

**1A, 55V, 0.750 Ohm, Voltage Clamping,
Current Limited, N-Channel Power
MOSFET**

The RLP1N06CLE is an intelligent monolithic power circuit which incorporates a lateral bipolar transistor, resistors, zener diodes, and a PowerMOS transistor. The current limiting of this device allows it to be used safely in circuits where it is anticipated that a shorted load condition may be encountered. The drain to source voltage clamping offers precision control of the circuit voltage when switching inductive loads. Logic level gates allow this device to be fully biased on with only 5V from gate to source. Input protection is provided for ESD up to 2kV.

Formerly developmental type TA09880.

Ordering Information

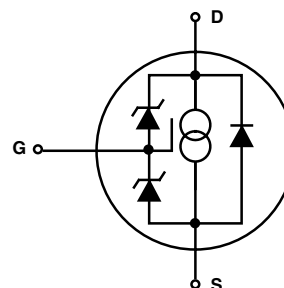
PART NUMBER	PACKAGE	BRAND
RLP1N06CLE	TO-220AB	L1N06CLE

NOTE: When ordering, use the entire part number.

Features

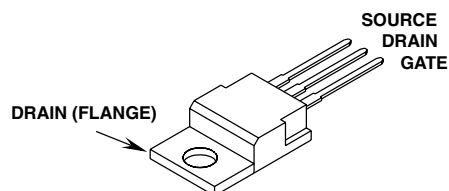
- 1A, 55V
- $r_{DS(ON)} = 0.750\Omega$
- I_{LIMIT} at 150°C = 1.1A to 1.5A Maximum
- Built-in Voltage Clamp
- Built-in Current Limiting
- ESD Protected, 2kV Minimum
- Controlled Switching Limits EMI and RFI
- 175°C Rated Junction Temperature
- Logic Level Gate
- Related Literature
 - TB334 "Guidelines for Soldering Surface Mount Components to PC Boards"

