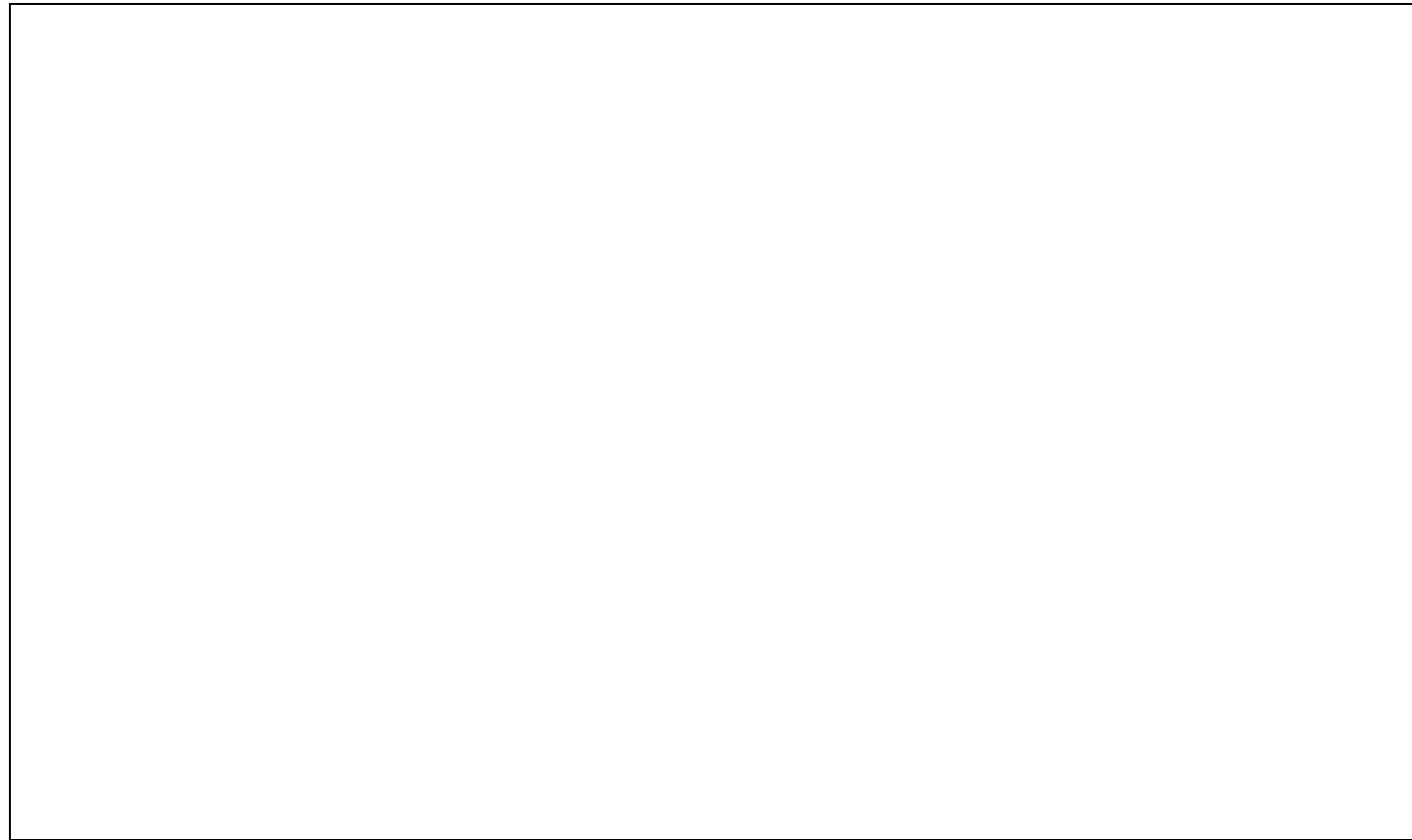


# **SIEMENS**



## **ICs for Consumer Electronics**

ADC with Built in Antialiasing filter and  
Clock generation UnitS  
ABACUS  
SDA 9206

Data Sheet 1999-02-10

**Edition 1999-02-10**

This edition was realized using  
the software system FrameMaker®

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**SIEMENS**

## ICs for Consumer Electronics

ADC with Built in Antialiasing filter and  
Clock generation UnitS

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<b>SDA 9206</b>		
<b>Revision History:</b>	<b>Current Version: 1999-02-10</b>	
Previous Version:		
Page (in previous Version)	Page (in current Version)	Subjects (major changes since last revision)
21	21	Update of <b>Table 2</b> concerning <b>Straight Binary</b>

## Data Classification

### Maximum Ratings

Maximum ratings are absolute ratings; exceeding only one of these values may cause irreversible damage to the integrated circuit.

### Recommended Operating Conditions

Under this conditions the functions given in the circuit description are fulfilled. Nominal conditions specify mean values expected over the production spread and are the proposed values for interface and application. If not stated otherwise, nominal values will apply at  $T_A=25^\circ\text{C}$  and the nominal supply voltage.

### Characteristics

The listed characteristics are ensured over the operating range of the integrated circuit.

### Edition 1999-02-10

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## ADC with Built in Antialiasing filter and Clock generation UnitS ABACUS

SDA 9206

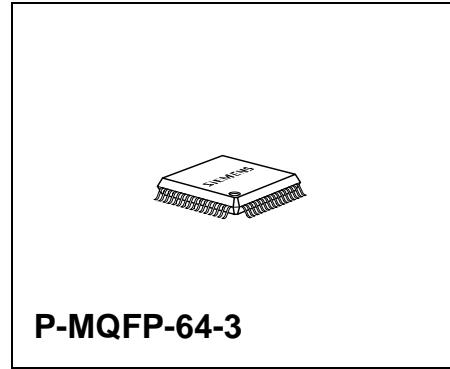
### Preliminary Data

CMOS

## 1 Overview

### 1.1 Features

- Three equivalent CMOS A/D converters on chip
- 30 MHz sampling rate
- 8-Bit resolution
- No external sample & hold required
- Internal clamping circuits for each of the ADCs
- Internal amplification of input signals can be set by I<sup>2</sup>C Bus
- Internal pre-filtering of analog input signals
- High performance decimation filters
- Two data sampling modes (4:2:2 and 4:1:1)
- 3 output data interfaces
  - CCIR 656 interface (8 wires)
  - Parallel data interface (2 x 8 wires)
  - Quasi Parallel data interface (8 + 4 wires)
- Overflow and underflow I<sup>2</sup>C status bits
- On-chip sync and clock generation
- Separate SYNC input with clamping for sync and clock generation (max. line frequency of SYNC input: 38 kHz)
- positive and negative polarity of SYNC signal (switchable by I<sup>2</sup>C Bus)
- Lock-in behavior can be set via I<sup>2</sup>C Bus
- Frequency generator function possible with digitally adjustable frequency
- Clock generation for single and double line input frequencies supported (1fh / 2fh mode)
- Vertical noise suppression and 50/60 Hz detection (for 1fh mode only)



Type	Ordering Code	Package
SDA 9206	Q67101-H5185-A704	P-MQFP-64-3

- 
- I<sup>2</sup>C-Bus interface
  - P-MQFP-64-3
  - 5 V supply voltage for input signals
  - 3.3 V or 5 V supply voltage for output signals

## 1.2 General Description

The SDA 9206 is a single monolithic IC containing three separate 8-Bit A/D converters for video (YUV) applications and a clock sync generator which is delivering the sample clock for the A/D converters. It utilizes an advanced VLSI 0.5 µm CMOS process providing 30 MHz sampling rates at 8-Bits.

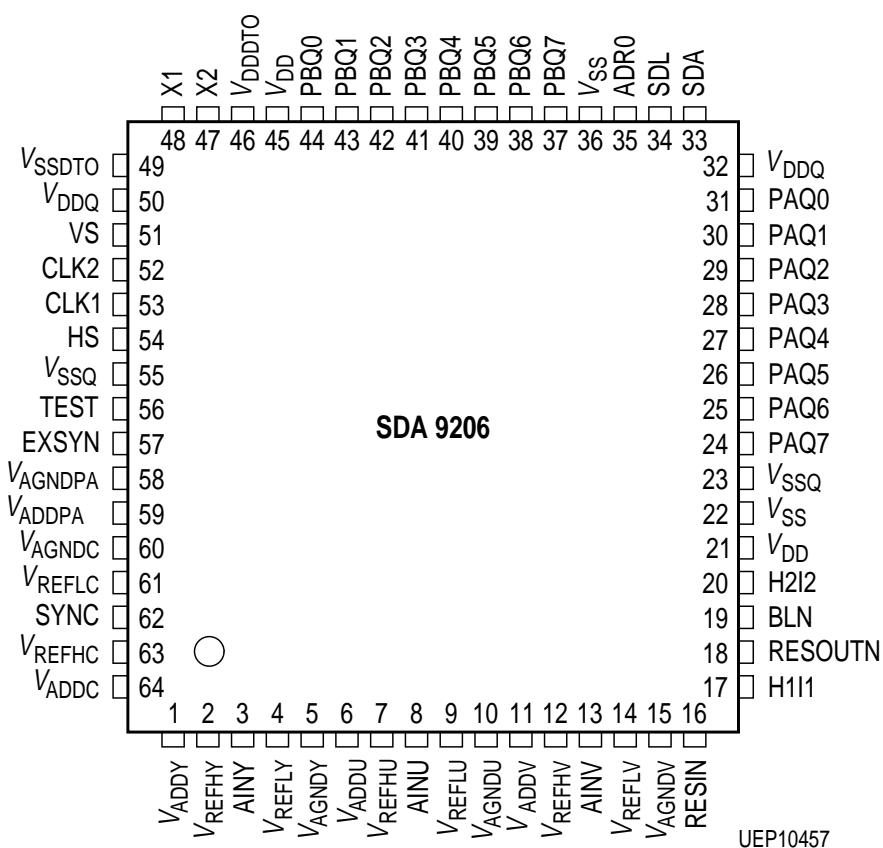
The YUV processing consists of following functional blocks:

- Analog input buffers and clamping circuits
- Three 30 MHz A/D converters
- Digital decimation filters
- Delay compensation in Y-path
- Output formatter and buffer

The clock sync generator consists essentially of the following functional blocks:

- Analog clamping
- 7-Bit A/D converter
- Sync processor with digital horizontal PLL, vertical sync processor and pulse generator
- Clock generator with discrete timing oscillator, D/A converter, analog PLL and divider, as well as a crystal oscillator

### 1.3 Pin Configuration



**Figure 1**

## 1.4 Pin Description

Pin No.	Symbol	Type	Description
22, 36	$V_{SS}$	S	Supply ground ( $V_{SS}$ ) for digital parts
21, 45	$V_{DD}$	S	Supply voltage ( $V_{DD}$ ) for digital parts
23, 55	$V_{SSQ}$	S	Supply ground for output stages and input stages
32, 50	$V_{DDQ}$	S	Supply voltage for output stages and input stages (3.3 V / 5 V)
1	$V_{ADDY}$	S	Analog positive supply voltage of ADC AINY (5 V)
2	$V_{REFHY}$		Reference voltage high of ADC AINY (4.2 V)
3	AINY	I/ana	Analog voltage input of ADC AINY input range selectable via I <sup>2</sup> C Bus (subaddress 11 <sub>H</sub> , YAMP)
4	$V_{REFLY}$		Reference voltage low of ADC AINY (2.2 V)
5	$V_{AGNDY}$	S	Analog ground of ADC AINY
6	$V_{ADDU}$	S	Analog positive supply voltage of ADC AINU (5 V)
7	$V_{REFHU}$		Reference voltage high of ADC AINU (4.2 V)
8	AINU	I/ana	Analog voltage input of ADC AINU input range selectable via I <sup>2</sup> C Bus (subaddress 12 <sub>H</sub> , UAMP)
9	$V_{REFLU}$		Reference voltage low of ADC AINU (2.2 V)
10	$V_{AGNDU}$	S	Analog ground of ADC AINU
11	$V_{ADDV}$	S	Analog positive supply voltage of ADC AINV (5 V)
12	$V_{REFHV}$		Reference voltage high of ADC AINV (4.2 V)
13	AINV	I/ana	Analog voltage input of ADC AINV input range selectable via I <sup>2</sup> C Bus (subaddress 12 <sub>H</sub> , VAMP)
14	$V_{REFLV}$		Reference voltage low of ADC AINV (2.2 V)
15	$V_{AGNDV}$	S	Analog ground of ADC AINV
16	RESIN	I/TTL/pu	Reset input signal: active low
17	H1I1	Q/TTL	Pin function defined by I <sup>2</sup> C Bus: Line frequent pulse output or programmable digital control output
18	RESOUTN	Q/TTL	Reset output signal: active low; reset for other ICs
19	BLN	Q/TTL	Blanking signal output, high level indicates active video line

## 1.4 Pin Description (cont'd)

Pin No.	Symbol	Type	Description
20	H2I2	Q/TTL	Pin function defined by I <sup>2</sup> C Bus: Line frequent pulse output or programmable digital control output
24 ... 31	PAQ7 ... 0	Q/TTL	Data output Port A ( <b>see Data Format</b> )
33	SDA	I/Q	I <sup>2</sup> C-Bus data line
34	SCL	I	I <sup>2</sup> C-Bus clock line
35	ADR0	I/TTL/pd	I <sup>2</sup> C-Chip select
37 ... 44	PBQ7 ... 0	Q/TTL	Data output port B ( <b>see Data Format</b> )
46	$V_{DDDTO}$	S	Positive supply voltage of DTO (5 V)
47	X2	Q/ana	Crystal connection
48	X1	I/ana	Crystal connection (clock input)
49	$V_{SSDTO}$	S	Ground of DTO
51	VS	Q/TTL	Vertical sync pulse output
52	CLK2	Q/TTL	Clock out: tristate / 6.75 / 13.5 / 27 MHz; selectable via I <sup>2</sup> C
53	CLK1	Q/TTL	Clock out: tristate / 6.75 / 13.5 / 27 MHz; selectable via I <sup>2</sup> C
54	HS	Q/TTL	Horizontal sync pulse output
56	TEST	I/TTL/pd	Input signal for test mode selection (0 V: no test mode selected) Leave unconnected or connect to $V_{ss}$
57	EXSYN	I/TTL/pd	Input signal for test mode selection (0 V: no test mode selected) Leave unconnected or connect to $V_{ss}$
58	$V_{AGNDPA}$		Analog ground of analog PLL and DACs
59	$V_{ADDPA}$		Analog positive supply voltage of analog PLL and DACs (5 V)
60	$V_{AGNDC}$		Analog ground of ADC SYNC
61	$V_{REFLC}$		Reference voltage low of ADC SYNC (2.2 V)
62	SYNC	I/ana	SYNC input Input range selectable via I <sup>2</sup> C Bus (subaddress 11 <sub>H</sub> , SYNAMP)

## 1.4 Pin Description (cont'd)

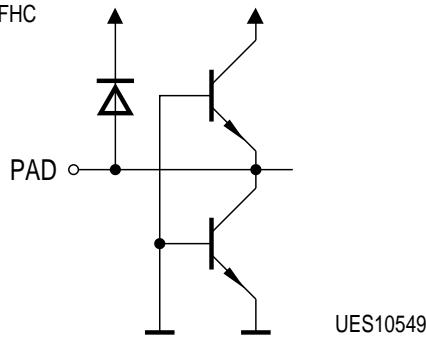
Pin No.	Symbol	Type	Description
63	$V_{REFHC}$		Reference voltage high of ADC SYNC (4.2 V)
64	$V_{ADDC}$		Analog positive supply voltage of ADC SYNC (5 V)

