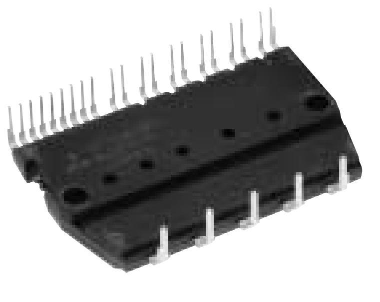
PS21562

TRANSFER-MOLD TYPE
INSULATED TYPE

PS21562



INTEGRATED POWER FUNCTIONS

600V/5A low-loss 5th generation IGBT inverter bridge for 3 phase DC-to-AC power conversion

INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

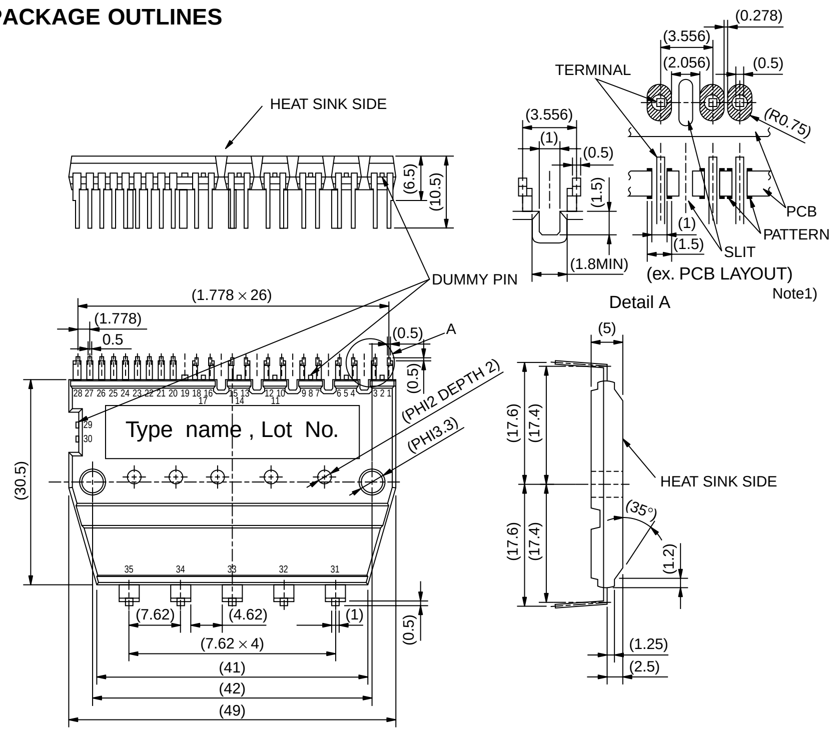
- For upper-leg IGBTS : Drive circuit, High voltage isolated high-speed level shifting, Control supply under-voltage (UV) protection.
- For lower-leg IGBTS : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC). (Fig.3)
- Fault signaling : Corresponding to an SC fault (Lower-side IGBT) or a UV fault (Lower-side supply).
- Input interface : 5V line CMOS/TTL compatible. (High Active)
- UL Approved : Yellow Card No. E80276

APPLICATION

AC100V~200V three-phase inverter drive for small power motor control.