Symbol



Packaging

JEDEC TO-220AB



Typical Performance Curves Unless Otherwise Specified

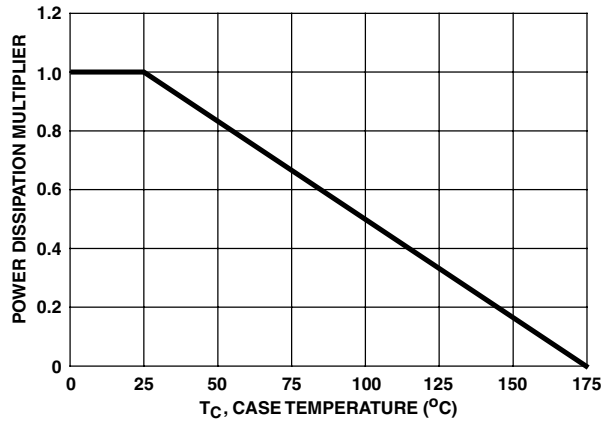


FIGURE 1. NORMALIZED POWER DISSIPATION vs CASE TEMPERATURE

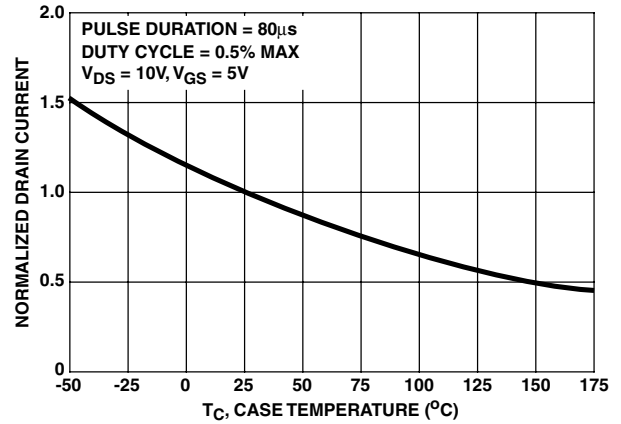


FIGURE 2. NORMALIZED CURRENT LIMIT vs CASE TEMPERATURE

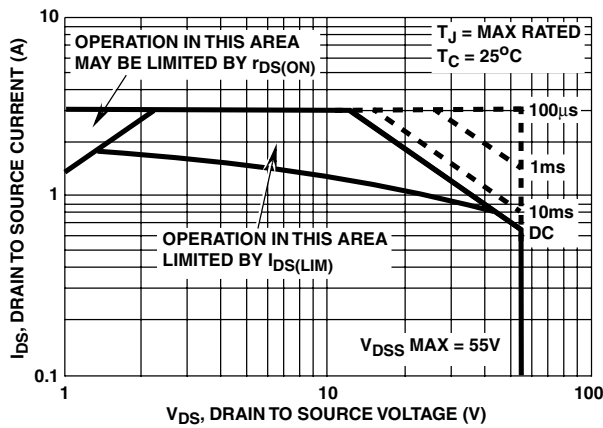


FIGURE 3. FORWARD BIAS SAFE OPERATING AREA

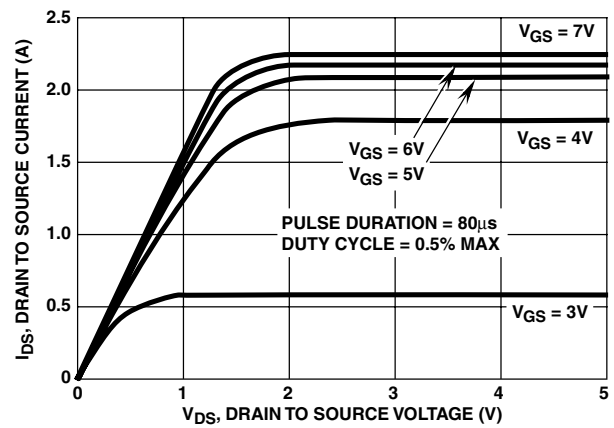


FIGURE 4. SATURATION CHARACTERISTICS

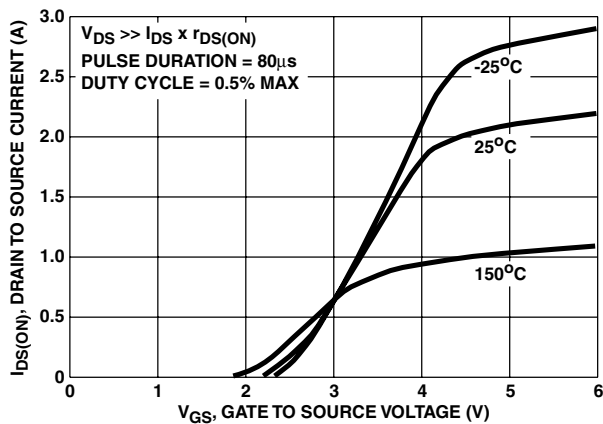


FIGURE 5. TRANSFER CHARACTERISTICS

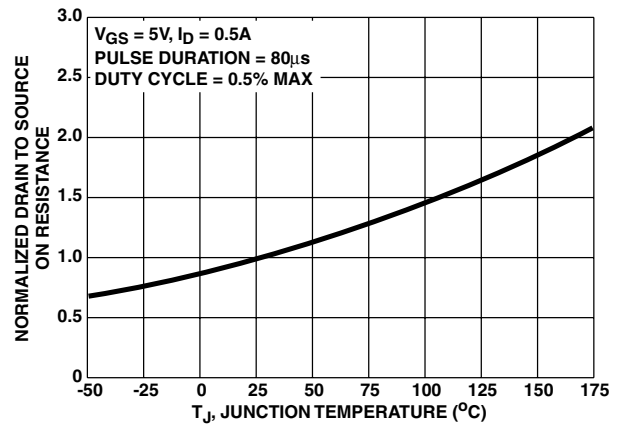


FIGURE 6. NORMALIZED DRAIN TO SOURCE ON RESISTANCE vs JUNCTION TEMPERATURE

Typical Performance Curves Unless Otherwise Specified (Continued)