S: supply, I: input, Q: output, TTL: digital (TTL)

ana: analog, pu: internal pullup-circuit, pd: internal pulldown-circuit

## 1.5 Internal Pin Configuration

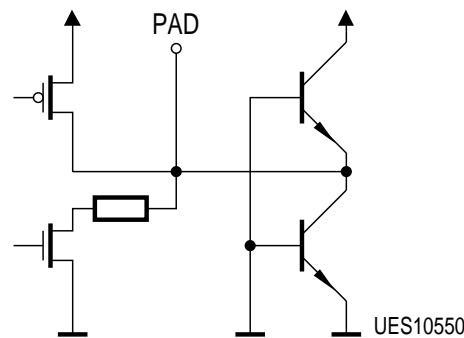
Pin 2, 4, 7, 9, 12, 14, 61, 63  
 $V_{REFHY}$ ,  $V_{REFLY}$ ,  $V_{REFHU}$ ,  $V_{REFLU}$ ,  $V_{REFHV}$ ,  $V_{REFLV}$ ,  
 $V_{REFLC}$ ,  $V_{REFHC}$



UES10549

Figure 2

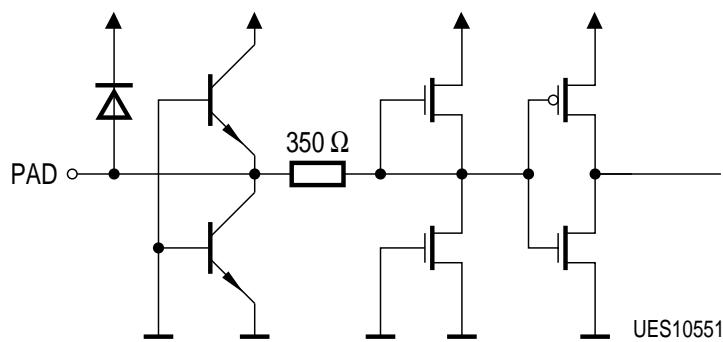
Pin 17, 18, 19, 20, 24...31, 37...44, 51, 52, 53, 54  
 H1I1, RESOUTN, BLN, H2I2, PAQ7...0, PBQ7...0, VS, CLK2,  
 CLK1, HS



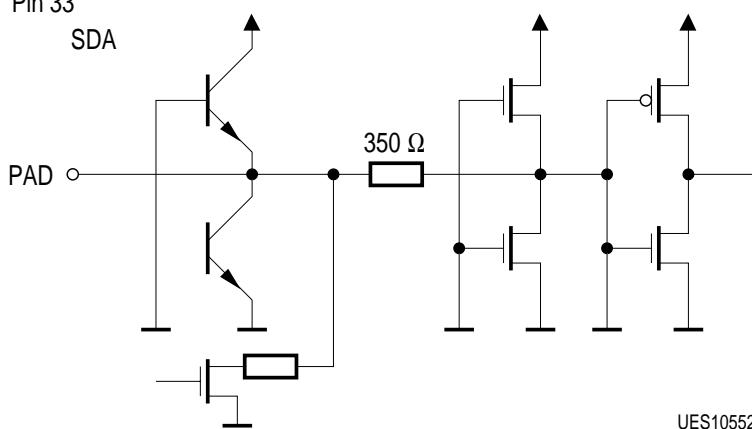
UES10550

Figure 3

Pin 16, 35, 56, 57  
RESIN, ADR0, TEST, EXSYN

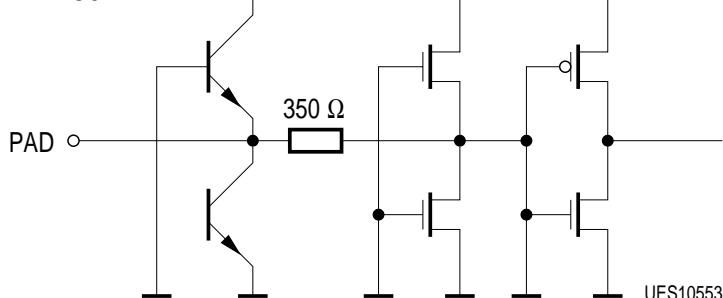
**Figure 4**

Pin 33  
SDA

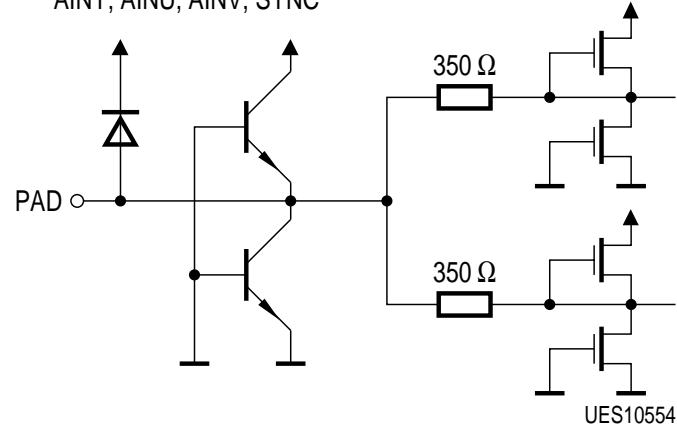
**Figure 5**

Pin 34

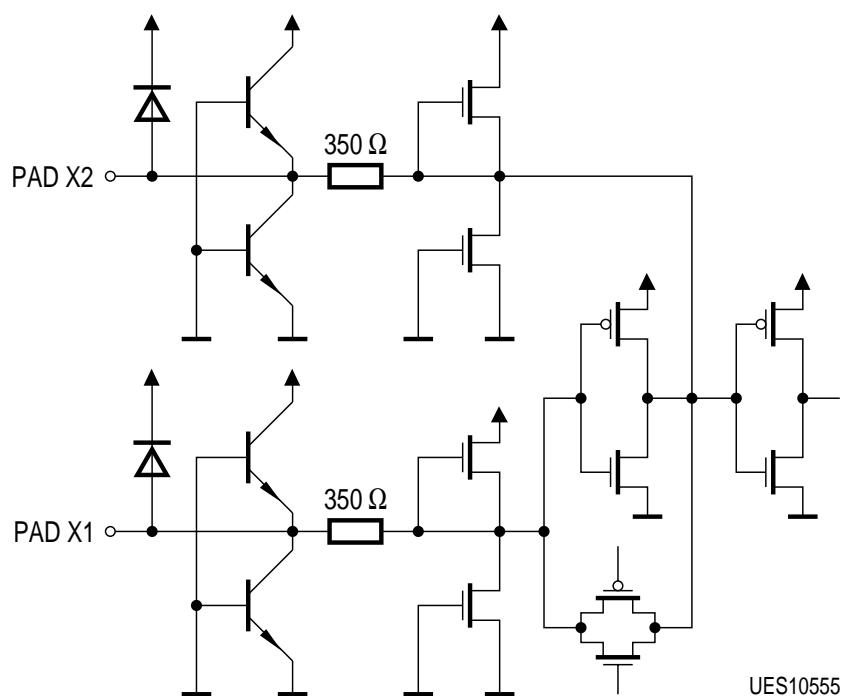
SCL

**Figure 6**

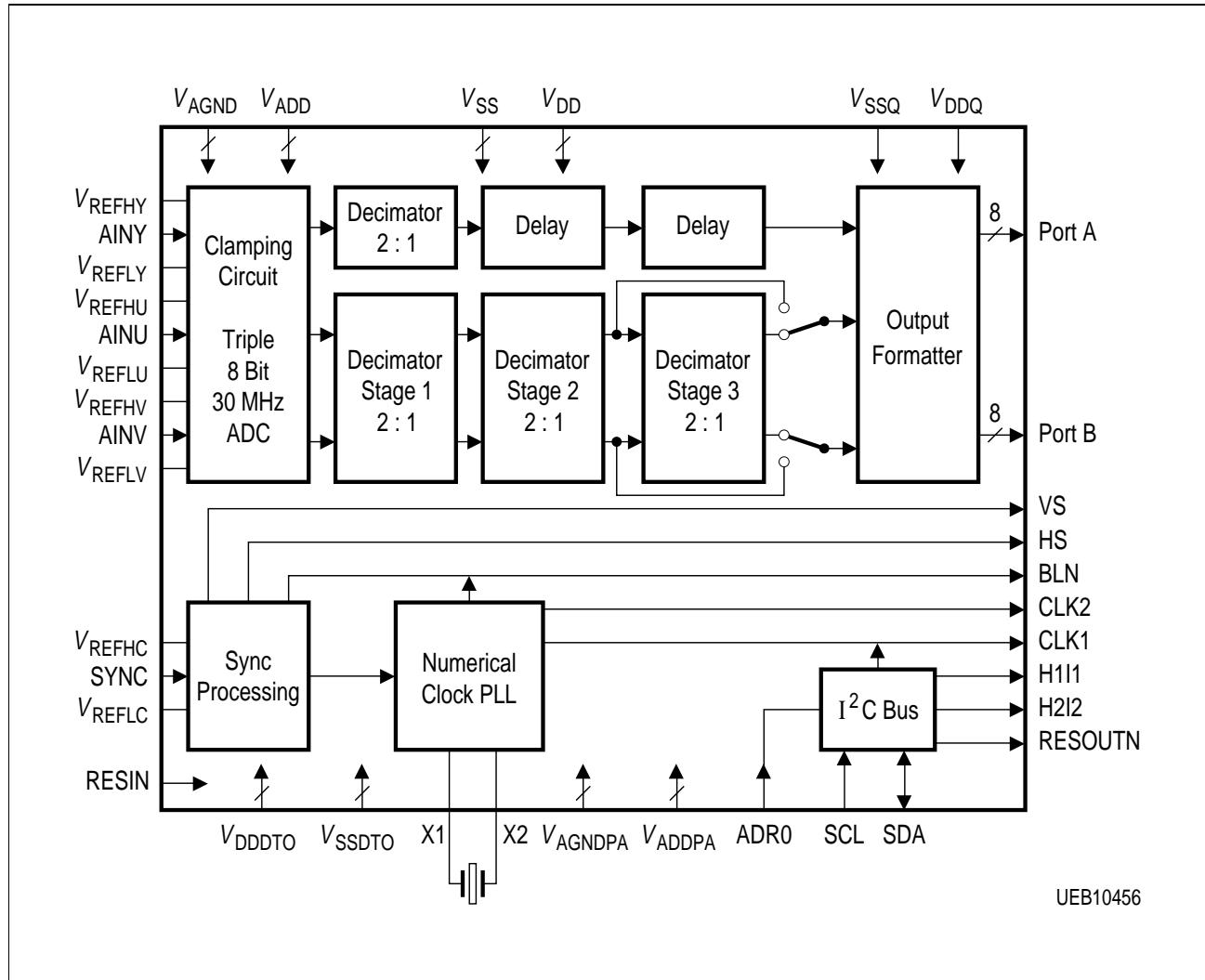
Pin 3, 8, 13, 62  
AINY, AINU, AINV, SYNC

**Figure 7**

Pin 47, 48  
X2, X1

**Figure 8**

## 1.6 Block Diagram



**Figure 9**

## 2 System Description

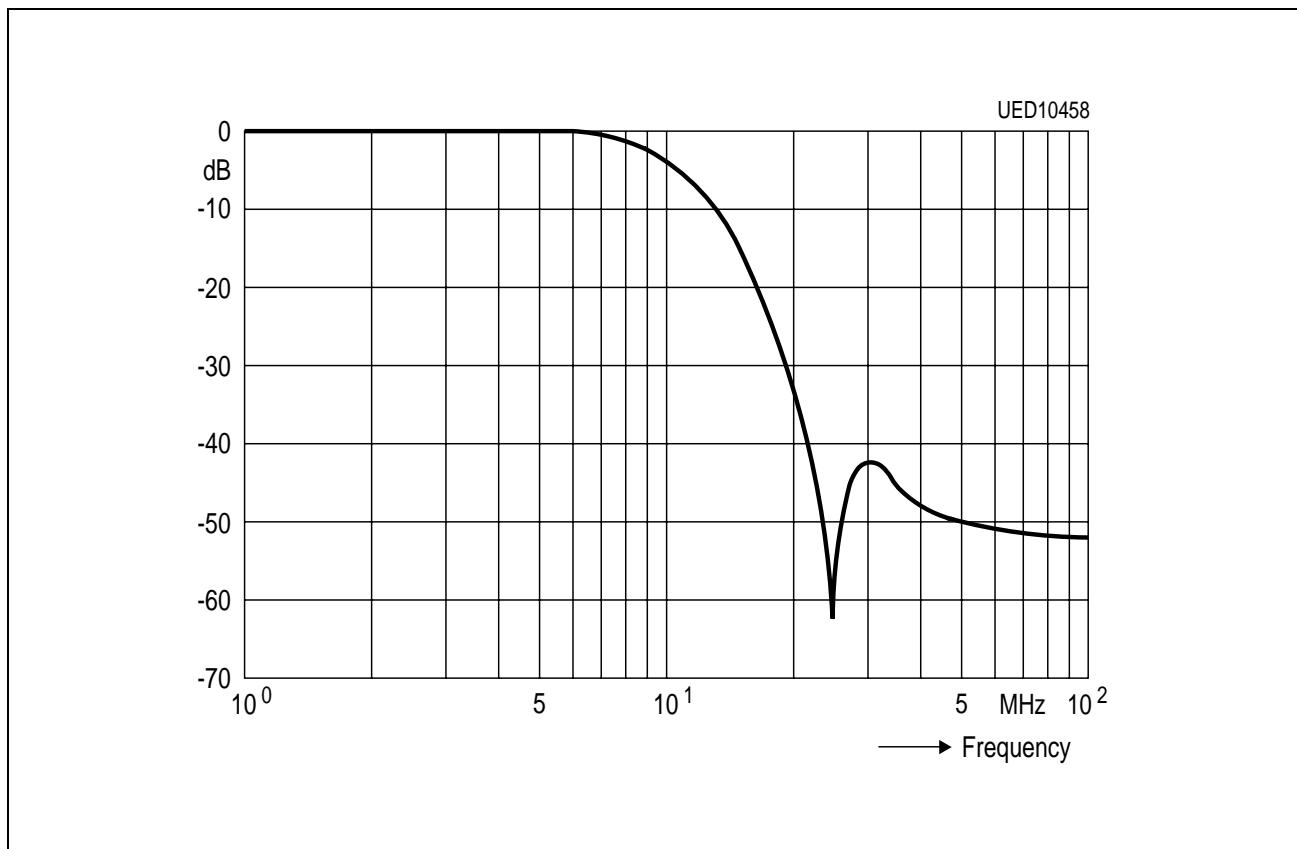
### 2.1 A/D Converter for YUV Inputs

#### 2.1.1 Introduction

The SDA 9206 implements 3 independent 8-Bit A/D converters.  
Maximum conversion rate is 30 MHz.

#### 2.1.2 Input Signal Amplification, Prefiltering

The amplification of the input signals can be adjusted via I<sup>2</sup>C Bus. An internal prefiltering of the analog input signals is implemented. The typ. frequency response of the analog antialiasing prefilter is shown in **figure 10**.



**Figure 10**  
**Frequency Response of the Analog Antialiasing Prefilter**

### 2.1.3 Clamping

The analog pins AINY, AINU, AINV are switched simultaneously to on chip generated clamping levels by an on chip clamping pulse H2.