Fig. 1 PACKAGE OUTLINES

Dimensions in mm



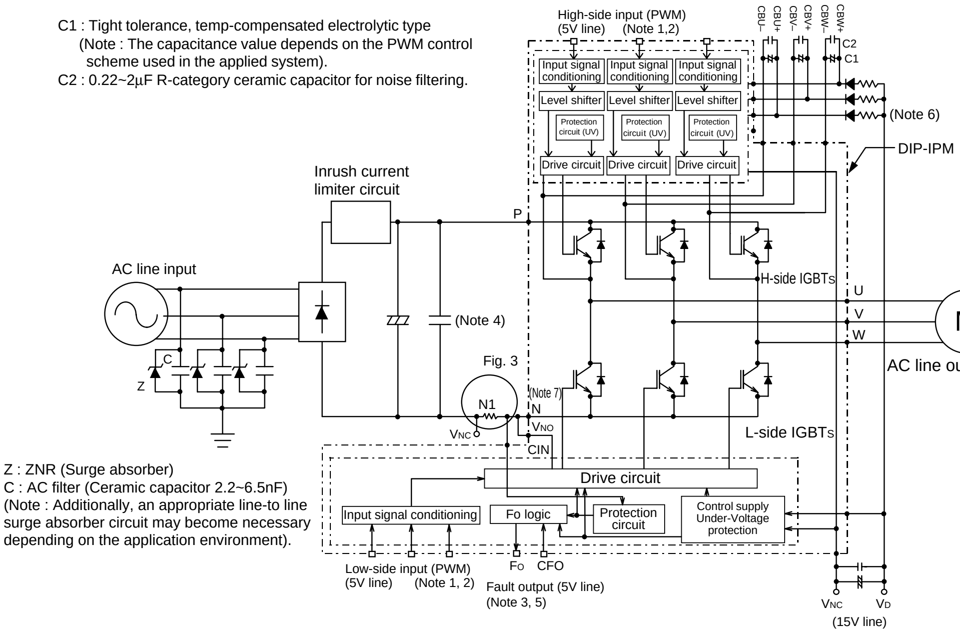
TERMINAL CODE

1	VUFS
2	(UPG)
3	VUFB
4	VP1
5	(COM)
6	UP
7	VVFS
8	(VPG)
9	VVFB
10	VP1
11	(COM)
12	VP
13	VWFS
14	(WPG)
15	VWFB
16	VP1
17	(COM)
18	WP
19	(UNG)
20	VNO Note2)
21	UN
22	VN
23	WN
24	FO
25	CFO
26	CIN
27	VNC
28	VN1
29	(WNG)
30	(VNG)
31	P
32	U
33	V
34	W
35	N

Note 1 : In order to get enough creepage distance between the terminals, please take some countermeasure such as a slit on PCB.
2 : The 20TH terminal VNO is treated as a NC in DIP-IPM ver.2, it should be connected with the terminal N outside in PS21562.

MITSUBISHI ELECTRIC

Fig. 2 INTERNAL FUNCTIONS BLOCK DIAGRAM (TYPICAL APPLICATION EXAMPLE)



Note1: Input logic is high-active. There is a 2.5kΩ (min) pull-down resistor built-in each input circuit. When using an external CR filter, please make it satisfy the input threshold voltage.

2: By virtue of integrating an application specific type HVIC inside the module, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible. (see also Fig. 8)

3: This output is open collector type. The signal line should be pulled up to the positive side of the 5V power supply with approximately 10kΩ resistance. (see also Fig. 8)

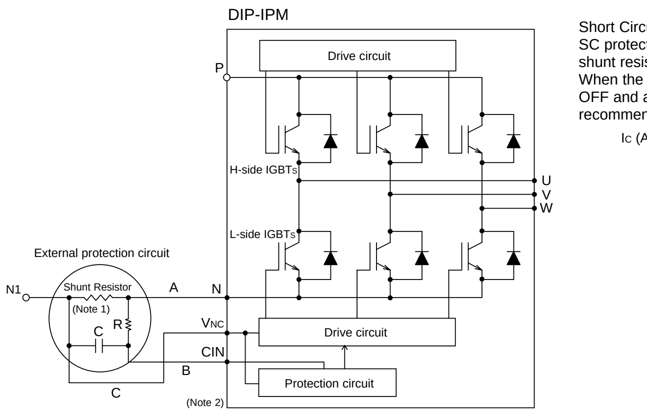
4: The wiring between the power DC link capacitor and the PN1 terminals should be as short as possible to protect the DIP-IPM against catastrophic high surge voltages. For extra precaution, a small film type snubber capacitor (0.1~0.22µF, high voltage type) is recommended to be mounted close to these PN1 DC power input pins.