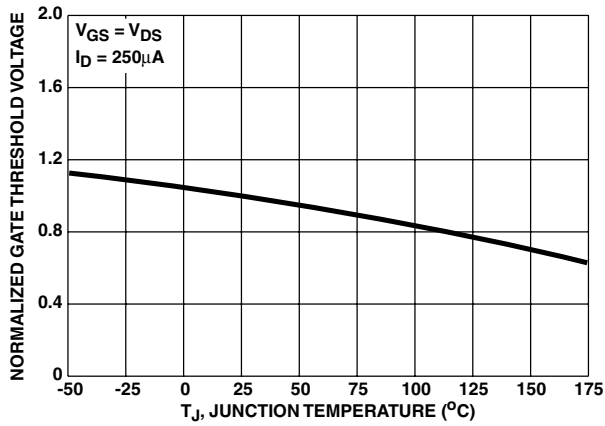


FIGURE 7. NORMALIZED GATE THRESHOLD VOLTAGE vs JUNCTION TEMPERATURE

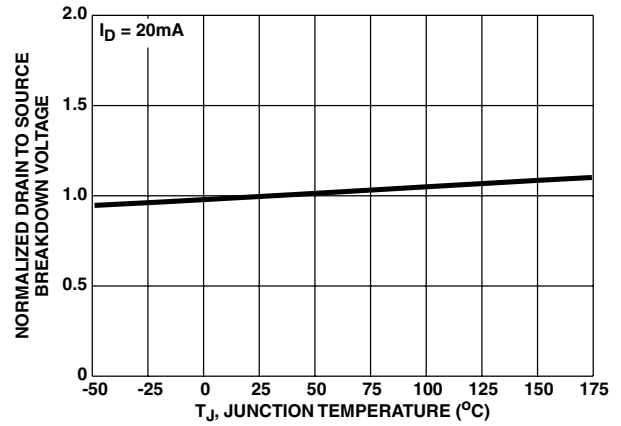


FIGURE 8. NORMALIZED DRAIN TO SOURCE BREAKDOWN VOLTAGE vs JUNCTION TEMPERATURE

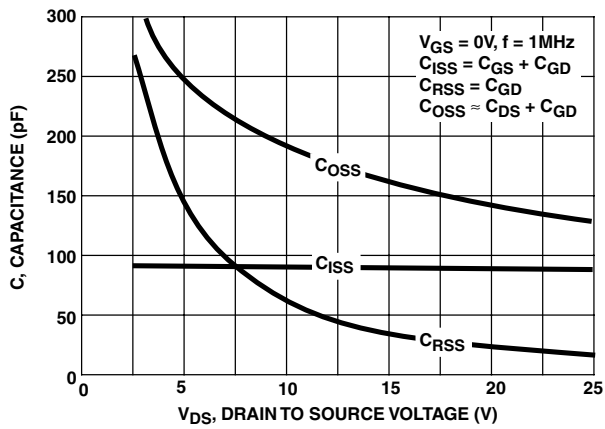


FIGURE 9. CAPACITANCE vs DRAIN TO SOURCE VOLTAGE

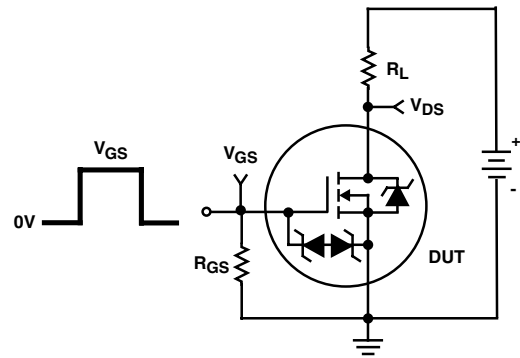