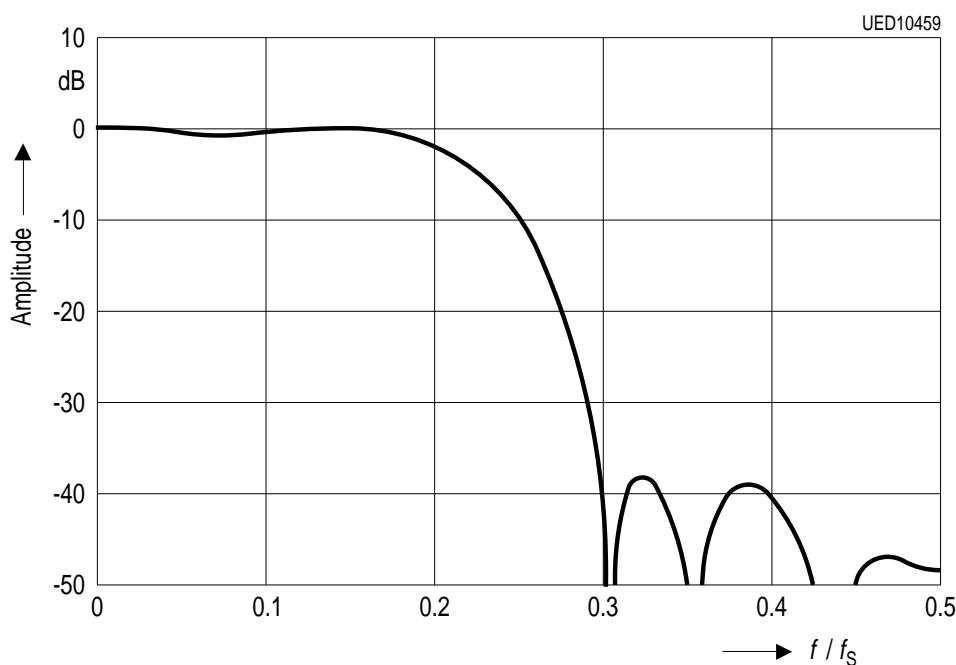
Analog Channel	Straight Binary Code	Two's Complement Code	Components
AINY	0001 0000	1001 0000	Y
AINU, AINV	1000 0000	0000 0000	U, V

The external clamping capacitance is loaded by on chip current sources during clamping. So loading time depends on the values of  $C_{\text{ext cl}}$ .

### 2.1.4 Digital Decimation Filters for YUV

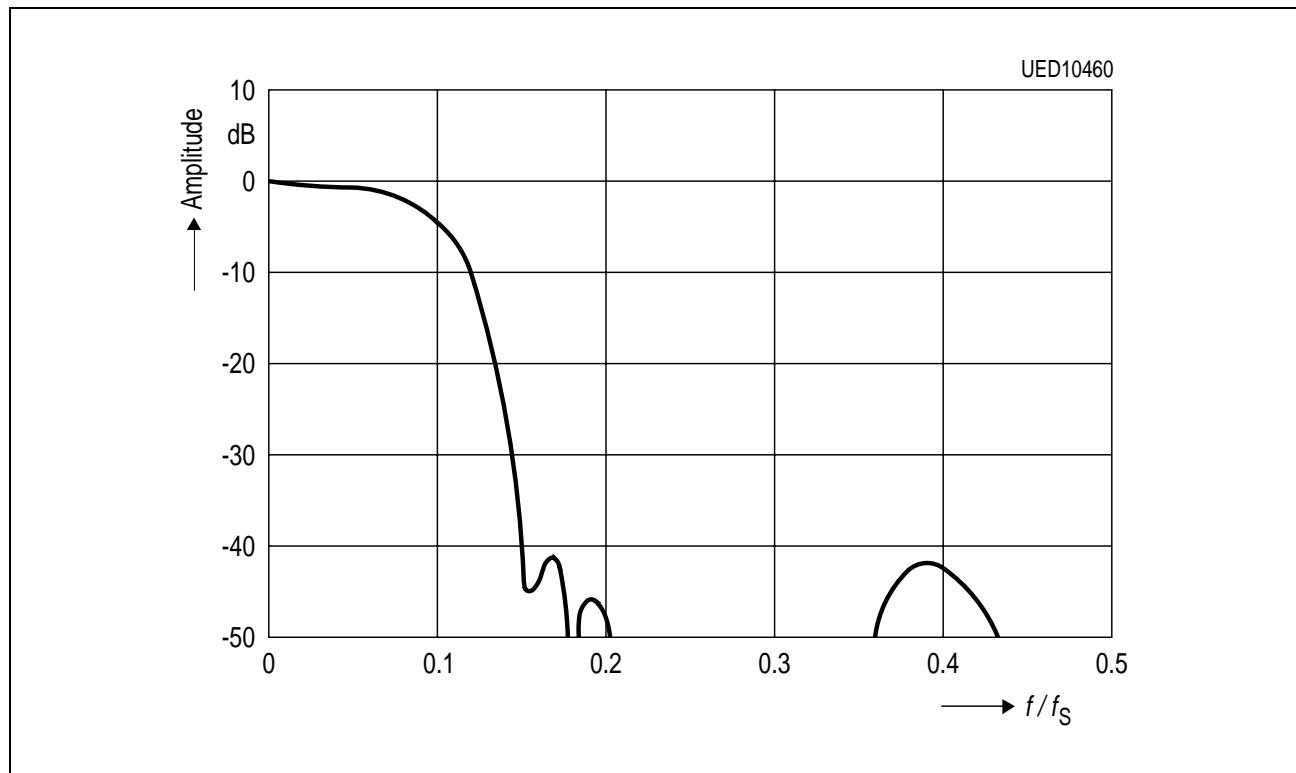
The data rates of digital YUV signals are reduced in decimation filters following the A/D conversion. The overall performance of the decimation filters is tuned to the requirements for TV signals.

In **figure 11** the frequency response of the filter for the Y channel is shown. The input sampling rate is 27 MHz, the output sampling rate is 13.5 MHz.



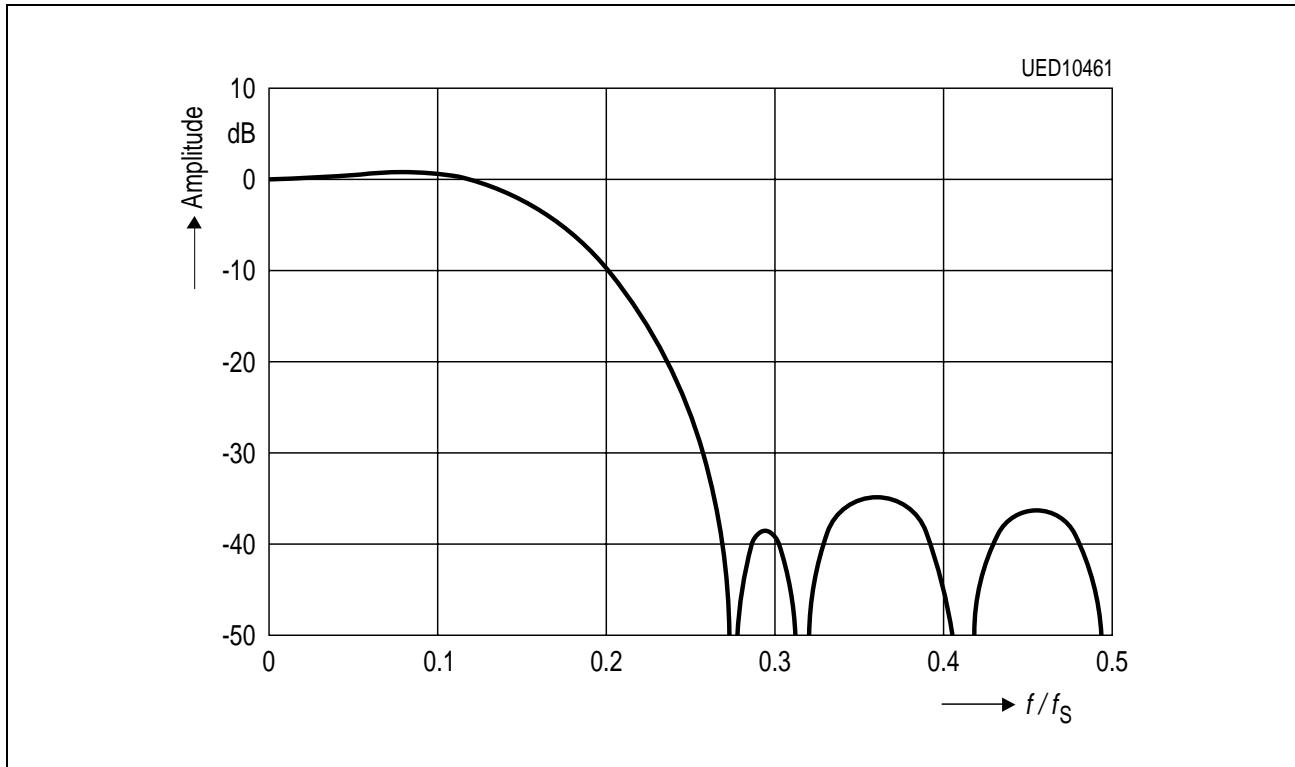
**Figure 11**  
**Magnitude Frequency Response of the Luminance Filter**  
**The Input Sampling Frequency  $f_s$  is 27 MHz**

The total frequency response of the decimator stages 1 and 2 of the UV channels for an input sampling rate of 27 MHz and an output sampling rate of 6.75 MHz is shown in **figure 12**.



**Figure 12**  
**Magnitude Frequency Response for Chroma Signals (Decimator Stages 1 and 2)**  
**The Input Sampling Frequency  $f_s$  is 27 MHz**

The frequency response of the decimator filter stage 3 of the UV channels for an input sampling rate of 6.75 MHz and an output sampling rate of 3.375 MHz is shown in **figure 13**. The decimator stage 3 is active for 4:1:1 mode and can also be activated for 4:2:2 mode by I<sup>2</sup>C Bus (control bit UV3FIL).



**Figure 13**  
**Frequency Response of the Chroma Decimator Stage 3**  
**The Input Sampling Frequency  $f_S$  is 6.75 MHz**

## 2.2 Data Output Formatter

Three output data formats can be selected via I<sup>2</sup>C Bus (control Bits FORMAT). One format corresponds to CCIR 656 (8-Bit bus at a data rate of 27 MHz), an other format makes available Y and UV data separately on 2 parallel 8-Bit buses for Y and UV at a data rate of 13.5 MHz each. The third format is a 12-Bit bus with 8 connections for Y and 4 connections for multiplexed UV data.

Output Pin	Quasiparallel Data FORMAT = 10 or 11 (13.5 MHz)				Parallel Data FORMAT = 01 (13.5 MHz)		CCIR 656 FORMAT = 00 (27 MHz)			
PAQ7	Y <sub>07</sub>	Y <sub>17</sub>	Y <sub>27</sub>	Y <sub>37</sub>	Y <sub>07</sub>	Y <sub>17</sub>	U <sub>07</sub>	Y <sub>07</sub>	V <sub>07</sub>	Y <sub>17</sub>
PAQ6	Y <sub>06</sub>	Y <sub>16</sub>	Y <sub>26</sub>	Y <sub>36</sub>	Y <sub>06</sub>	Y <sub>16</sub>	U <sub>06</sub>	Y <sub>06</sub>	V <sub>06</sub>	Y <sub>16</sub>
PAQ5	Y <sub>05</sub>	Y <sub>15</sub>	Y <sub>25</sub>	Y <sub>35</sub>	Y <sub>05</sub>	Y <sub>15</sub>	U <sub>05</sub>	Y <sub>05</sub>	V <sub>05</sub>	Y <sub>15</sub>
PAQ4	Y <sub>04</sub>	Y <sub>14</sub>	Y <sub>24</sub>	Y <sub>34</sub>	Y <sub>04</sub>	Y <sub>14</sub>	U <sub>04</sub>	Y <sub>04</sub>	V <sub>04</sub>	Y <sub>14</sub>
PAQ3	Y <sub>03</sub>	Y <sub>13</sub>	Y <sub>23</sub>	Y <sub>33</sub>	Y <sub>03</sub>	Y <sub>13</sub>	U <sub>03</sub>	Y <sub>03</sub>	V <sub>03</sub>	Y <sub>13</sub>
PAQ2	Y <sub>02</sub>	Y <sub>12</sub>	Y <sub>22</sub>	Y <sub>32</sub>	Y <sub>02</sub>	Y <sub>12</sub>	U <sub>02</sub>	Y <sub>02</sub>	V <sub>02</sub>	Y <sub>12</sub>
PAQ1	Y <sub>01</sub>	Y <sub>11</sub>	Y <sub>21</sub>	Y <sub>31</sub>	Y <sub>01</sub>	Y <sub>11</sub>	U <sub>01</sub>	Y <sub>01</sub>	V <sub>01</sub>	Y <sub>11</sub>
PAQ0	Y <sub>00</sub>	Y <sub>10</sub>	Y <sub>20</sub>	Y <sub>30</sub>	Y <sub>00</sub>	Y <sub>10</sub>	U <sub>00</sub>	Y <sub>00</sub>	V <sub>00</sub>	Y <sub>10</sub>
PBQ7	U <sub>07</sub>	U <sub>05</sub>	U <sub>03</sub>	U <sub>01</sub>	U <sub>07</sub>	V <sub>07</sub>	Z	Z	Z	Z
PBQ6	U <sub>06</sub>	U <sub>04</sub>	U <sub>02</sub>	U <sub>00</sub>	U <sub>06</sub>	V <sub>06</sub>	Z	Z	Z	Z
PBQ5	V <sub>07</sub>	V <sub>05</sub>	V <sub>03</sub>	V <sub>01</sub>	U <sub>05</sub>	V <sub>05</sub>	Z	Z	Z	Z
PBQ4	V <sub>06</sub>	V <sub>04</sub>	V <sub>02</sub>	V <sub>00</sub>	U <sub>04</sub>	V <sub>04</sub>	Z	Z	Z	Z
PBQ3	Z	Z	Z	Z	U <sub>03</sub>	V <sub>03</sub>	Z	Z	Z	Z
PBQ2	Z	Z	Z	Z	U <sub>02</sub>	V <sub>02</sub>	Z	Z	Z	Z
PBQ1	Z	Z	Z	Z	U <sub>01</sub>	V <sub>01</sub>	Z	Z	Z	Z
PBQ0	Z	Z	Z	Z	U <sub>00</sub>	V <sub>00</sub>	Z	Z	Z	Z

X<sub>AB</sub>: X: signal component A: sample number B: bit number

Z: Pin is in tristate mode.

The BLN signal marks the active part of the video line (**see figure 14**).

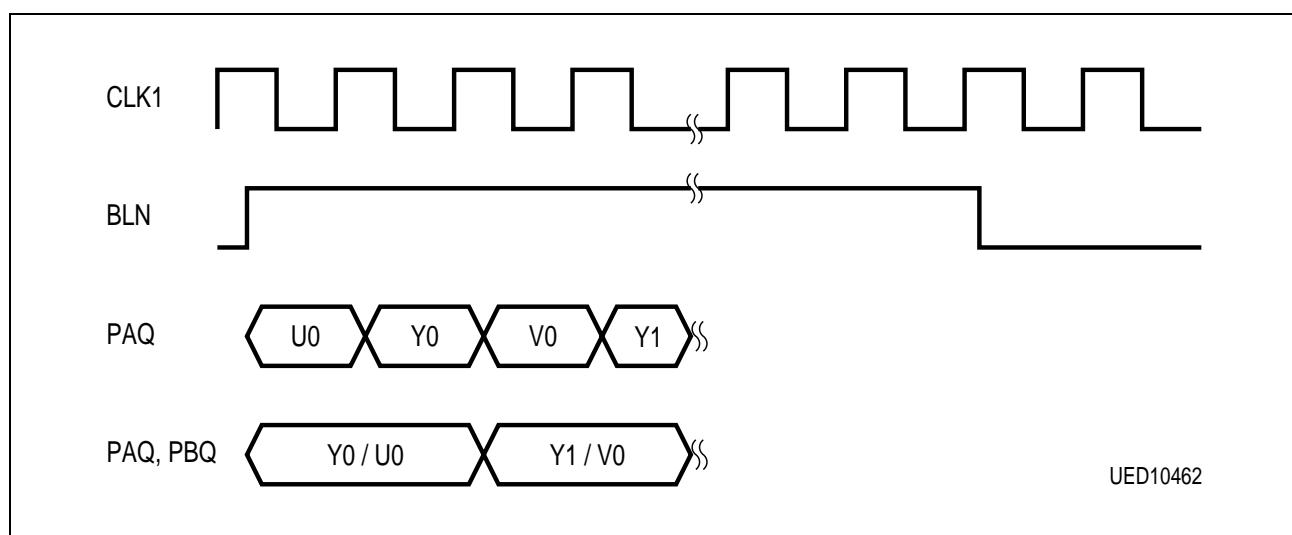


Figure 14

### 2.2.1 Output Coding for Straight Binary / Two's Complement Mode

Straight binary or Two's complement output coding is selectable for each separate signal component (Y and UV) via I<sup>2</sup>C-Bus control bits YCODE and UVCODE.