5: Fo output pulse width should be decided by putting external capacitor between CFO and V_{NC} terminals. (Example : CFO=22nF → t_{FO}=1.8ms (Typ.))

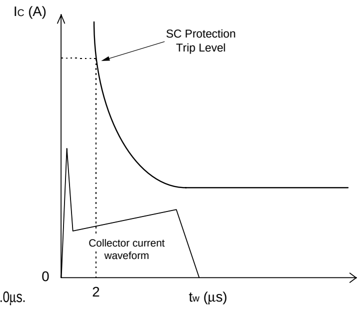
6: High voltage (600V or more) and fast recovery type (less than 100ns) diodes should be used in the bootstrap circuit.

7: The terminal V_{NO} should be connected to the terminal N outside of DIP-IPM.

Fig. 3 EXTERNAL PART OF THE DIP-IPM PROTECTION CIRCUIT



Short Circuit Protective Function (SC) :
SC protection is achieved by sensing the L-side DC-Bus current (through the external shunt resistor) after allowing a suitable filtering time (defined by the RC circuit).
When the sensed shunt voltage exceeds the SC trip-level, all the L-side IGBTs are turned OFF and a fault signal (Fo) is output. Since the SC fault may be repetitive, it is recommended to stop the system when the Fo signal is received and check the fault.



Note1: In the recommended external protection circuit, please select the RC time constant in the range 1.5~2.0µs.
2: To prevent erroneous protection operation, the wiring of A, B, C should be as short as possible.

MAXIMUM RATINGS (T_j = 25°C, unless otherwise noted)

INVERTER PART

Symbol	Parameter	Condition	Ratings	Unit
V_{CC}	Supply voltage	Applied between P-N	450	V
$V_{CC(surge)}$	Supply voltage (surge)	Applied between P-N	500	V
V_{CES}	Collector-emitter voltage		600	V
$\pm I_C$	Each IGBT collector current	T_f = 25°C	5	A
$\pm I_{CP}$	Each IGBT collector current (peak)	T_f = 25°C, less than 1ms	10	A
P_C	Collector dissipation	T_f = 25°C, per 1 chip	16.7	W
T_j	Junction temperature	(Note 1)	–20~+125	°C

Note 1 : The maximum junction temperature rating of the power chips integrated within the DIP-IPM is 150°C (@ $T_f \leq$ 100°C) however, to ensure safe operation of the DIP-IPM, the average junction temperature should be limited to $T_{j(ave)} \leq$ 125°C (@ $T_f \leq$ 100°C).

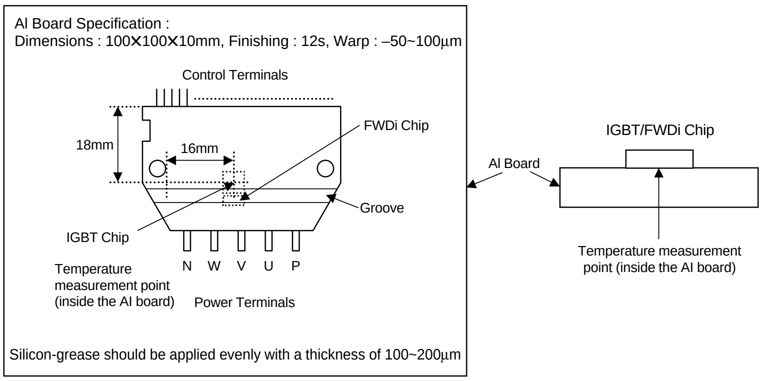
CONTROL (PROTECTION) PART

Symbol	Parameter	Condition	Ratings	Unit
V_D	Control supply voltage	Applied between V_{P1}-V_{NC}, V_{N1}-V_{NC}	20	V
V_{DB}	Control supply voltage	Applied between V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}	20	V
V_{IN}	Input voltage	Applied between U_P, V_P, W_P-V_{NC}, U_N, V_N, W_N-V_{NC}	–0.5~V_D+0.5	V
V_{FO}	Fault output supply voltage	Applied between F_O-V_{NC}	–0.5~V_D+0.5	V
I_{FO}	Fault output current	Sink current at F_O terminal	1	mA
V_{SC}	Current sensing input voltage	Applied between CIN-V_{NC}	–0.5~V_D+0.5	V