FIGURE 10. SWITCHING TEST CIRCUIT

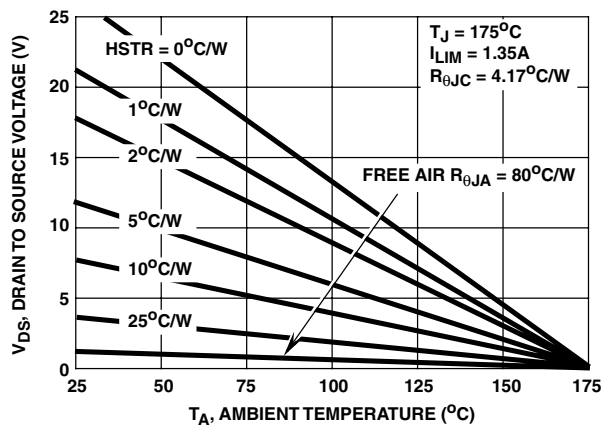
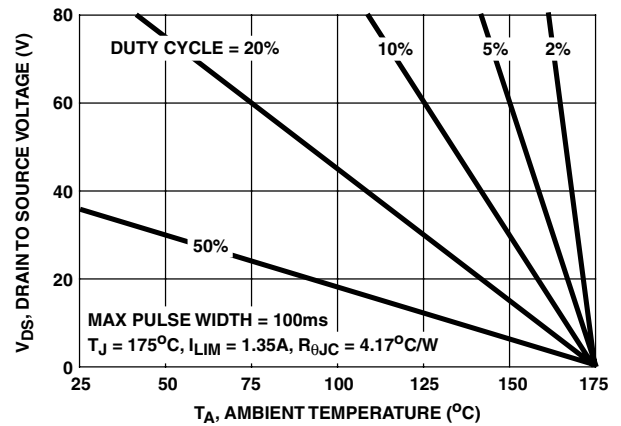


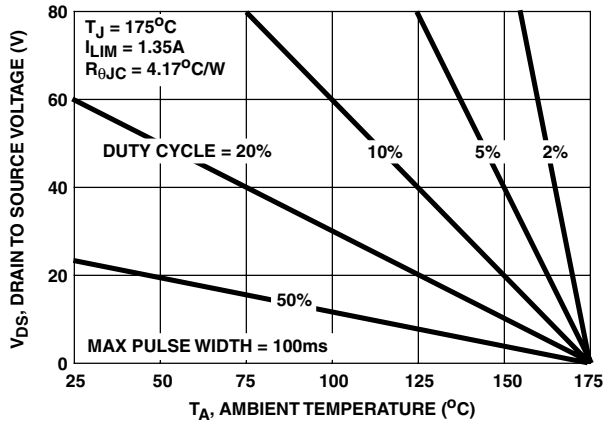
FIGURE 11. DC OPERATION IN CURRENT LIMITING



NOTE: Heatsink thermal resistance = $2^\circ C/W$

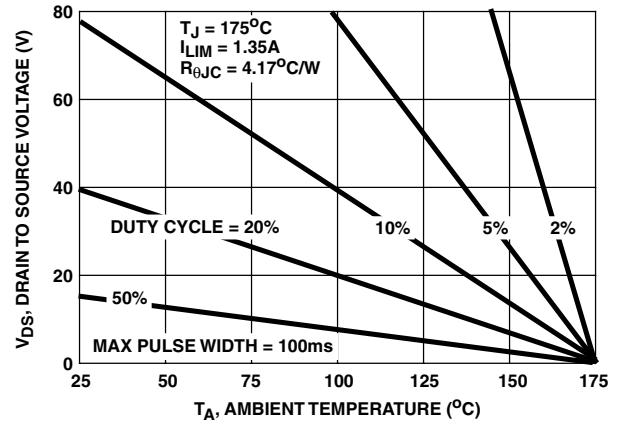
FIGURE 12. MAXIMUM V_{DS} vs T_A IN CURRENT LIMITING

