + A ₁ DI ₁ + A ₂ DI ₂ + A ₃ DI ₃
7	SI ₇	DI ₇	C ₂	10	SI ₃ + A ₁ DI ₂ + A ₂ DI ₃ + A ₃ DI ₄
8	SI ₈	DI ₈	C ₃	11	SI ₄ + A ₁ DI ₃ + A ₂ DI ₄ + A ₃ DI ₅
9	SI ₉	DI ₉		00	SI ₅ + B ₁ DI ₄ + B ₂ DI ₅ + B ₃ DI ₆
10	SI ₁₀	DI ₁₀		00	SI ₆ + B ₁ DI ₅ + B ₂ DI ₆ + B ₃ DI ₇
11	SI ₁₁	DI ₁₁		00	$SI_7 + B_1DI_6 + B_2DI_7 + B_3DI_8$
12	SI ₁₂	DI ₁₂		00	$SI_8 + C_1DI_7 + C_2DI_8 + C_3DI_9$
13	SI ₁₃	DI ₁₃		00	$SI_9 + C_1DI_8 + C_2DI_9 + C_3DI_{10}$

Figure 4. Typical Adaptive Filter Operation Sequence

with $FT_1 = LOW$ and $FT_{2, 3} = HIGH$ (one zero stage)



Equivalent Circuits and Transition Levels

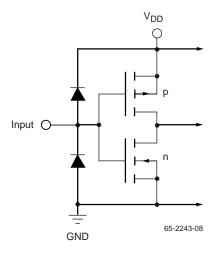


Figure 5. Equivalent Input Circuit

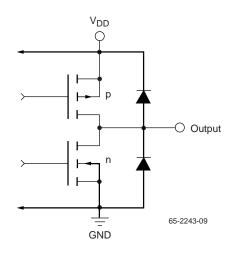


Figure 6. Equivalent Output Circuit

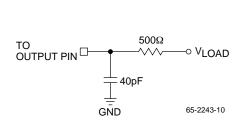
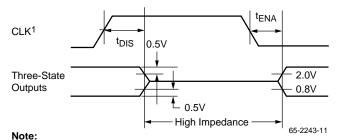


Figure 7. Test Load



1. Assumes \overline{OE} has gone LOW, within the Input Setup requirements.

Figure 8. Transition Levels for Three-State Measurements

Absolute Maximum Ratings (beyond which the device may be damaged)¹

Parameter	Min	Max	Unit
Supply Voltage	-0.5	7.0	V
Input Voltage	-0.5	VDD + 0.5	V
Output Applied Voltage ²	0.5	VDD + 0.5	V
Output Forced Current ^{3,4}	-1.0	6.0	mA
Short circuit duration (single output in HIGH state to ground)		1	sec
Operating, Case Temperature	-60	130	°C
Operating, Junction Temperature		175	°C
Storage Temperature	-65	150	°C
Lead Soldering Temperature (10 seconds)		300	°C

- 1. Absolute maximum ratings are limiting values applied individually while all parameters are within specified operating conditions. Functional operation under any of these conditions is NOT implied.
- 2. Applied voltage must be current limited to specified range, and measured with respect to GND.
- 3. Forcing voltage must be limited to specified range.
- 4. Current is specified as conventional current flowing into the device.

Operating Conditions

				Te	mperat	ure Rar	nge		
				Standar	d	E	xtende	d	
Param	eter	Test Conditions	Min	Nom	Max	Min	Nom	Max	Units
VDD	Supply Voltage		4.75	5.0	5.25	4.5	5.0	5.5	V
VIL	Input Voltage, Logic LOW		2.0			2.0			V
VIH	Input Voltage, Logic HIGH				0.8			0.8	V
loL	Output Current, Logic LOW				4.0			4.0	mA
Іон	Output Current, Logic HIGH				-2.0			-2.0	mA
tCY	Cycle Time	V _{DD} = Min	50			50			ns
tpwL	Clock Pulse Width LOW	V _{DD} = Min	20			20			ns
tPWH	Clock Pulse Width HIGH	V _{DD} = Min	20			20			ns
ts	Input Setup Time		15			20			ns
ts(SI)	Input Setup Time	Sl _{21-6,} FT ₁ = HIGH	25			28			ns
		FT ₁ = LOW	18			20			ns
tH	Input Hold Time		2			3			ns
tH(SI)	Input Hold Time, SI ₂₁₋₆		5			5			ns
TA	Ambient Temperature, Still Air		0		70				°C
TC	Case Temperature					-55		125	°C