For straight binary coding a special suppression of code 0 and code 255 is provided in output format mode according CCIR 656.

**Table 1**  
**Output Coding**

Step	AINY	AINU, AINV	OFL Bit	UFL Bit	Straight Binary 7654 3210	Two's Complement 7654 3210
Underflow	< $V_{CY} - 0.125 \text{ V}$	< $V_{CU,v} - 1.0 \text{ V}$	0	1	0000 0000	1000 0000
0	$V_{CY} - 0.125 \text{ V}$	$V_{CU,v} - 1.0 \text{ V}$	0	0	0000 0000	1000 0000
1	$V_{CY} - 0.117 \text{ V}$	$V_{CU,v} - 0.992 \text{ V}$	0	0	0000 0001	1000 0001
2	$V_{CY} - 0.109 \text{ V}$	$V_{CU,v} - 0.984 \text{ V}$	0	0	0000 0010	1000 0010
•	•	•	•	•	•	•
•	•	•	•	•	•	•
253	$V_{CY} + 1.859 \text{ V}$	$V_{CU,v} + 0.984 \text{ V}$	0	0	1111 1101	0111 1101
254	$V_{CY} + 1.867 \text{ V}$	$V_{CU,v} + 0.992 \text{ V}$	0	0	1111 1110	0111 1110
255	$V_{CY} + 1.875 \text{ V}$	$V_{CU,v} + 1.0 \text{ V}$	0	0	1111 1111	0111 1111
Overflow	> $V_{CY} + 1.875 \text{ V}$	> $V_{CU,v} + 1.0 \text{ V}$	1	0	1111 1111	0111 1111

$V_{CY}$ ,  $V_{CU,v}$ : ext. clamping level during clamping at  $C_{ext\ cl}$  on channel AINY  
resp. AINU, AINV

**Table 1** is valid for  $V_{REFL} = 2.2 \text{ V}$  and  $V_{REFH} = 4.2 \text{ V}$ , xAMP = 0000

**Table 2****Output Coding in Case of CCIR 656 Format, FORMAT = 00**

Step	AINY	AINU, AINV	OFL Bit	UFL Bit	Straight Binary 7654 3210	Two's Complement 7654 3210
Underflow	< $V_{CY} - 0.125 \text{ V}$	< $V_{CU,v} - 1.0 \text{ V}$	0	1	0000 0001	1000 0000
0	$V_{CY} - 0.125 \text{ V}$	$V_{CU,v} - 1.0 \text{ V}$	0	0	0000 0001	1000 0000
1	$V_{CY} - 0.117 \text{ V}$	$V_{CU,v} - 0.992 \text{ V}$	0	0	0000 0001	1000 0001
2	$V_{CY} - 0.109 \text{ V}$	$V_{CU,v} - 0.984 \text{ V}$	0	0	0000 0010	1000 0010
•	•	•	•	•	•	•
•	•	•	•	•	•	•
253	$V_{CY} + 1.859 \text{ V}$	$V_{CU,v} + 0.984 \text{ V}$	0	0	1111 1101	0111 1101
254	$V_{CY} + 1.867 \text{ V}$	$V_{CU,v} + 0.992 \text{ V}$	0	0	1111 1110	0111 1110
255	$V_{CY} + 1.875 \text{ V}$	$V_{CU,v} + 1.0 \text{ V}$	0	0	1111 1110	0111 1111
Overflow	> $V_{CY} + 1.875 \text{ V}$	> $V_{CU,v} + 1.0 \text{ V}$	1	0	1111 1110	0111 1111

$V_{CY}$ ,  $V_{CU,v}$ : ext. clamping level during clamping at  $C_{ext\ cl}$  on channel AINY resp. AINU, AINV

**Table 2** is valid for  $V_{REFL} = 2.2 \text{ V}$  and  $V_{REFH} = 4.2 \text{ V}$ , xAMP = 0000

## 2.3 Clock Sync Generation

The clock sync generator is a phase locked loop that locks on a horizontal SYNC input signal and generates the clock signals as well as additional control output signals.

### 2.3.1 Horizontal PLL (HPLL)

The input signal SYNC may be either a CVBS signal or a composite sync signal. The polarity of the SYNC signal can be both positive or negative (I<sup>2</sup>C-Bit SYPOL). The edges of the SYNC input pulses should not be steeper than 100 ns. The frequency of the SYNC signal can be of normal or double line frequency (I<sup>2</sup>C-Bit 2FH).

The SYNC is clamped before A/D conversion. For DC-input signals clamping can be disabled (I<sup>2</sup>C-Bit CLOF). A/D conversion takes place with 7 bits and a nominal frequency of 27 MHz.

The digital HPLL filters the signal with a cutoff frequency of 1 MHz (2 MHz for 2fh mode). If 1fh mode is used the sampling frequency is decimated to 13.5 MHz. Following the low pass filtering a black- and sync bottom- level measurement takes place in order to calculate a threshold value. By means of this value the phase difference between the HPLL output and the SYNC input pulse is determined. Using a digital PI filter an increment is calculated from this for the Discrete Timing Oscillator (DTO). It is possible

to adapt the nominal frequency of the DTO by means of 5 I<sup>2</sup>C-Bus bits (INC4...INC0) such shifting the center frequency according to the momentary standard used.

For the different applications the following values of INC are allowed (values valid for a crystal frequency of 24.576 MHz):

Application	FH [Hz]	2FH	YUV-ADCs	INC
PAL	15625	0	active	6
NTSC	15750	0	active	6
PAL (100 Hz/VGA)	31250	1	inactive	6
NTSC (120 Hz/VGA)	31500	1	inactive	6
ATV	32400	1	inactive	8
MUSE	33750	1	inactive	11
Macintosh	35000	1	inactive	14
VGA	38000	1	inactive	21

**Note:** A change of INC causes spontaneous changes of the generated clock frequencies!

The DTO generates a saw-tooth with a frequency that is proportional to the increment. The saw-tooth is converted into a sinusoidal clock signal by means of a D/A converter and applied to an analog PLL which multiplies the frequency and minimizes residual jitter.

By means of the I<sup>2</sup>C bits S1CL and S2CL the output frequency on pins CLK1 and CLK2 can be set. In this manner a clock is provided that is line-locked with the SYNC-input signal. The ratio of these clock frequencies to the horizontal frequency of SYNC depends only on the I<sup>2</sup>C-Bus bits S1CL, S2CL, HPLL and 2FH.

For the different modes the following values of S1CL and S2CL are allowed:

Mode	YUV-ADC	2FH	S1CL	S2CL	f <sub>CLK1</sub> (MHz)	f <sub>CLK2</sub> (MHz)
CCIR	enabled	0	11	11	27	27
CCIR	enabled	0	11	00	27	tristate
4:2:2, 4:1:1	enabled	0	11	11	27	27
4:2:2, 4:1:1	enabled	0	10	00	13.5	tristate
VGA	disabled	1	01	11	6.25 ... 8.75	25 ... 35
VGA	disabled	1	10	11	12.5 ... 17.5	25 ... 35

The digital horizontal PLL supplies a noise-suppressed horizontal pulse.

During 1fh mode ( $2FH = 0$ ) the digital HPLL also supplies a noise-suppressed vertical pulse obtained by digital integration of the main equalizing pulses. An integration time of  $26.6 \mu s$  or  $11.3 \mu s$  can be set by the I<sup>2</sup>C Bus. This functionality is switched off during 2fh mode ( $2FH = 1$ ).

### 2.3.2 Vertical Sync Processing (only available for 1fh mode)

Vertical sync processing consists of:

- 625/525 line detection
- vertical noise suppression

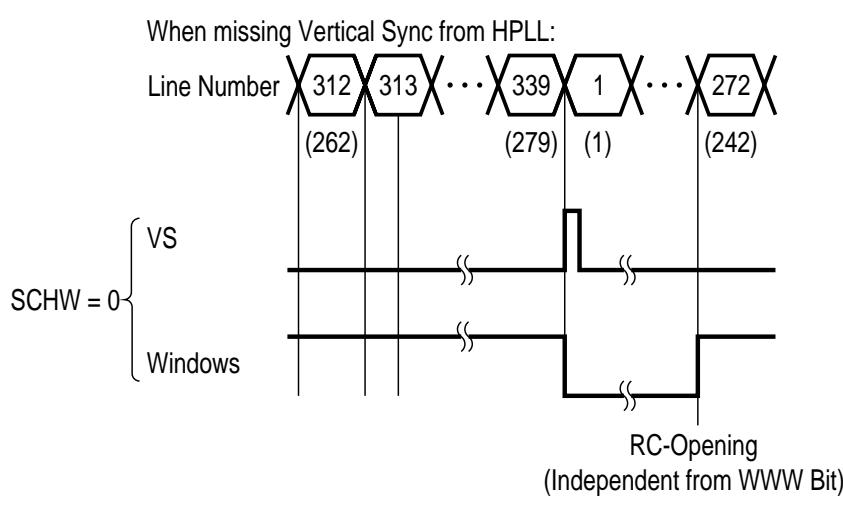
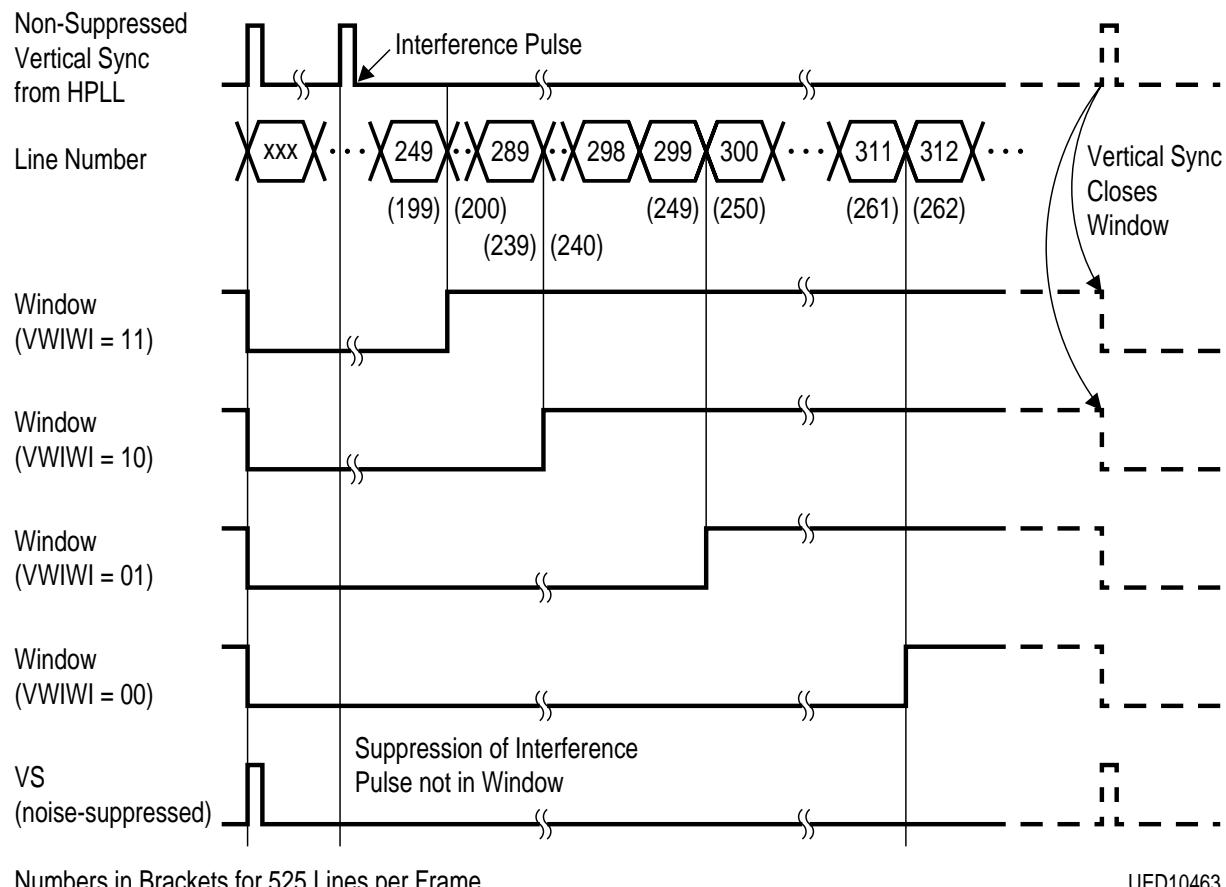
The vertical pulses are obtained from the SYNC signal by integration. The 625/525 line detector measures the number of lines per field. By taking the average of the individual measurements with two up/down counters, the status bits 'FF' ad 'FFGF' are obtained.

When vertical noise suppression is switched on ( $VOFF = 0$ ), the vertical pulse obtained from the SYNC signal by integration is admitted only within a preset window (refer to timing diagram) and appears as a VS pulse. The width of the window can be set via the I<sup>2</sup>C Bus.

In the temporary absence of vertical pulses in SYNC, a continuous VS can be generated by switching on a 'flywheel mode' ( $SCHW = 1$ ) providing a number of lines per field of 312.5 or 262.5 respectively.

When interference to SYNC is heavy, missing vertical pulses can be supplemented by switching on the flywheel mode and vertical interference can be eliminated by switching on the noise suppression circuitry. Noise suppression and the flywheel mode can be enabled independently of each other.

There is also the possibility of generating VS in the free-running mode. The VS pulses are then completely independent of the vertical sync pulse in SYNC. When  $FREE = 1$  and  $SCHW = 1$ , a VS pulse is generated every 262.5 or 312.5 lines ( $VF = 1$  or 0 respectively). When  $FREE = 1$  and  $SCHW = 0$ , a VS pulse is generated every 279 or 339 lines ( $VF = 1$  or 0 respectively). Free-running generation of VS occurs every 262 or 312 in the terminal mode ( $TERM = 1$ ).