Typical Performance Curves Unless Otherwise Specified (Continued)



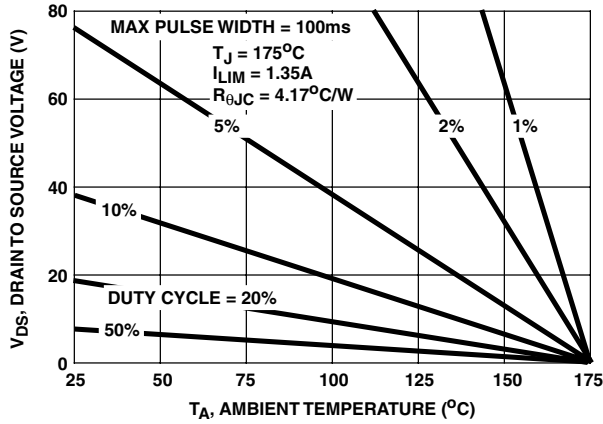
NOTE: Heatsink thermal resistance = 5°C/W

FIGURE 13. MAXIMUM V_{DS} vs T_A IN CURRENT LIMITING



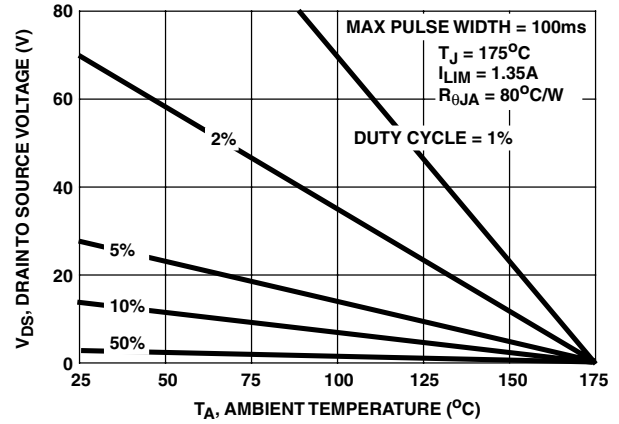
NOTE: Heatsink thermal resistance = 10°C/W

FIGURE 14. MAXIMUM V_{DS} vs T_A IN CURRENT LIMITING



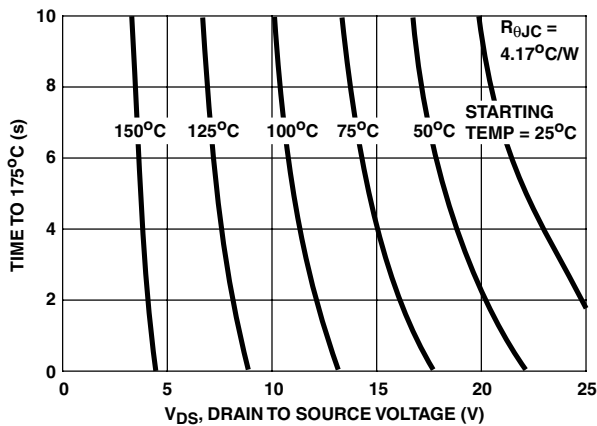
NOTE: Heatsink thermal resistance = 25°C/W