TOTAL SYSTEM

Symbol	Parameter	Condition	Ratings	Unit
$V_{CC(PROT)}$	Self protection supply voltage limit (short circuit protection capability)	V_D = 13.5~16.5V, Inverter part T_j = 125°C, non-repetitive, less than 2 µs	400	V
T_f	Module case operation temperature	(Note 2)	–20~+100	°C
T_{stg}	Storage temperature		–40~+125	°C
V_{iso}	Isolation voltage	60Hz, Sinusoidal, AC 1 minute, connection pins to heat-sink plate	2500	V_{rms}

Note 2 : T_f MEASUREMENT POINT



THERMAL RESISTANCE

Symbol	Parameter	Condition	Limits Min.	Limits Typ.	Limits Max.	Unit
$R_{th(j-f)Q}$	Junction to case thermal resistance (Note 3)	Inverter IGBT part (per 1/6 module)	—	—	6.0	°C/W
$R_{th(j-f)F}$		Inverter FWDi part (per 1/6 module)	—	—	6.5	°C/W

Note 3: Grease with good thermal conductivity should be applied evenly with about +100µm~+200µm on the contacting surface of DIP-IPM and heat-sink.

ELECTRICAL CHARACTERISTICS (T_j = 25°C, unless otherwise noted)

INVERTER PART

Symbol	Parameter	Condition		Limits Min.	Limits Typ.	Limits Max.	Unit
$V_{CE(sat)}$	Collector-emitter saturation voltage	$V_D = V_{DB}$ = 15V, V_{IN} = 5V	I_C = 5A, T_j = 25°C	—	1.60	2.10	V
			I_C = 5A, T_j = 125°C	—	1.70	2.20	
V_{EC}	FWDi forward voltage	T_j = 25°C, $-I_C$ = 5A, V_{IN} = 0V		—	1.50	2.00	V
t_{on}	Switching times	V_{CC} = 300V, $V_D = V_{DB}$ = 15V, I_C = 5A, T_j = 125°C, V_{IN} = 0 ↔ 5V, Inductive load (upper-lower arm)		0.60	1.20	1.80	µs
t_{rr}				—	0.30	—	µs
$t_{c(on)}$				—	0.40	0.60	µs
t_{off}				—	1.30	2.00	µs
$t_{c(off)}$				—	0.50	0.80	µs
I_{CES}	Collector-emitter cut-off current	$V_{CE} = V_{CES}$	T_j = 25°C	—	—	1	mA
			T_j = 125°C	—	—	10	