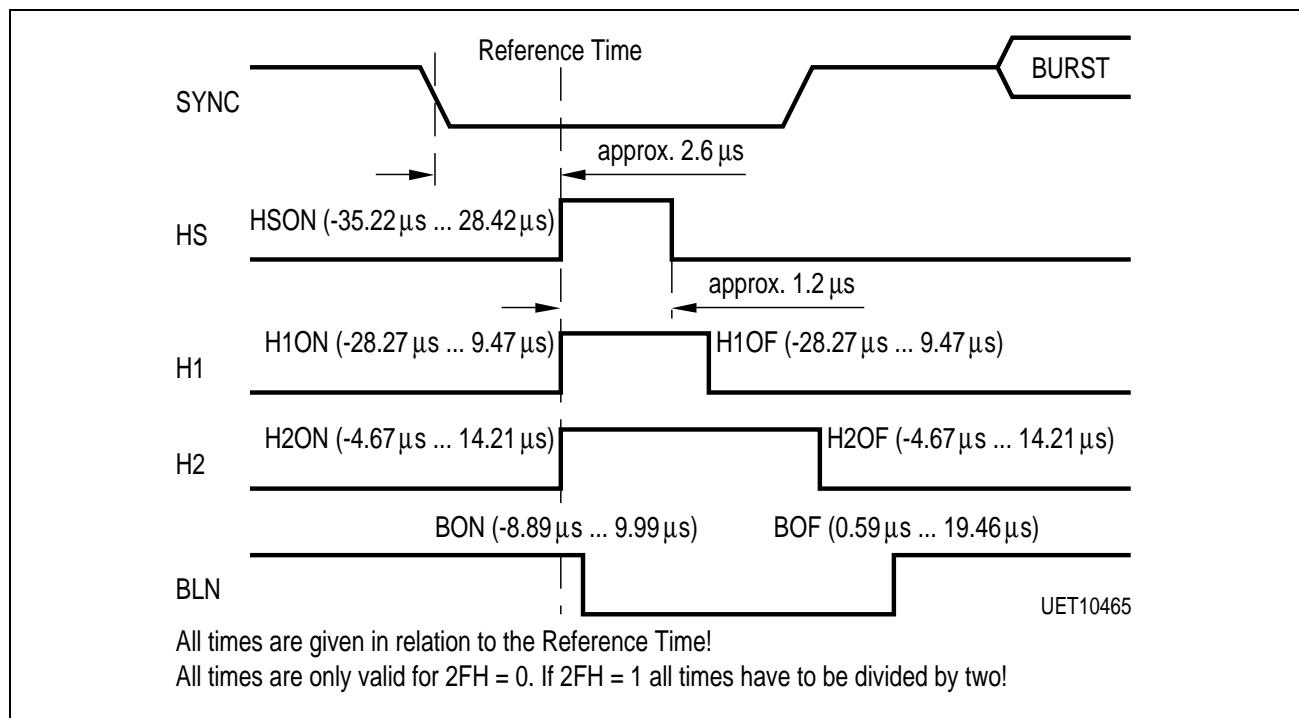
**Figure 15**  
**Window for Vertical Pulse Noise Suppression**

### 2.3.3 Pulse Generation

The clock sync generator supplies the following pulses:

- HS
- VS
- BLN
- Two clamping pulses H1 and H2. H2 is also the internal clamping pulse of the YUV-ADCs.
- The HS pulse is 16 13.5 MHz clock periods long and can be shifted by the I<sup>2</sup>C-Bus in increments of four 13.5 MHz clock periods.
- For the VS pulse refer to vertical noise suppression.
- With the BLN pulse the start time (high-to-low edge) and the stop time (low-to-high edge) can be set within a certain range of lines in increments of 13.5 MHz clock periods by I<sup>2</sup>C Bus. The timing of BLN does not change during the field blanking interval.
- During the BLN pulse the Y-U-V output data are set to their clamping level.
- For pulse H1 the start time (low-to-high edge) and stop time can be set in increments of two 13.5 MHz clock periods.
- For pulse H2 the start time (low-to-high edge) and stop time can be set in increments of 13.5 MHz clock periods.

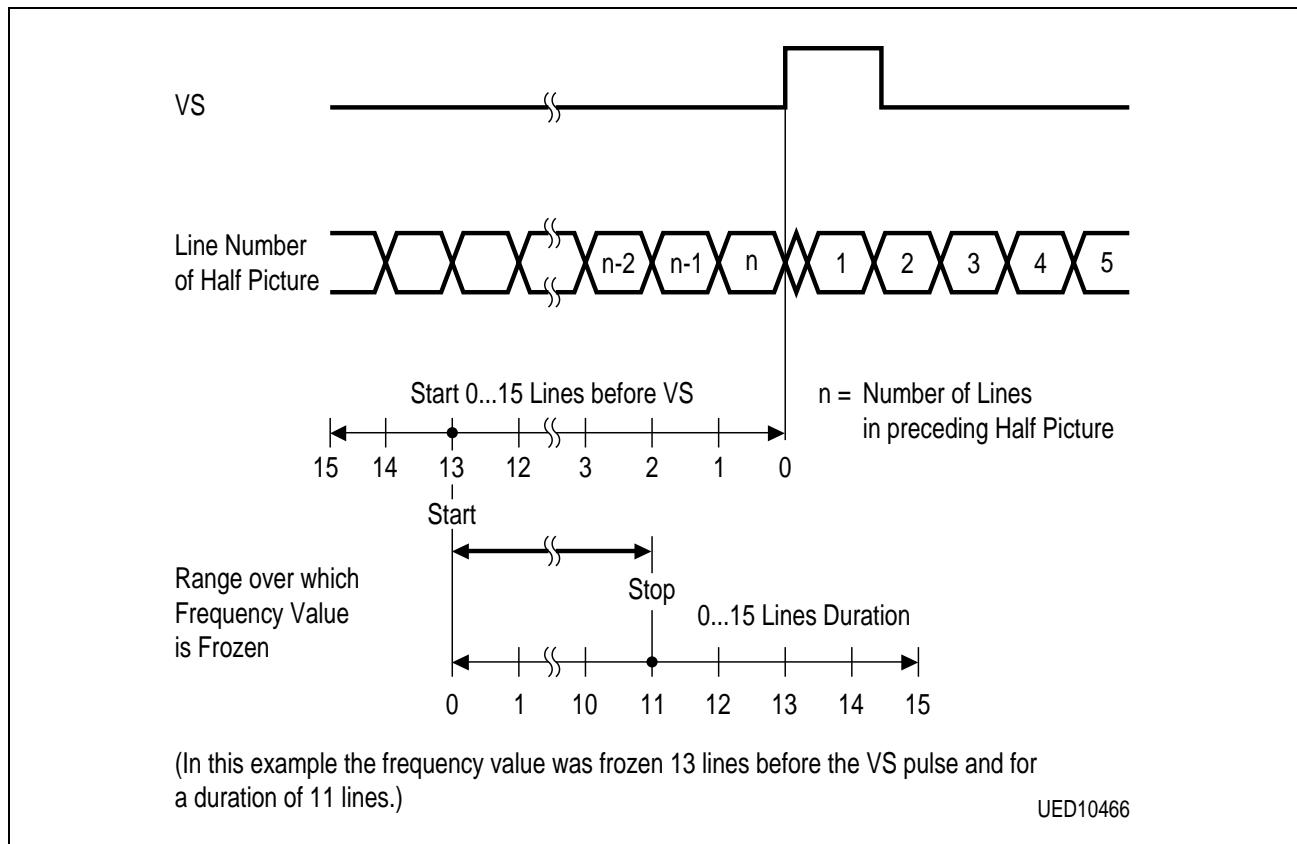
The timing of the BLN, H1, H2, VS and HS pulses can be set by the customer using the specified I<sup>2</sup>C-Bus bits. **Figure 9** shows the ranges of those settings.



**Figure 16**  
**I<sup>2</sup>C-Bus Programming Areas of Horizontal-Frequency Pulses**

### 2.3.4 Miscellaneous Circuit Sections

To suppress bottom flutter in VCR mode, the frequency of the clock can be 'hold' by 'freezing' the increment of the HPLL. The vertical-frequency 'freezing-time' starts a number of lines (programmable by the I<sup>2</sup>C Bus) before the vertical pulse and then lasts for a number (programmable) of lines. The settings do not depend on I<sup>2</sup>C-Bit TV. This functionality is only available for the 1fh mode (2FH = 0).



**Figure 17**  
**I<sup>2</sup>C-Bus Programming Area which Clock Frequency Value Generated by HPLL can be Frozen**

An active low reset signal for other chips is available at pin RESOUTN. It is activated when the chip supply voltage  $V_{DD}$  is switched on or when voltage glitches occur on it. RESOUTN also is activated by pin RESIN. The RESOUTN pulse signal is not cancelled until the crystal oscillator resonates and in addition stretched by an internal circuit for approximately 127 lines (8 ms).

## 2.4 I<sup>2</sup>C Bus

### 2.4.1 I<sup>2</sup>C-Bus Address

1	0	1	1	0	0	B	B: equal to the value set on pin ADR0				
---	---	---	---	---	---	---	---------------------------------------	--	--	--	--

### 2.4.2 I<sup>2</sup>C-Bus Format

Write:

S	1	0	1	1	0	0	B	0	A	Subaddress	A	Data Byte	A	****	A	P
---	---	---	---	---	---	---	---	---	---	------------	---	-----------	---	------	---	---

Read:

S	1	0	1	1	0	0	B	1	A	...	Status Byte 0	A	Status Byte 1	A	Data****	
***** Byte n   A   Data Byte (n+1)   A   *****   NA   P																

Reading starts with status byte 0, followed by status byte 1 and then in succession by data byte n, data byte n+1..., where n is the last write address. Specification of a subaddress in reading mode is not possible.

S: Start condition

A: Acknowledge

P: Stop condition

NA: Not Acknowledge

An automatic address increment function is implemented.

After switching on the IC or RESIN = 0, all bits are set to defined states. Particularly:

Register	Default value	Register	Default value
00 <sub>H</sub>	10 <sub>H</sub>	0C <sub>H</sub>	00 <sub>H</sub>
01 <sub>H</sub>	40 <sub>H</sub>	0D <sub>H</sub>	00 <sub>H</sub>
02 <sub>H</sub>	00 <sub>H</sub>	0E <sub>H</sub>	00 <sub>H</sub>
03 <sub>H</sub>	00 <sub>H</sub>	0F <sub>H</sub>	00 <sub>H</sub>
04 <sub>H</sub>	28 <sub>H</sub>	10 <sub>H</sub>	00 <sub>H</sub>
05 <sub>H</sub>	00 <sub>H</sub>	11 <sub>H</sub>	00 <sub>H</sub>
06 <sub>H</sub>	00 <sub>H</sub>	12 <sub>H</sub>	00 <sub>H</sub>
07 <sub>H</sub>	06 <sub>H</sub>	13 <sub>H</sub>	00 <sub>H</sub>
08 <sub>H</sub>	00 <sub>H</sub>	14 <sub>H</sub>	13 <sub>H</sub>
09 <sub>H</sub>	00 <sub>H</sub>	15 <sub>H</sub>	00 <sub>H</sub>
0A <sub>H</sub>	00 <sub>H</sub>	16 <sub>H</sub>	00 <sub>H</sub>
0B <sub>H</sub>	00 <sub>H</sub>	17 <sub>H</sub>	00 <sub>H</sub>

### 2.4.3 I<sup>2</sup>C-Bus Commands

Subadd.	Data Byte							
	D7	D6	D5	D4	D3	D2	D1	D0
00 <sub>H</sub>	0	UV3FIL	FORMAT1	FORMAT0	UVCODE	YCODE	OENB	OENA
01 <sub>H</sub>	YD3	YD2	YD1	YD0	0	0	0	0
02 <sub>H</sub>	I2	I1	SELH2I2	SELH1I1	0	0	0	0
03 <sub>H</sub>	0	0	0	0	CGSUP1	CGSUP0	VWIWI1	VWIWI0
04 <sub>H</sub>	0	OEFB	S1CL1	S1CL0	S2CL1	S2CL0	0	0
05 <sub>H</sub>	0	SCHW	HPLL	VTHRE	CLOF	0	0	0
06 <sub>H</sub>	TV	FREE	VOFF	VF	TERM	GENMOD	0	SYPOL
07 <sub>H</sub>	2FH	HSWMA	HSWMIN	INC4	INC3	INC2	INC1	INC0
08 <sub>H</sub>	BON7	BON6	BON5	BON4	BON3	BON2	BON1	BON0
09 <sub>H</sub>	BOF7	BOF6	BOF5	BOF4	BOF3	BOF2	BOF1	BOF0
0A <sub>H</sub>	H1ON7	H1ON6	H1ON5	H1ON4	H1ON3	H1ON2	H1ON1	H1ON0
0B <sub>H</sub>	H1OF7	H1OF6	H1OF5	H1OF4	H1OF3	H1OF2	H1OF1	H1OF0
0C <sub>H</sub>	H2ON7	H2ON6	H2ON5	H2ON4	H2ON3	H2ON2	H2ON1	H2ON0
0D <sub>H</sub>	H2OF7	H2OF6	H2OF5	H2OF4	H2OF3	H2OF2	H2OF1	H2OF0
0E <sub>H</sub>	HSON7	HSON6	HSON5	HSON4	HSON3	HSON2	HSON1	HSON0
0F <sub>H</sub>	0	0	0	0	0	0	0	0
10 <sub>H</sub>	FION3	FION2	FION1	FION0	FILE3	FILE2	FILE1	FILE0
11 <sub>H</sub>	SYNAMP3	SYNAMP2	SYNAMP1	SYNAMP0	YAMP3	YAMP2	YAMP1	YAMP0
12 <sub>H</sub>	UAMP3	UAMP2	UAMP1	UAMP0	VAMP3	VAMP2	VAMP1	VAMP0
13 <sub>H</sub>	DATDEL2	DATDEL1	DATDELO	DATSLOP	CLKSLOP1	CLKSLOPO	0	0
14 <sub>H</sub>	0	0	0	1	0	0	1	1
15 <sub>H</sub>	0	0	0	0	0	0	0	0
16 <sub>H</sub>	0	0	0	0	0	0	0	0
17 <sub>H</sub>	0	0	0	0	0	0	0	0

## 2.4.4 Detailed Description

### Subaddress 00<sub>H</sub>

Bit	Name	Function
D7	0	Reserved
D6	UV3FIL	Filter stage 3 for UV data (FORMAT = 0X) 0: <u>OFF</u> 1: <u>ON</u> <i>Note: For FORMAT = 1X filter stage 3 for UV data is "on"</i> <i>(UV3FIL = don't care)</i>
D5...D4	FORMAT	Selection of output data interface: 00: Output data format according CCIR 656 (8 wires at Port A) <u>01: Parallel output data format (2 x 8 wires)</u> 10: Quasiparallel 12 wire interface 11: Quasiparallel 12 wire interface
D3	UVCODE	Coding of UV data: <u>0: Straight binary code</u> 1: Two's complement code
D2	YCODE	Coding of Y data: <u>0: Straight binary code</u> 1: Two's complement code
D1	OENB	Output enable port B: <u>0: Tristate</u> 1: Port enabled
D0	OENA	Output enable port A: <u>0: Tristate</u> 1: Port enabled

**Subaddress 01<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D4	YD	Delay compensation in Y-signal path (13.5 MHz clocks): 0000: ... - 0.30 µs 0001 0010 0011 <u>0100:</u> ... 0 µs : 1110 1111: ... 0.81 µs
D3...D0	<u>0000</u>	Reserved

**Subaddress 02<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7	I2	Voltage level of H2I2 output (SELH2I2 = 1): 0: <u>Low voltage at pin H2I2</u> 1: High voltage at pin H2I2
D6	I1	Voltage level of H1I1 output (SELH1I1 = 1): 0: <u>Low voltage at pin H1I1</u> 1: High voltage at pin H1I1
D5	SELH2I2	Function of pin H2I2: 0: <u>H2 (line frequency, start and stop programmable)</u> 1: I2 (low/high programmable)
D4	SELH1I1	Function of pin H1I1: 0: <u>H1 (line frequency, start and stop programmable)</u> 1: I1 (low/high programmable)
D3	<u>0</u>	Reserved
D2	<u>0</u>	Reserved
D1	<u>0</u>	Reserved
D0	<u>0</u>	Reserved

**Subaddress 03<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D4	0000	Reserved
D3...D2	CGSUP	Suppression of black level disturbances caused by copy guarded tapes 00: <u>No function</u> 01: Black level error is limited to + / - 32 (~ 27 mV) 10: Black level error is limited to + / - 16 (~ 14 mV) 11: Black level error is limited to + / - 8 (~ 7 mV)
D1...D0	VWIWI	Width of Window in Vertical Processing: 00: <u>Narrow window: open from line 312 for PAL and 262 for NTSC</u> 01: Window: open from line 300 for PAL and 250 for NTSC 10: Window: open from line 290 for PAL and 240 for NTSC 11: Very wide window: open from line 250 for PAL and 200 for NTSC

**Subaddress 04<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7	0	Reserved
D6	OEFB	Output enable for Featurebox signals BLN, HS and VS: 0: <u>BLN, HS, VS outputs tristate</u> 1: BLN, HS, VS outputs enabled (2FH = 0) BLN, HS outputs enabled, VS output tristate (2FH = 1)
D5...D4	S1CL	Selection of clock frequency on pin CLK1: 00: Tristate 01: 6.75 MHz 10: <u>13.5 MHz</u> 11: 27 MHz  For the allowed values of S1CL refer to table <b>chapter 2.3.1!</b>
D3...D2	S2CL	Selection of clock frequency on pin CLK2: 00: Tristate 01: 6.75 MHz 10: <u>13.5 MHz</u> 11: 27 MHz  For the allowed values of S2CL refer to table <b>chapter 2.3.1!</b>
D1...D0	00	Reserved