DC Characteristics¹

			Temp		ure Rar	nge	
			Stan	dard	Exte	nded	
Param	eter	Test Conditions	Min	Max	Min	Max	Units
IDDQ	Supply Current, Quiescent	V _{DD} = Max, V _{IN} = OV, OE = HIGH		15		15	mA
IDDU	Supply Current, Unloaded	V _{DD} = Max, OE = HIGH					
		f = 20MHz		90		90	mA
		f = 10MHz		48		48	mA
ΠL	Input Current, Logic LOW	V _{DD} = Max, V _{IN} = OV	-75	75	-75	75	μΑ
lн	Input Current, Logic HIGH	V _{DD} = Max, V _{IN} = V _{DD}	-75	75	-75	75	μΑ
VOL	Output Voltage, Logic LOW	V _{DD} = Min, I _{OL} = Max		0.4		0.4	V
Vон	Output Voltage, Logic HIGH	V _{DD} = Min. I _{OH} = Max	2.4		2.4		V
lozL	Hi-Z Output Leakage Current, Output LOW	V _{DD} = Max, V _{IN} = OV	-40	40	-40	40	μА
lozh	Hi-Z Output Leakage Current, Output HIGH	$V_{DD} = Max, V_{IN} = V_{DD}$	-40	40	-40	40	μА
los	Short-Circuit Output Current	VDD = Max, Output HIGH, one pin to ground, one second duration max		-150		-150	mA
Cı	Input Capacitance	T _A = 25°C, f = 1MHz		10		10	pF
C ₀	Input Capacitance	T _A = 25°C, f = 1MHz		10		10	pF

^{1.} Actual test conditions may vary from those shown, but guarantee operation as specified.

AC Characteristics

			Temperature Range		nge		
			Stan	Standard Extend		nded	
Paran	neter	Test Conditions	Min	Max	Min	Max	Units
tD	Output Delay	V _{DD} = Min, C _{LOAD} = 40pF		30		30	ns
tDC	Output Delay, Cascaded	VDD = Min, CLOAD = 10pF		20		20	ns
tHO	Output Hold Time	V _{DD} = Max, C _{LOAD} = 40pF	5		5		ns
tENA	Three-State Output, Enable Delay ¹	VDD = Min, CLOAD = 40pF		20		25	ns
tDIS	Three-State Output, Disable Delay ¹	V _{DD} = Min, C _{LOAD} = 40pF		15		20	ns

Note:

1. All transitions are measured at a 1.5V level except for t_{DIS} and t_{ENA}.

Application Discussion

Loading and Updating of Coefficients

Because of the TMC2243's internal architecture, its impulse response is C_3 , C_2 , C_1 , where C_3 is the rightmost coefficient and C_1 is the leftmost. However, for glitchless performance, coefficients must be updated from left to right: C_1 then C_2 then C_3 .

For example, consider an adaptive filter whose first set of coefficients is A_i , second set is B_i and third set is C_i (Figure 4). First, the TMC2243 is initialized with A_i . If these are loaded in numerical (left to right) sequence, two of the first three data points can be loaded with them, as shown in Figure 4. Immediately after the third coefficient is loaded, the first coefficient of the next set can be loaded, if desired, along with the third data point.

Table 1. Impulse Response

FT ₃₋₁		Response					
000	C ₃	0	C ₂	0	C ₁	0	
001	C ₃	0	C ₂	0	C ₁		
010	C ₃	0	C ₂	C ₁	0		
011	C ₃	0	C ₂	C ₁			
100	C ₃	C ₂	0	C ₁	0		
101	C ₃	C ₂	0	C ₁			
110	C ₃	C ₂	C ₁	0			
111	C ₃	C ₂	C ₁				

Notes:

- 1. C₃ is she rightmost coefficient, C₁ is the leftmost
- FT₁ is relevant only it SUMIN is used. When multiple chips are cascaded, FT₁ LOW places a zero stage between their concatenated impulse responses.

Building Longer Filters

To build a filter of more than three non-zero stages, merely concatenate a series of TMC2243s. The coefficient inputs may be connected to the data bus, a separate common coefficient bus, or separate buses, depending on system architecture, memory and bus resources, and coefficient updating requirements. The data inputs are connected to a common bus. If the first feedthrough register is used (and a zero stage is not desired there), an external register should be inserted in the data input path for proper timing (Figure 9).

The 16-bit Sum-Out port of each TMC2243 is connected to the Sum-In port of the next TMC2243 in the chain; the filter output is the Sum-Out port of the last TMC2243. Since the 6 LSBs of each TMC2243's accumulation pipeline are not output, each TMC2243 incorporates a rounding increment of 1 into the sixth bit, to minimize bias.

When TMC2243s are cascaded in this fashion, the minimum permissible clock period is the sum of the output delay and the Sum-In port's input setup time. When the Input Registers are enabled (that is, $FT_1 = LOW$), full 20MHz performance can be obtained.