CONTROL (PROTECTION) PART

Symbol	Parameter	Condition		Limits Min.	Limits Typ.	Limits Max.	Unit
I_D	Circuit current	$V_D = V_{DB}$ = 15V, V_{IN} = 5V	Total of V_{P1}-V_{NC}, V_{N1}-V_{NC}	—	—	5.00	mA
			V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}	—	—	0.40	mA
		$V_D = V_{DB}$ = 15V, V_{IN} = 0V	Total of V_{P1}-V_{NC}, V_{N1}-V_{NC}	—	—	7.00	mA
			V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}	—	—	0.55	mA
V_{FOH}	Fault output voltage	V_{SC} = 0V, F_O circuit pull-up to 5V with 10kΩ		4.9	—	—	V
V_{FOL}		V_{SC} = 1V, I_{FO} = 1mA		—	—	0.95	V
$V_{SC(ref)}$	Short circuit trip level	T_j = 25°C, V_D = 15V (Note 4)		0.43	0.48	0.53	V
I_{IN}	Input current	V_{IN} = 5V		1.0	1.5	2.0	mA
UV_{DBt}	Supply circuit under-voltage protection	T_j ≤ 125°C	Trip level	10.0	—	12.0	V
UV_{DBr}			Reset level	10.5	—	12.5	V
UV_{Dt}			Trip level	10.3	—	12.5	V
UV_{Dr}			Reset level	10.8	—	13.0	V
t_{FO}	Fault output pulse width	C_{FO} = 22nF (Note 5)		1.0	1.8	—	ms
$V_{th(on)}$	ON threshold voltage	Applied between U_P, V_P, W_P-V_{NC}, U_N, V_N, W_N-V_{NC}		2.1	2.3	2.6	V
$V_{th(off)}$	OFF threshold voltage			0.8	1.4	2.1	V

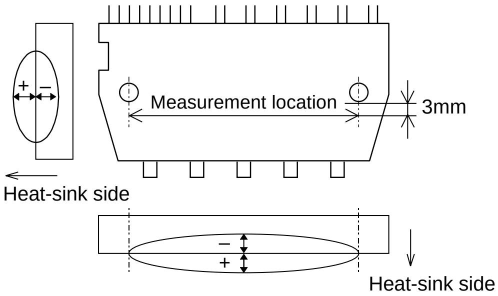
Note 4: Short circuit protection is functioning only at the low-arms. Please select the value of the external shunt resistor such that the SC trip-level is less than 8.5 A.

5: Fault signal is output when the low-arms short circuit or control supply under-voltage protective functions operate. The fault output pulse-width t_{FO} depends on the capacitance value of C_{FO} according to the following approximate equation : $C_{FO} = 12.2 \times 10^{-6} \times t_{FO}$ [F].

MECHANICAL CHARACTERISTICS AND RATINGS

Parameter	Condition		Limits Min.	Limits Typ.	Limits Max.	Unit
Mounting torque	Mounting screw : M3	Recommended 0.78 N·m	0.59	—	0.98	N·m
Weight			—	20	—	g
Heat-sink flatness		(Note 6)	–50	—	100	µm

Note 6: Measurement point of heat-sink flatness



RECOMMENDED OPERATION CONDITIONS

Symbol	Parameter	Condition	Limits Min.	Limits Typ.	Limits Max.	Unit
V_{CC}	Supply voltage	Applied between P-N	0	300	400	V
V_D	Control supply voltage	Applied between V_{P1}-V_{NC}, V_{N1}-V_{NC}	13.5	15.0	16.5	V
V_{DB}	Control supply voltage	Applied between V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}	13.0	15.0	18.5	V
ΔV_D, ΔV_{DB}	Control supply variation		–1	—	1	V/µs
t_{dead}	Arm shoot-through blocking time	For each input signal, $T_f \leq 100°C$	1.5	—	—	µs
f_{PWM}	PWM input frequency	$T_f \leq 100°C$, $T_j \leq 125°C$	—	10	—	kHz
I_O	Allowable r.m.s. current	V_{CC} = 300V, V_D = 15V, fc = 10kHz P.F = 0.8, sinusoidal $T_j \leq 125°C$, $T_f \leq 100°C$ (Note 7)	—	—	3.0	Arms
PWIN	Minimum input pulse width	ON (Note 8)	300	—	—	ns
V_{NC}	V_{NC} variation	between V_{NC}-N (including surge)	–5.0	—	5.0	V

Note 7 : The allowable r.m.s. current value depends on the actual application conditions.
8 : The input pulse width less than PWIN might make no response.