**Subaddress 05<sub>H</sub>**

Bit	Name	Function
D7	<u>0</u>	Reserved
D6	SCHW	Mode of vertical pulse generation: 0: <u>No flywheel mode</u> 1: Flywheel mode
D5	HPLL	Relationship between horizontal frequency in SYNC and default frequency on CLK1 and CLK2: 0: <u>864</u> 1: 858
D4	VTHRE	Minimum sync pulse length from which a vertical pulse is detected: 0: <u>26.6 µs</u> 1: 11.3 µs
D3	CLOF	Clamping of SYNC for clock generator: 0: <u>Clamping on</u> 1: Clamping off
D2...D0	<u>000</u>	Reserved

**Subaddress 06<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>																														
D7	TV	Selection of HPLL lock-in behavior: 0: <u>Optimum for VCR</u> 1: Optimum for SYNC from network																														
D6	FREE	Generation of V pulse: 0: <u>V derived from SYNC</u> 1: Free-running generation; vertical frequency is determined by VF bit, VOFF bit is enabled, SCHW bit should be set to 1																														
D5	VOFF	Vertical noise suppression: 0: <u>Noise suppression enabled</u> 1: No noise suppression																														
D4	VF	Number of lines per field: 0: <u>312.5 or 312</u> 1: 262.5 or 262  <b>Note:</b> VF must be set to the number of lines present in SYNC for fly-wheel and noise suppression modes. VF determines the number of lines per field for the free-running or terminal mode.																														
D3	TERM	Terminal mode: FREE TERM SCHW VF Number of Lines per Field generated in Free-Running Mode  <table> <tbody> <tr> <td>don't care</td> <td>1</td> <td>don't care</td> <td>0</td> <td>312</td> </tr> <tr> <td>don't care</td> <td>1</td> <td>don't care</td> <td>1</td> <td>262</td> </tr> <tr> <td>1</td> <td><u>0</u></td> <td>1</td> <td>0</td> <td>312.5</td> </tr> <tr> <td>1</td> <td><u>0</u></td> <td>1</td> <td>1</td> <td>262.5</td> </tr> <tr> <td>1</td> <td><u>0</u></td> <td>0</td> <td>0</td> <td>339</td> </tr> <tr> <td>1</td> <td><u>0</u></td> <td>0</td> <td>1</td> <td>279</td> </tr> </tbody> </table>	don't care	1	don't care	0	312	don't care	1	don't care	1	262	1	<u>0</u>	1	0	312.5	1	<u>0</u>	1	1	262.5	1	<u>0</u>	0	0	339	1	<u>0</u>	0	1	279
don't care	1	don't care	0	312																												
don't care	1	don't care	1	262																												
1	<u>0</u>	1	0	312.5																												
1	<u>0</u>	1	1	262.5																												
1	<u>0</u>	0	0	339																												
1	<u>0</u>	0	1	279																												
D2	GENMOD	Clock generator mode 0: <u>Normal PLL mode</u> 1: Generator mode (fixed frequency output, controlled by INC)																														
D1	0	Reserved																														
D0	SYPOL	SYNC polarity: 0: <u>Negative sync signals</u> (normal SYNC input) 1: Positive sync signals																														

**Subaddress 07<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7	2FH	Selection of input frequency range: 0: <u>Normal line frequencies (around 15.6 kHz)</u> 1: Double line frequencies (31.2...38 kHz) [YUV A/D converters are switched off]
D6	HSWMA	Maximum width of HSYNC (input SYNC): 0: <u>6.2 µs for low FH-range</u> 3.1 µs for high FH-range (2FH = 1) 1: 9.0 µs for low FH-range 4.5 µs for high FH-range (2FH = 1)
D5	HSWMI	Minimum width of HSYNC (input SYNC): 0: <u>3.0 µs for low FH-range</u> 1.5 µs for high FH-range (2FH = 1) 1: 1.7 µs for low FH-range 0.8 µs for high FH-range (2FH = 1)
D4...D0	INC	Nominal PLL output frequency: INC = <u>00110</u> For the allowed values of INC refer to table <b>chapter 2.3.1!</b> Calculation of INC for low FH range:  $\text{INC} = \text{INT}\left(\frac{f_h}{f_q} * 110592 - 64,625\right)$  for high FH range (2FH = 1):  $\text{INC} = \text{INT}\left(\frac{f_h}{f_q} * 55292 - 64,625\right)$

**Subaddress 08<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	BON	BLN start time in relation reference time (refer to the following table and to timing diagram)

<b>BON7...BON0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
1000 0000	- (- 128) + 7	= 135	9.99 µs
...	...	...	...
1111 1111	- (- 1) + 7	= 8	0.60 µs
<u>0000 0000</u>	<u>- (0) + 7</u>	<u>= 7</u>	<u>0.52 µs</u>
0000 0001	- (+ 1) + 7	= 6	0.44 µs
...	...	...	...
0111 1111	- (+ 127) + 7	= - 120	- 8.89 µs

**Subaddress 09<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	BOF	BLN stop time in relation to reference time: (refer to the following table and to timing diagram)

<b>BOF7...BOF0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
<u>0000 0000</u>	<u>(0) + 8</u>	<u>= 8</u>	<u>0.59 µs</u>
0000 0001	(+ 1) + 8	= 9	0.67 µs
...	...	...	...
0111 1111	(+ 127) + 8	= 135	9.99 µs
1000 0000	(+ 128) + 8	= 136	10.06 µs
1000 0001	(+ 129) + 8	= 137	10.14 µs
...	...	...	...
1111 1110	(+ 254) + 8	= 262	19.39 µs
1111 1111	(+ 255) + 8	= 263	19.46 µs

**Subaddress 0A<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	H1ON	H1 start time in relation to reference time: (refer to the following table and to timing diagram)

<b>H1ON7...H1ON0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
1100 0000	- (- 64) x 2	= 128	9.47 µs
...	...	...	...
1111 1111	- (- 1) x 2	= 2	0.15 µs
0000 0000	<u>- (0) x 2</u>	<u>= 0</u>	<u>0 µs</u>
0000 0001	- (+ 1) x 2	= - 2	- 0.15 µs
...	...	...	...
0111 1111	- (+ 127) x 2	= - 254	- 18.79 µs
1000 0000	- (+ 128) x 2	= - 256	- 18.94 µs
1000 0001	- (+ 129) x 2	= - 258	- 19.09 µs
...	...	...	...
1011 1111	- (+ 191) x 2	= - 382	- 28.27 µs

**Subaddress 0B<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	H1OF	H1 stop time in relation to reference time: (refer to the following table and to timing diagram)

<b>H1OF7...H1OF0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
1100 0000	- (- 64) x 2	= 128	9.47 µs
...	...	...	...
1111 1111	- (- 1) x 2	= 2	0.15 µs
0000 0000	<u>- (0) x 2</u>	<u>= 0</u>	<u>0 µs</u>
0000 0001	- (+ 1) x 2	= - 2	- 0.15 µs
...	...	...	...
0111 1111	- (+ 127) x 2	= - 254	- 18.79 µs
1000 0000	- (+ 128) x 2	= - 256	- 18.94 µs
1000 0001	- (+ 129) x 2	= - 258	- 19.09 µs
...	...	...	...
1011 1111	- (+ 191) x 2	= - 382	- 28.27 µs

**Subaddress 0C<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	H2ON	H2 start time in relation to reference time: (H2 is always used as clamping reference for the YUV ADCs) (refer to the following table and to timing diagram)

<b>H2ON7...H2ON0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
0100 0000	- (- 192)	= 192	14.21 µs
...	...	...	...
0111 1111	- (- 129)	= 129	9.55 µs
1000 0000	- (- 128)	= 128	9.47 µs
1000 0001	- (- 127)	= 127	9.40 µs
...	...	...	...
1111 1111	- (- 1)	= 1	0.07 µs
0000 0000	- (0)	= 0	0 µs
0000 0001	- (+ 1)	= - 1	- 0.07 µs
...	...	...	...
0011 1111	- (+ 63)	= - 63	- 4.67 µs

**Subaddress 0D<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	H2OF	H2 stop time in relation to reference time: (H2 is always used as clamping reference for the YUV ADCs) (refer to the following table and to timing diagram)

<b>H2OF7...H2OF0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
0100 0000	- (- 192)	= 192	14.21 µs
...	...	...	...
0111 1111	- (- 129)	= 129	9.55 µs
1000 0000	- (- 128)	= 128	9.47 µs
1000 0001	- (- 127)	= 127	9.40 µs
...	...	...	...
1111 1111	- (- 1)	= 1	0.07 µs
0000 0000	- (0)	= 0	0 µs
0000 0001	- (+ 1)	= - 1	- 0.07 µs
...	...	...	...
0011 1111	- (+ 63)	= - 63	- 4.67 µs

**Subaddress 0E<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	HSON	HS start time in relation to reference time: (refer to the following table and to timing diagram)

<b>HSON7...HSON0</b>	<b>Number</b>	<b>13.5 MHz Cycles</b>	<b>Time (2FH = 0)</b>
1010 0000	- (- 96) x 4	= 384	28.42 µs
...	...	...	...
1111 1111	- (- 1) x 4	= 4	0.30 µs
0000 0000	- (0) x 4	= 0	0 µs
0000 0001	- (+ 1) x 4	= - 4	- 0.30 µs
...	...	...	...
0111 0110	- (+ 118) x 4	= - 472	- 34.93 µs
0111 0111	- (+ 119) x 4	= - 476	- 35.22 µs

**Subaddress 0F<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	0000 0000	Reserved

**Subaddress 10<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D4	FION	Start of clock frequency freezing in number of lines before the vertical pulse (only valid for 2FH = 0): <u>0000</u> : 0 (no freezing) 0001: 1 : 1111: 15
D3...D0	FILE	Duration of clock frequency freezing in number of lines: <u>0000</u> : 0 (no freezing) 0001: 1 : 1111: 15

**Subaddress 11<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D4	SYNAMP	Internal amplification of SYNC input signal. Allowed values: SYNAMP = <u>0000</u> : amplification 0 dB : SYNC input nom. 2 Vpp SYNAMP = 0110 : amplification 6 dB : SYNC input nom. 1 Vpp
D3...D0	YAMP	Internal amplification of AINY input signal. Allowed values: YAMP = <u>0000</u> : amplification 0 dB : AINY input nom. 2 Vpp YAMP = 0110 : amplification 6 dB : AINY input nom. 1 Vpp

**Subaddress 12<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D4	UAMP	<p>Internal amplification of AINU input signal.</p> <p>Allowed values:</p> <p>UAMP = <u>0000</u> : internal amplification 0 dB : AINU input nom. 2 Vpp</p> <p>UAMP = 0110 : internal amplification 6 dB : AINU input nom. 1 Vpp</p>
D3...D0	VAMP	<p>Internal amplification of AINV input signal.</p> <p>Allowed values:</p> <p>VAMP = <u>0000</u> : amplification 0 dB : AINV input nom. 2 Vpp</p> <p>VAMP = 0110 : internal amplification 6 dB : AINV input nom. 1 Vpp</p>

**Subaddress 13<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D5	DATDEL	<p>Programmable output delay for PAQ7...PAQ0, PBQ7...PBQ0, BLN, HS, H1I1, H2I2, VS.</p> <p>Allowed values:</p> <p><u>000</u></p> <p>001 (description <b>see chapter 5.3</b>)</p>
D4	DATSLOP	<p>Adaptation of the output driver stages for PAQ7...PAQ0, PBQ7...PBQ0, BLN, HS, H1I1, H2I2, VS.</p> <p>Allowed values:</p> <p><u>0</u> ... to be used only for 5 V output stage supply voltage</p> <p>1 ... to be used only for 3.3 V output stage supply voltage and FORMAT = 00</p>
D3...D2	CLKSLOP	<p>Adaptation of the output driver stages for CLK1 and CLK2.</p> <p>Allowed values:</p> <p><u>00</u> ... to be used only for 5 V output stage supply voltage</p> <p>10 ... to be used only for 3.3 V output stage supply voltage</p>
D1...D0	<u>00</u>	Reserved

**Subaddress 14<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	0001 0011	Reserved

**Subaddress 15<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	0000 0000	Reserved

**Subaddress 16<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	0000 0000	Reserved

**Subaddress 17<sub>H</sub>**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D0	0000 0000	Reserved

## 2.4.5 Read Mode

### Status Byte 0

Bit	Name	Function																																																
D7	CON	Absolute difference between the horizontal sync pulse in SYNC and the HPLL: 0: Larger than or equal to 32 system clock cycles 1: Less than 32 system clock cycles																																																
D6	THRELIM	Absolute difference between the horizontal sync pulse in SYNC and the HPLL: 0: Larger than 8 system clock cycles 1: Less than 8 system clock cycles for 8 or more successive lines (i.e. HPLL well locked in)																																																
D5, D4	FFGF, FF	Identified number of lines per field (refer also to timing diagram figure 16): <table> <tr> <td>&lt; N1</td> <td>0</td> <td>0</td> </tr> <tr> <td><math>\geq N1</math> and <math>\leq N2</math></td> <td>1</td> <td>0</td> </tr> <tr> <td><math>&gt; N2</math> and <math>&lt; 287</math></td> <td>0</td> <td>0</td> </tr> <tr> <td><math>\geq 287</math> and <math>&lt; N3</math></td> <td>0</td> <td>1</td> </tr> <tr> <td><math>\geq N3</math> and <math>\leq N4</math></td> <td>1</td> <td>1</td> </tr> <tr> <td><math>&gt; N4</math></td> <td>0</td> <td>1</td> </tr> </table> N1 to N4 depends on Control Bits VWIWI: <table> <thead> <tr> <th>VWIWI1</th> <th>VWIWI0</th> <th>N1</th> <th>N2</th> <th>N3</th> <th>N4</th> </tr> </thead> <tbody> <tr> <td>0</td> <td>0</td> <td>262</td> <td>264</td> <td>312</td> <td>314</td> </tr> <tr> <td>0</td> <td>1</td> <td>250</td> <td>275</td> <td>300</td> <td>325</td> </tr> <tr> <td>1</td> <td>0</td> <td>240</td> <td>285</td> <td>290</td> <td>335</td> </tr> <tr> <td>1</td> <td>1</td> <td>200</td> <td>312</td> <td>250</td> <td>362</td> </tr> </tbody> </table>	< N1	0	0	$\geq N1$ and $\leq N2$	1	0	$> N2$ and $< 287$	0	0	$\geq 287$ and $< N3$	0	1	$\geq N3$ and $\leq N4$	1	1	$> N4$	0	1	VWIWI1	VWIWI0	N1	N2	N3	N4	0	0	262	264	312	314	0	1	250	275	300	325	1	0	240	285	290	335	1	1	200	312	250	362
< N1	0	0																																																
$\geq N1$ and $\leq N2$	1	0																																																
$> N2$ and $< 287$	0	0																																																
$\geq 287$ and $< N3$	0	1																																																
$\geq N3$ and $\leq N4$	1	1																																																
$> N4$	0	1																																																
VWIWI1	VWIWI0	N1	N2	N3	N4																																													
0	0	262	264	312	314																																													
0	1	250	275	300	325																																													
1	0	240	285	290	335																																													
1	1	200	312	250	362																																													
D3		don't care																																																
D2 ... D0	POR	Status bit POR is set by power on reset or by activating the reset pin. POR is reset after reading the status byte.																																																

**Status Byte 1**

<b>Bit</b>	<b>Name</b>	<b>Function</b>
D7...D6		don't care
D5	OFLY	Overflow detection of ADC for input AINY
D4	UFLY	Underflow detection of ADC for input AINY
D3	OFLU	Overflow detection of ADC for input AINU
D2	UFLU	Underflow detection of ADC for input AINU
D1	OFLV	Overflow detection of ADC for input AINV
D0	UFLV	Underflow detection of ADC for input AINV

### 3 Absolute Maximum Ratings

Parameter	Symbol	Limit Values		Unit	Remark
		min.	max.		
Storage temperature	$T_{\text{stg}}$	- 40	125	°C	
Soldering temperature	$T_{\text{sold}}$		260	°C	
Soldering time	$t_{\text{sold}}$		10	sec	
Input/output voltage	$V_{I/Q}$	$V_{SSQ} - 0.3 \text{ V}$	$V_{DDQ} + 0.3 \text{ V}$	1	Not valid for I <sup>2</sup> C-Bus pins
Input/output voltage I <sup>2</sup> C-pins 33, 34 (SCL, SDA)	$V_{I/Q, I2C}$	$V_{SSQ} - 0.3 \text{ V}$	6 V	1	
Power supply voltage	$V_{DD}, V_{ADDx}, V_{DDQ}, V_{DDDTO}$	- 0.3	6	V	
Total power dissipation	$P_{\text{tot}}$		1.25	W	
Latch-up protection		- 100	100	mA	All inputs/outputs
ESD protection	ESD	- 1	1	kV	MIL STD 883C method 3015-6, 100 pF, 1500 Ω

All voltages listed are referenced to ground (0 V,  $V_{SS}$ ) except where noted.