All data and coefficient inputs and outputs are two's complement representation, whose relative scaling is presented in the Data Formats table, Figure 1. Although the data values are shown in fractional format, the user can arbitrarily rescale them, as long as consistency is maintained.

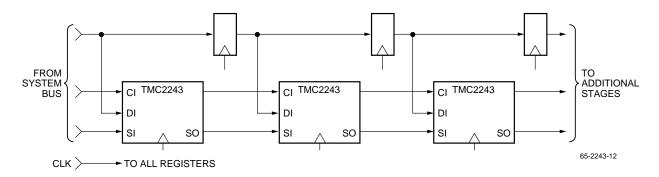


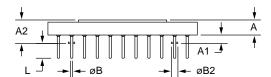
Figure 9. Basic Diagram for Stacking the TMC2243 for High–Speed Operation (no zero tap desired between each TMC2243, all FT₁ = LOW)

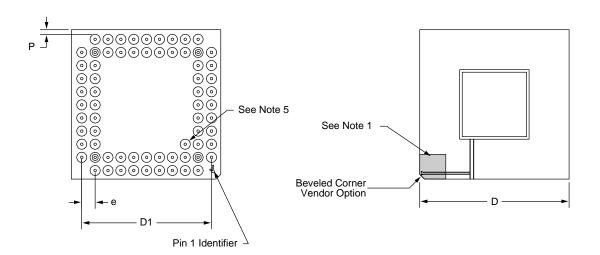
Mechanical Dimensions

68 Lead PGA Package

	Inc	hes	Millin		
Symbol	Min.	Max.	Min.	Max.	Notes
A	.080	.125	2.03	3.18	
A1	.040	.060	1.02	1.52	
A2	.115	.190	2.92	4.83	
øΒ	.017	.020	0.43	0.51	
øB2	.050 I	NOM.	1.27	NOM.	
D	1.140	1.180	28.96	29.97	
D1	1.000	BSC	25.40	BSC	
е	.100	BSC	2.54	BSC	
L	.120	.140	3.05	3.56	
М	1	1	11		2
N	6	8	6	8	3
Р	.003	_	.076	_	

- 1. Pin #1 identifier shall be within shaded area shown.
- 2. Dimension "M" defines matrix size.
- 3. Dimension "N" defines the maximum possible number of pins.
- 4. Controlling dimension: inch.
- 5. Optional index pin.



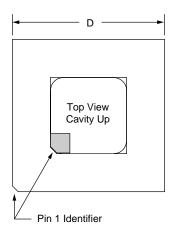


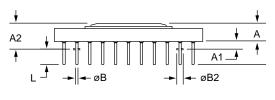
Mechanical Dimensions (continued)

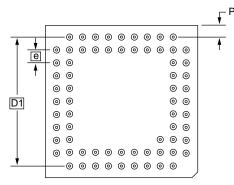
69 Lead PGA Package

Cumbal	Inc	hes	Millin	Notes	
Symbol	Min.	Max.	Min.	Max.	Notes
А	.080	.160	2.03	4.06	
A1	.040	.060	1.01	1.53	
A2	.125	.215	3.17	5.46	
øΒ	.016	.020	0.40	0.51	2
øB2	.050 l	NOM.	1.27	NOM.	2
D	1.140	1.180	28.96	29.97	SQ
D1	1.000	BSC	25.40	BSC	
е	.100	BSC	2.54	BSC	
L	.110	.145	2.79	3.68	
М	1	1	11		3
N	6	8	68		4
Р	.003	_	.076	_	

- 1. Pin #1 identifier shall be within shaded area shown.
- 2. Pin diameter excludes solder dip finish.
- 3. Dimension "M" defines matrix size.
- 4. Dimension "N" defines the maximum possible number of pins.
- 5. Orientation pin is at supplier's option.
- 6. Controlling dimension: inch.







Ordering Information

Product Number	Product Number Temperature Range		Package	Package Marking
TMC2243G8C	STD- $T_A = 0^{\circ}C$ to $70^{\circ}C$	Commercial	68 Pin Grid Array	2243G8C
TMC2243G8V	EXT-TC = 55° C to 125° C	MIL-STD-883	68 Pin Grid Array	2243G8V
TMC2243H8C	STD- $T_A = 0^{\circ}C$ to $70^{\circ}C$	Commercial	69 Pin Plastic Pin Grid Array	2243H8C

LIFE SUPPORT POLICY

FAIRCHILD'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF FAIRCHILD SEMICONDUCTOR CORPORATION. As used herein:

- Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and (c) whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury of the user.
- A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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