Fig. 4 THE DIP-IPM INTERNAL CIRCUIT

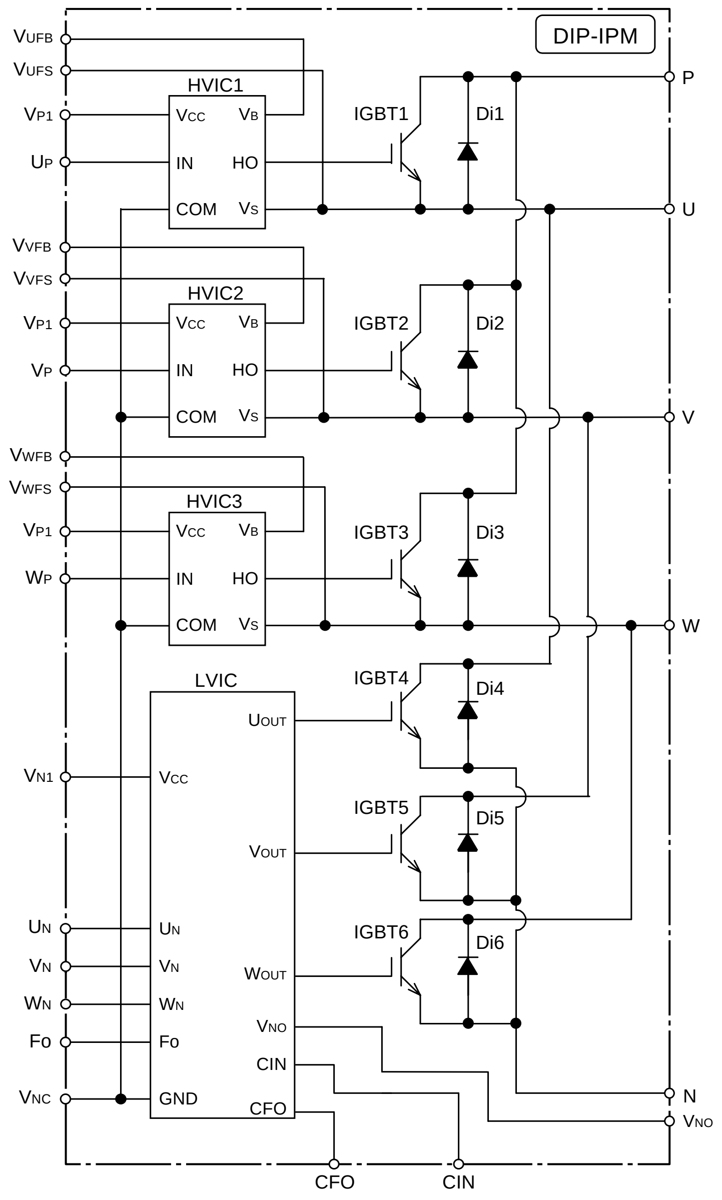
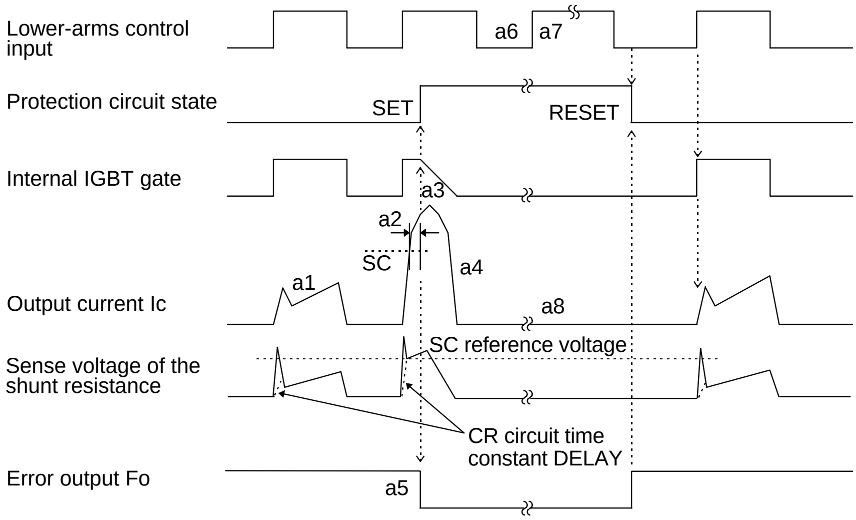