**Note:** Absolute maximum ratings are those values beyond which damage to the device may occur. Functional operation under these conditions or at any other condition beyond those indicated in the operational sections of this specification is not implied.

### 3.1 Recommended Operating Conditions

Parameter	Symbol	Limit Values			Unit	Remark
		min.	nom.	max.		
Ambient temperature	$T_A$	0	25	70	°C	

#### Power Requirements

Analog supply voltage	$V_{ADDx}$	4.75	5.0	5.25	V	
Digital supply voltage	$V_{DD}$	4.75	5.0	5.25	V	
DTO supply voltage	$V_{DDDTO}$	4.75	5.0	5.25	V	
Output stage supply voltage	$V_{DDQ}$	4.75	5.0	5.25	V	5 V-Mode
		3.0	3.3	3.6	V	3.3 V mode, only to be used for FORMAT = 00
Supply voltage differential	$V_{DD}$ , diff	- 0.25		0.25	V	Supply pins $V_{ADDx}$

#### All TTL Inputs

H-input voltage	$V_{IH}$	2.0V		$V_{DDQ}$	1	
L-input voltage	$V_{IL}$	0		0.8	V	

#### I<sup>2</sup>C-Bus (Values are Referred to min. ( $V_{IH}$ ) and max. ( $V_{IL}$ ))

H-input voltage	$V_{IH}$	0.7 x $V_{DDQ}$		$V_{DD}$	1	
L-input voltage	$V_{IL}$	0V		0.3 x $V_{DDQ}$	1	
SCL clock frequency	$f_{SCL}$	0		400	kHz	
Rise times of SCL, SDA	$t_R$			0.3	μs	
Fall times of SCL, SDA	$t_F$			0.3	μs	
Set-up time data	$t_{SU, Dat}$	100			ns	
Hold time data	$t_{HD, Dat}$	0			ns	
Bus free time before start condition	$t_{Buf}$	1.3			μs	
Set-up time start condition	$t_{SU, Sta}$	0.6			μs	
Hold time start condition	$t_{HD, Sta}$	0.6			μs	

### 3.1 Recommended Operating Conditions (cont'd)

Parameter	Symbol	Limit Values			Unit	Remark
		min.	nom.	max.		
SCL low time	$t_{\text{Low}}$	1.3			μs	
SCL high time	$t_{\text{High}}$	0.6			μs	
Load capacitance			400	pF		

#### Reference Inputs for Analog Inputs AINY, AINU, AINV, SYNC

Reference voltage high	$V_{\text{REFHx}}$	3.2	4.2	4.7	V	$V_{\text{ADDx}} = 5 \text{ V}$
Reference voltage low	$V_{\text{REFLx}}$	1.7	2.2	3.2	V	

#### Analog Inputs AINY, AINU, AINV

Input range (Peak-Peak)	$V_{\text{IPP}}$		2 V	$V_{\text{REFHx}} - V_{\text{REFLx}}$	1	YAMP, UAMP, VAMP = 0000, Prefiltering <b>see chapter 2.1.2</b>
Required ext clamp capacitance	$C_{\text{ext cl}}$		100		nF	AINY, AINU, AINV each
Required signal source resistance	$R_S$	0		200	Ω	

#### SYNC Input for Sync and Clock Generation

Input range (Peak-Peak)	$V_{\text{IPP}}$	0.5 V	2 V	$V_{\text{REFHC}} - V_{\text{REFLC}}$	1	SYNAMP = 0000
Input frequency	f	0		12	MHz	To avoid aliasing
Required ext clamp capacitance	$C_{\text{ext cl}}$		100		nF	
Required signal source resistance	$R_S$			200	Ω	

#### Inputs Crystal Connections X1, X2

Crystal frequency	$f_c$		24.576		MHz	
-------------------	-------	--	--------	--	-----	--

#### Crystal Type Fundamental Crystal

Equivalent parallel C	$C_O$		3.6		pF	
Crystal resonant impedance	$Z_R$			40	Ω	
Pin capacitance	$C_I$			10	pF	
External capacitance	$C_{\text{ext}}$		18		pF	Each

### 3.2 Characteristics (Assuming Recommended Operating Conditions)

Parameter	Symbol	Limit Values			Unit	Remark
		min.	nom.	max.		
<b>Supply Currents</b>						
Analog supply current	$I_{ADD}$		120		mA	Sum of all $V_{ADDx}$ pins
Digital supply current	$I_{DD}$		40		mA	Sum of all $V_{DD}$ pins + $V_{DDDT0}$
Output stage supply current	$I_{DDQ}$		40		mA	Sum of all $V_{DDQ}$ pins
<b>Reference Inputs for Analog Inputs AINY, AINU, AINV</b>						
Reference ladder resistance	$R_{REF}$	175	250	325	$\Omega$	For each converter between REFH and REFL
<b>Reference Inputs for Analog Input SYNC</b>						
Reference ladder resistance	$R_{REF}$	280	400	520	$\Omega$	For each converter between REFH and REFL
<b>All TTL Inputs</b>						
Input current		- 300		300	$\mu A$	$V_I = 0 \text{ V} \dots V_{DDQ}$ <b>Note:</b> internal pullup/pulldown-circuits
<b>I<sup>2</sup>C Input/Output SDA</b>						
L-output voltage	$V_{QL}$		0.6	V		$I = 4 \text{ mA}$
<b>Analog Inputs</b>						
Analog input leakage current	$I_{AIN}$	- 100		100	nA	AINY, AINU, AINV, SYNC each
Analog input capacitance	$C_I$		10	pF		AINY, AINU, AINV, SYNC each
<b>TTL Outputs Port A, Port B, VS, HS, BLN, H1, H2, RESOUTN (<math>V_{DDQ} = 3.3 \text{ V or } 5 \text{ V}</math>)</b>						
L-output voltage	$V_{QL}$	0		0.4	V	$I = 1 \text{ mA}$
H-output voltage	$V_{QH}$	2.4 V		$V_{DDQ}$	1	$I = - 0.5 \text{ mA}$
High impedance state output current	$I_{QZ}$	- 20		20	$\mu A$	$V_Q = 0 \text{ V} \dots V_{DDQ}$ Port A, Port B, VS, HS, BLN
Load capacitance	$C_L$		25	pF		

### 3.2 Characteristics (Assuming Recommended Operating Conditions) (cont'd)

Parameter	Symbol	Limit Values			Unit	Remark
		min.	nom.	max.		
Output data delay time, referenced to CLK1 (not valid for RESOUTN)	$t_{QD}$			25	ns	DATDEL = 000 $C_L = 15 \text{ pF}$ 5 V output stage supply voltage, DATSLOP = 0
				35	ns	DATDEL = 000 $C_L = 25 \text{ pF}$ 5 V output stage supply voltage, DATSLOP = 0
				25	ns	DATDEL = 000 $C_L = 25 \text{ pF}$ 3.3 V output stage supply voltage, FORMAT = 00 DATSLOP = 1
Output data hold time, referenced to CLK1 (not valid for RESOUTN)	$t_{QH}$	6			ns	
Pin RESOUTN Data delay/ data hold time		-	-	-		Asynchronous output signal

#### Clock TTL Outputs CLK1, CLK2

L-output voltage	$V_{QL}$	0		0.4	V	$I = 1 \text{ mA}$
H-output voltage	$V_{QH}$	2.4 V		$V_{DD}$	1	$I = -0.5 \text{ mA}$
Load capacitance	$C_L$			30	pF	
Transition times	$t_R, t_F$			5	ns	5 V output stage supply voltage, CLKSLOP = 00
						3.3 V output stage supply voltage, CLKSLOP = 10

### 3.2 Characteristics (Assuming Recommended Operating Conditions) (cont'd)

Parameter	Symbol	Limit Values			Unit	Remark
		min.	nom.	max.		
Low time 13.5 MHz	$t_{WL13}$	26			ns	13.5 MHz
High time 13.5 MHz	$t_{WH13}$	26			ns	13.5 MHz
Low time 27 MHz	$t_{WL27}$	10			ns	27 MHz
High time 27 MHz	$t_{WH27}$	10			ns	27 MHz
Skew	$t_{SK}$	- 2	0	2	ns	$C_{L,CLK1} = C_{L,CLK2}$
Frequency range when PLL is locked at SYNC input signal	$f$	25	27	35	MHz	$\pm 4.8\%$ at 27 MHz S1CL = 11, S2CL = 11

#### Performance of A/D Conversion (8-Bit)

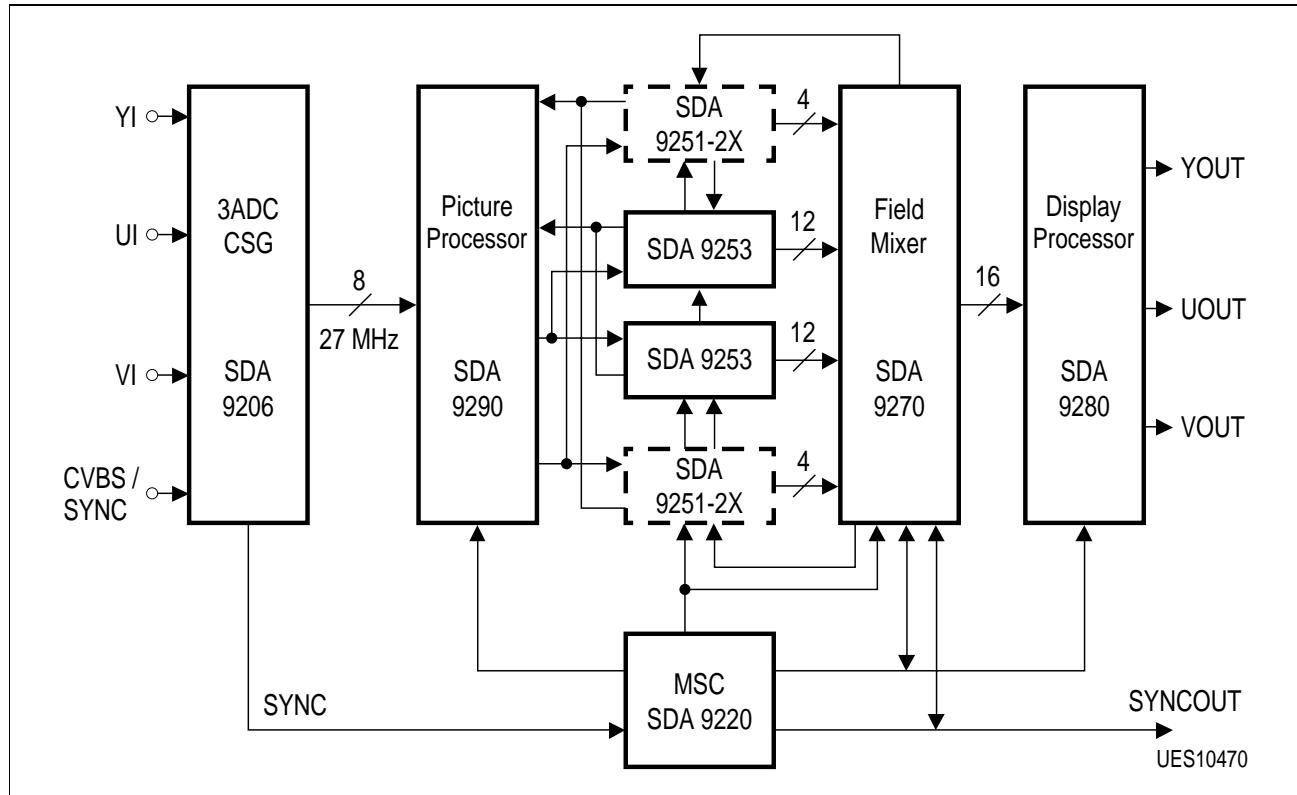
Test Conditions: ADC Clock = 27 MHz, DATDEL = 000, xAMP = 0000, VIPP = 2 Vpp

Sampling rate		27		MHz		
Differential linearity (DC)	DNLE			$\pm 0.5$	LSB	
Integral linearity (DC)	INLE			$\pm 1$	LSB	
Clamping level accuracy	CLA		0.5	$\pm 3$	LSB	
Gain error (DC)	GE			$\pm 6$	LSB	
Gain matching error (DC)	GME			$\pm 3$	LSB	
Differential gain	DG		3	%	Not tested	
Differential phase	DP		3	deg	Not tested	
Signal to noise ratio at 4.4 MHz sinus	$\alpha_{S/N}$	45	48		dB	Without harmonics

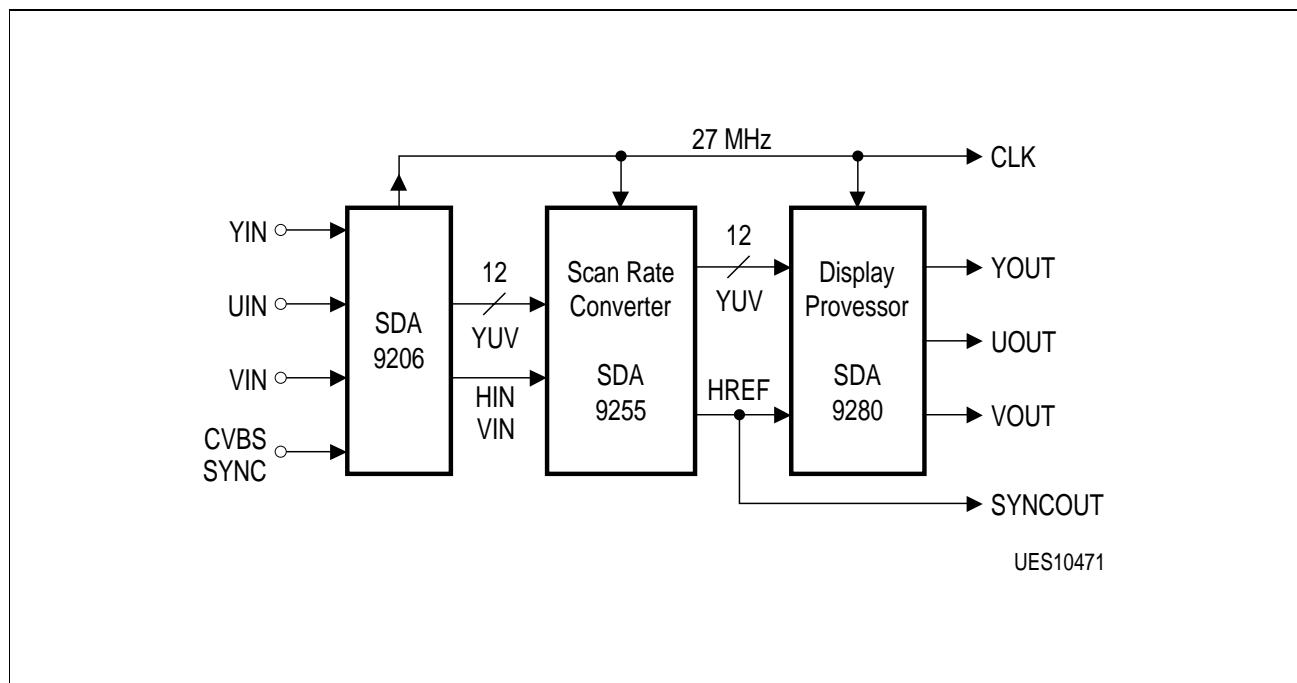
#### Harmonic Distortion

2./4. order				- 42	dB	4.4 MHz fundamental
3. order				- 42	dB	4.4 MHz fundamental
5./6. order				- 48	dB	4.4 MHz fundamental