FIGURE 15. MAXIMUM V_{DS} vs T_A IN CURRENT LIMITING



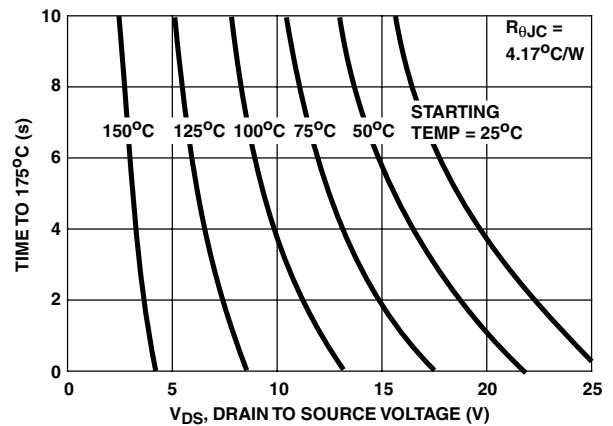
NOTE: No external heatsink

FIGURE 16. MAXIMUM V_{DS} vs T_A IN CURRENT LIMITING



NOTE: Heatsink thermal resistance = 2°C/W
Heatsink thermal capacitance = 4J/°C

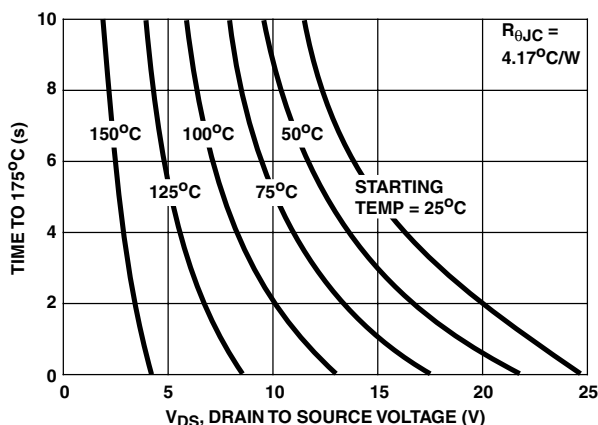
FIGURE 17. TIME TO 175°C IN CURRENT LIMITING



NOTE: Heatsink thermal resistance = 5°C/W
Heatsink thermal capacitance = 2J/°C

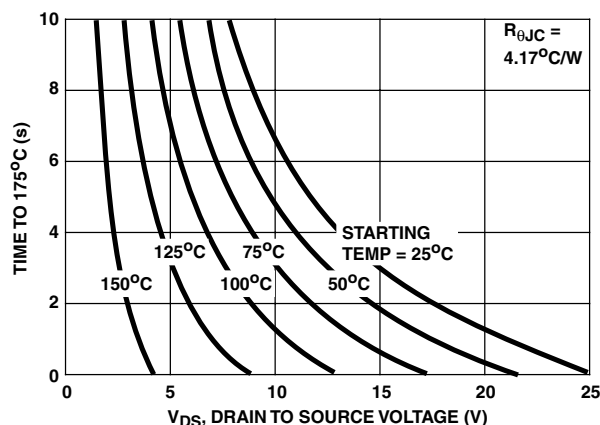
FIGURE 18. TIME TO 175°C IN CURRENT LIMITING

Typical Performance Curves Unless Otherwise Specified (Continued)



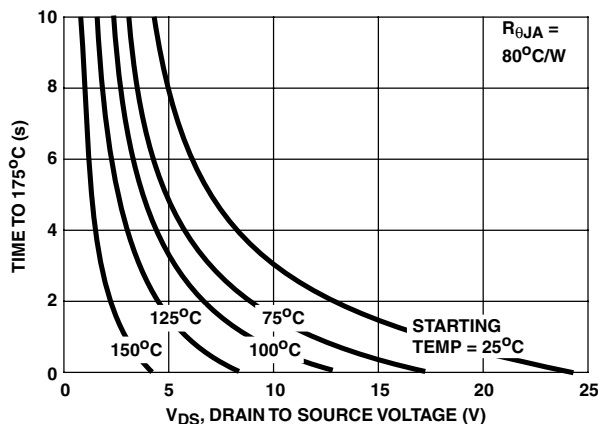
NOTE: Heatsink thermal resistance = 10°C/W
Heatsink thermal capacitance = $1\mu\text{J}/^{\circ}\text{C}$