Fig. 5 TIMING CHARTS OF THE DIP-IPM PROTECTIVE FUNCTIONS

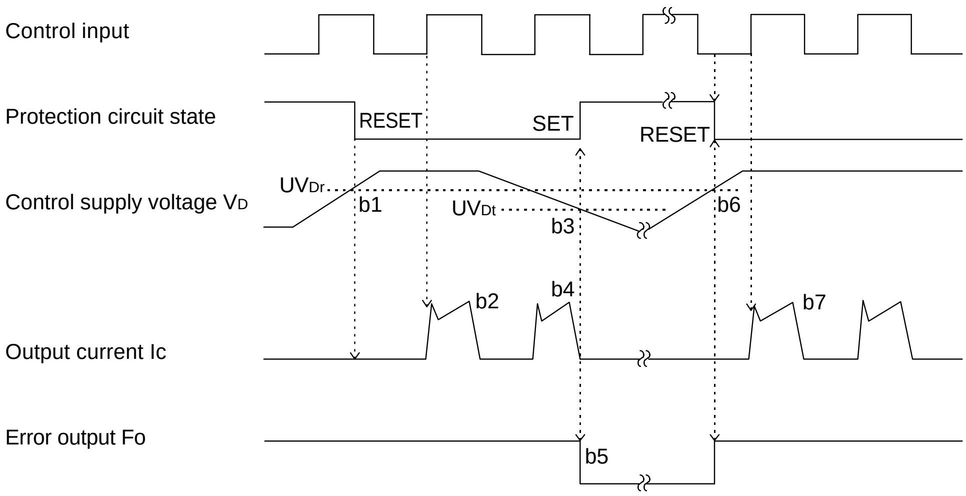
[A] Short-Circuit Protection (Lower-arms only)

(With the external shunt resistance and CR connection)
a1. Normal operation : IGBT ON and carrying current.
a2. Short circuit current detection (SC trigger).
a3. Hard IGBT gate interrupt.
a4. IGBT turns OFF.
a5. F_O timer operation starts : The pulse width of the F_O signal is set by the external capacitor C_{FO}.
a6. Input “L” : IGBT OFF state.
a7. Input “H” : IGBT ON state, but during the F_O active signal period the IGBT doesn’t turn ON.
a8. IGBT OFF state.



[B] Under-Voltage Protection (Lower-arm, UV_D)

b1. Control supply voltage rises : After the voltage level reaches UV_{Dr}, the circuits start to operate when next input is applied.
b2. Normal operation : IGBT ON and carrying current.
b3. Under voltage trip (UV_{Dt}).
b4. IGBT OFF in spite of control input condition.
b5. F_O operation starts.
b6. Under voltage reset (UV_{Dr}).
b7. Normal operation : IGBT ON and carrying current.



[C] Under-Voltage Protection (Upper-arm, UV_{DB})

c1. Control supply voltage rises : After the voltage reaches UV_{DBr}, the circuits start to operate when next input is applied.
c2. Normal operation : IGBT ON and carrying current.
c3. Under voltage trip (UV_{DBt}).
c4. IGBT OFF in spite of control input condition, but there is no F_O signal output.
c5. Under voltage reset (UV_{DBr}).
c6. Normal operation : IGBT ON and carrying current.