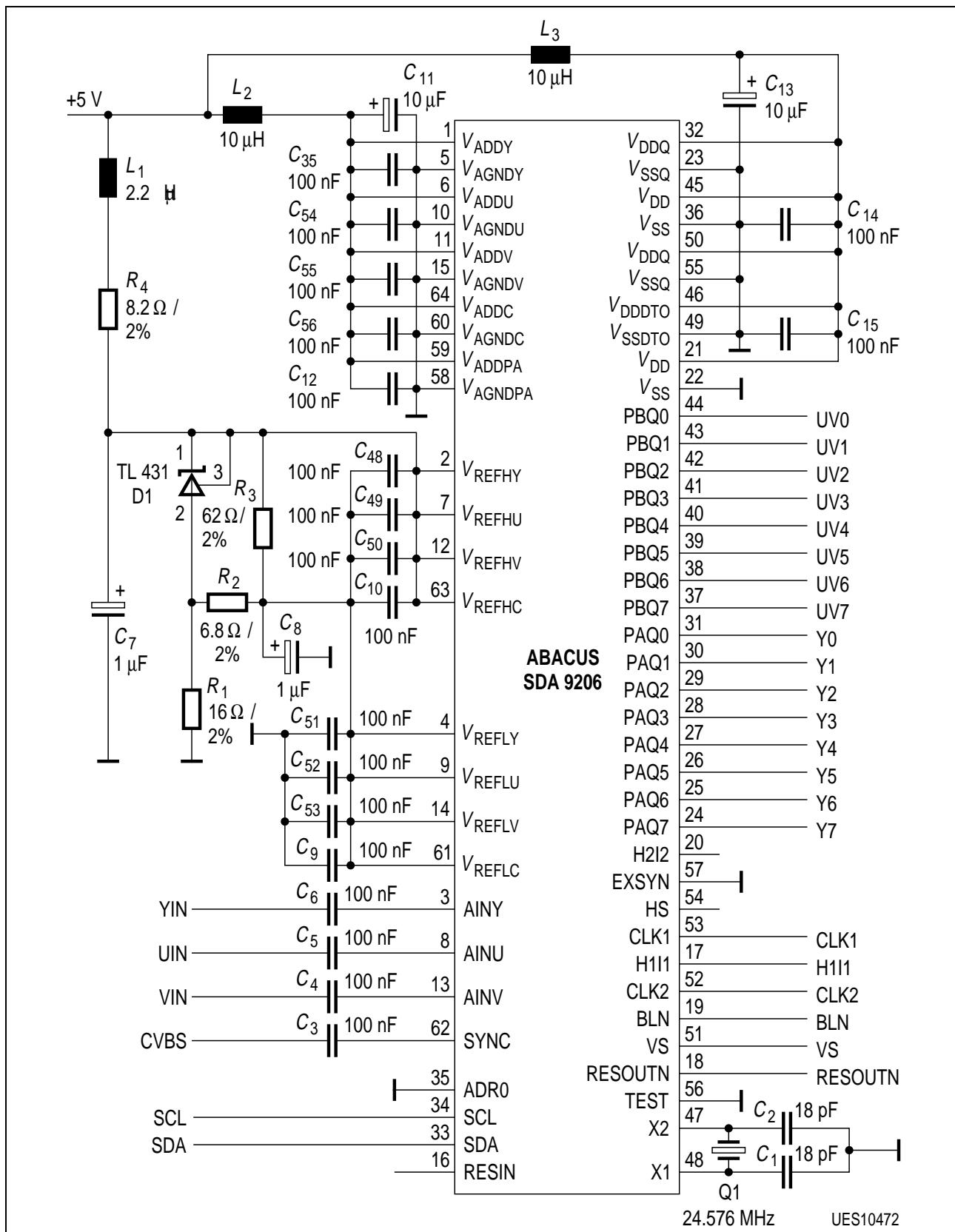
#### 4 Application Information



**Figure 18**  
**Application Circuit 1**



**Figure 19**  
**Application Circuit 2**



**Figure 20**  
**Application Circuit 3**

## 5 Waveforms

### 5.1 Timing Diagram Data Input/Output Referenced to the Clock CLK1

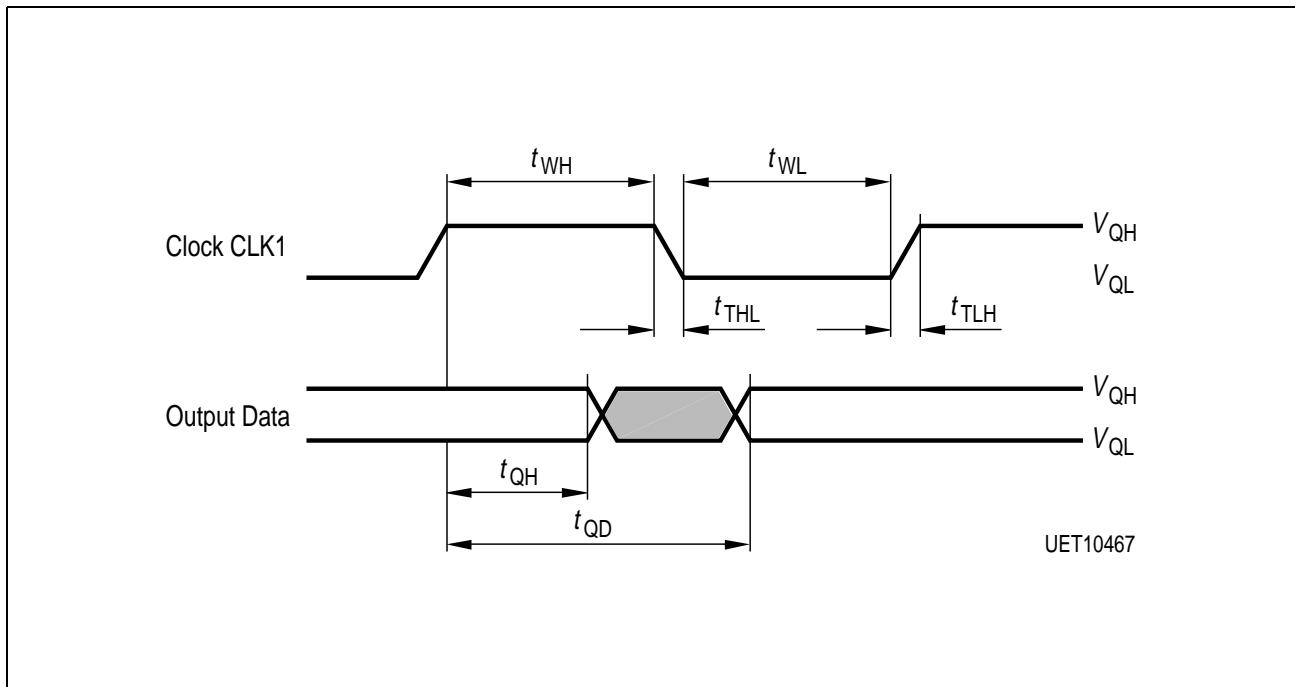


Figure 21

### 5.2 Timing Diagram Clock Skew CLK2 - CLK1

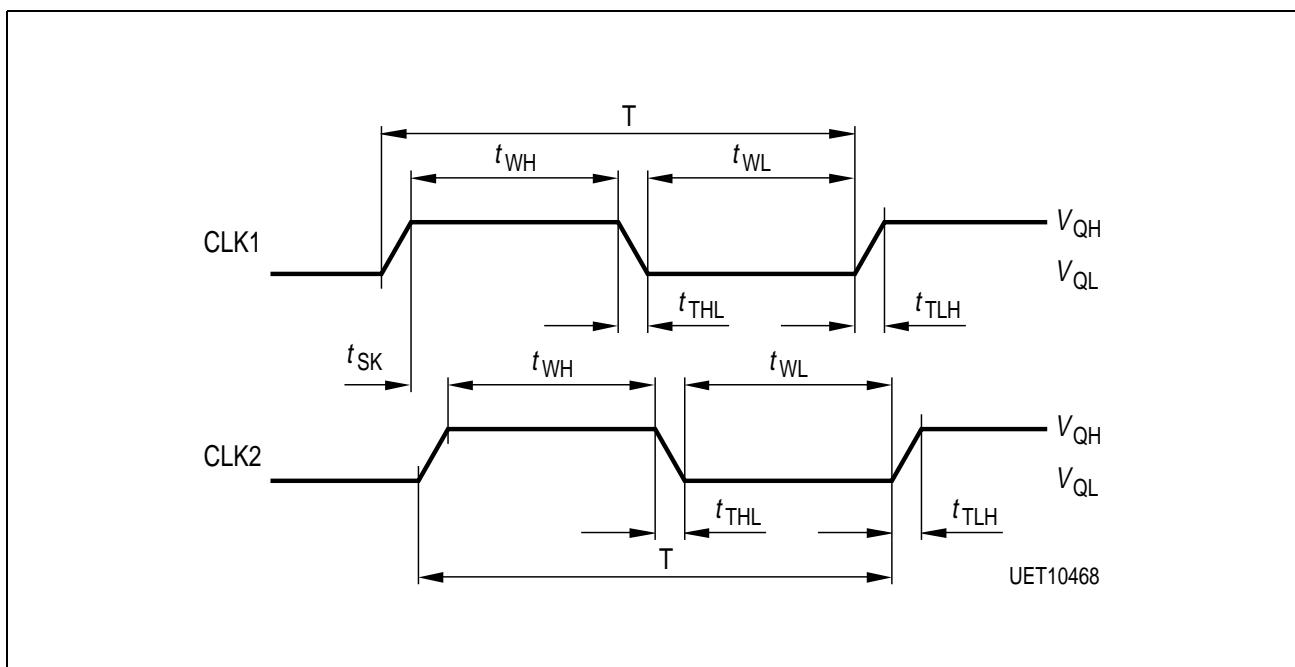


Figure 22

### 5.3 Programmable Data Output Delay: DAT\_OUT: Pins PAQ7...0, PBQ7...0, BLN, HS, H1I1, H2I2 and VS

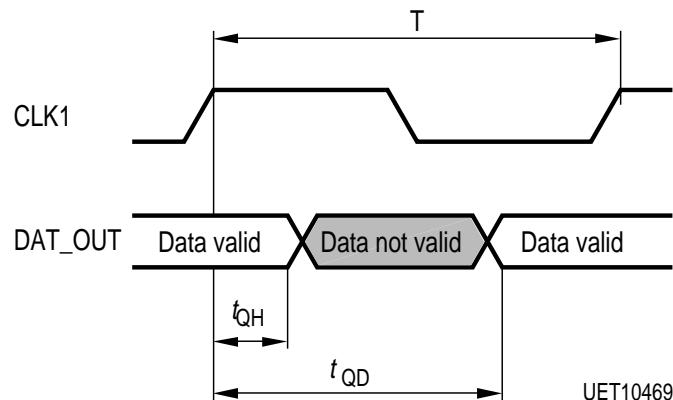


Figure 23

DATDEL	$t_{QH}$ ; min.	$t_{QD}$ ; max.
000	6 ns	25 ns
001	10 ns	29 ns

The delay times are valid for a clock rate of the analog PLL of 27 MHz.

## 6 Package Outlines

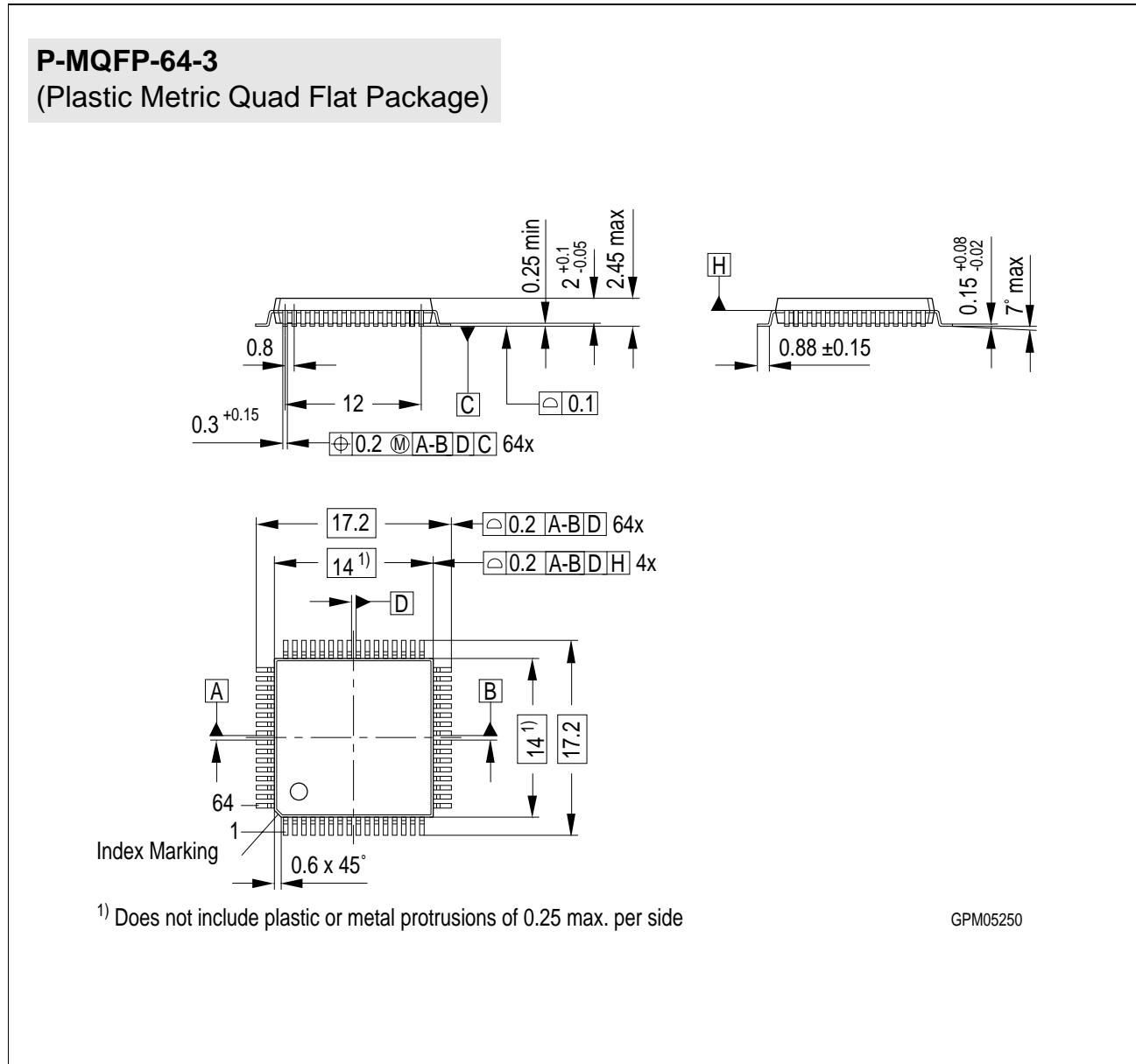


Figure 24

### Sorts of Packing

Package outlines for tubes, trays etc. are contained in our Data Book "Package Information".

SMD = Surface Mounted Device

Dimensions in mm