FIGURE 19. TIME TO 175°C IN CURRENT LIMITING



NOTE: Heatsink thermal resistance = 25°C/W
Heatsink thermal capacitance = $0.5\mu\text{J}/^{\circ}\text{C}$

FIGURE 20. TIME TO 175°C IN CURRENT LIMITING



NOTE: No external heatsink

FIGURE 21. TIME TO 175°C IN CURRENT LIMITING

Detailed Description

Temperature Dependence of Current Limiting and Switching Speed

The RLP1N06CLE is a monolithic power device which incorporates a logic level PowerMOS transistor with a resistor in series with the source. The base and emitter of a lateral bipolar transistor is connected across this resistor, and the collector of the bipolar transistor is connected to the gate of the PowerMOS transistor. When the voltage across the resistor reaches the value required to forward bias the emitter base junction of the bipolar transistor, the bipolar transistor “turns on”. A series resistor is incorporated in series with the gate of the PowerMOS transistor allowing the bipolar transistor to drive the gate of the PowerMOS transistors to a voltage which just maintains a constant current in the PowerMOS transistor. Since both the resistance of the resistor

in series with the PowerMOS transistor source and voltage required to forward bias the base emitter junction of the bipolar transistor vary with the temperature, the current at which the device limits is a function of temperature. This dependence is shown in figure 2.

The resistor in series with the gate of the PowerMOS transistor results in much slower switching than in most PowerMOS transistors. This is an advantage where fast switching can cause EMI or RFI. The switching speed is very predictable, and a minimum as well as maximum fall time is given in the device characteristics for this type.

DC Operation of the RLP1N06CLE

The limit of the drain to source voltage for operation in current limiting on a steady state (DC) basis is shown as Figure 11. The dissipation in the device is simply the applied drain to source voltage multiplied by the limiting current. This

device, like most Power MOSFET devices today, is limited to 175°C. The maximum voltage allowable can, therefore be expressed as:

$$V_{DS} = \frac{(175^{\circ}\text{C} - T_{\text{AMBIENT}})}{I_{\text{LIM}} \times (R_{\theta\text{JC}} + R_{\theta\text{CA}})} \quad (\text{EQ. 1})$$

Duty Cycle Operation of the RLP1N06CLE

In many applications either the drain to source voltage or the gate drive is not available 100% of the time. The copper header on which the RLP1N06CLE is mounted has a very large thermal storage capability, so for pulse widths of less than 100 milliseconds, the temperature of the header can be considered a constant case temperature calculated simply as:

$$T_{\text{C}} = (V_{\text{DS}} \times I_{\text{D}} \times D \times R_{\theta\text{CA}}) + T_{\text{AMBIENT}} \quad (\text{EQ. 2})$$

Generally the heat storage capability of the silicon chip in a power transistor is ignored for duty cycle calculations. Making this assumption, limiting junction temperature to 175°C and using the T_{C} calculated above, the expression for maximum V_{DS} under duty cycle operation is:

$$V_{\text{DS}} = \frac{175 - T_{\text{C}}}{I_{\text{LIM}} \times D \times R_{\theta\text{JC}}} \quad (\text{EQ. 3})$$

These values are plotted as Figures 12 thru 16.

Limited Time Operations of the RLP1N06CLE

Protection for a limited period of time is sufficient for many applications. As stated above the heat storage in the silicon chip can usually be ignored for computations of over 10 milliseconds and the thermal equivalent circuit reduces to a simple enough circuit to allow easy computation on the limiting conditions. The variation in limiting current with temperature complicates the calculation of junction temperature, but a simple straight line approximation of the variation is accurate enough to allow meaningful computations. The curves shown as figures 17 thru 21 give an accurate indication of how long the specified voltage can be applied to the device in the current limiting mode without exceeding the maximum specified 175°C junction temperature. In practice this tells you how long you have to alleviate the condition causing the current limiting to occur.

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PRODUCT STATUS DEFINITIONS

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