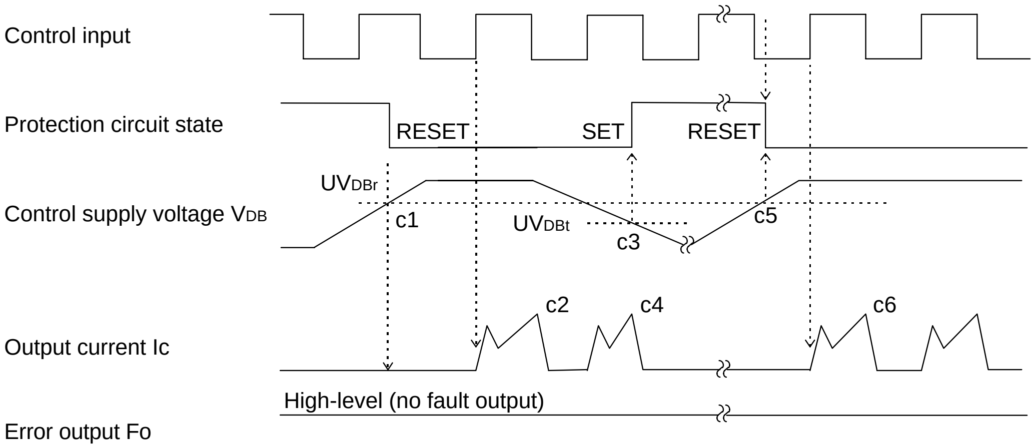
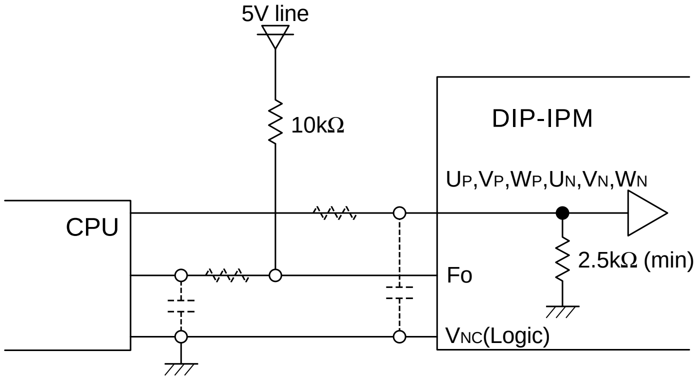


Fig. 6 RECOMMENDED CPU I/O INTERFACE CIRCUIT



Note : RC coupling at each input (parts shown dotted) may change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board.
The DIP-IPM input signal section integrates a 2.5kΩ(min) pull-down resistor. Therefore, when using a external filtering resistor, care must be taken to satisfy the turn-on threshold voltage requirement.

Fig. 7 RECOMMENDED WIRING OF SHUNT RESISTANCE

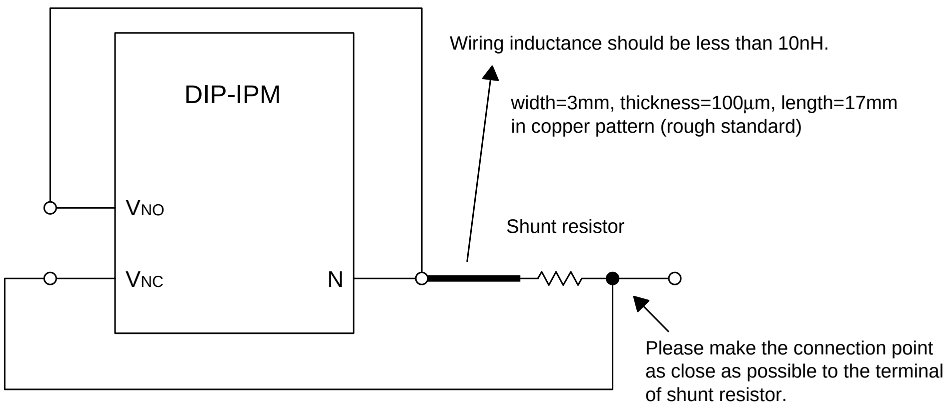
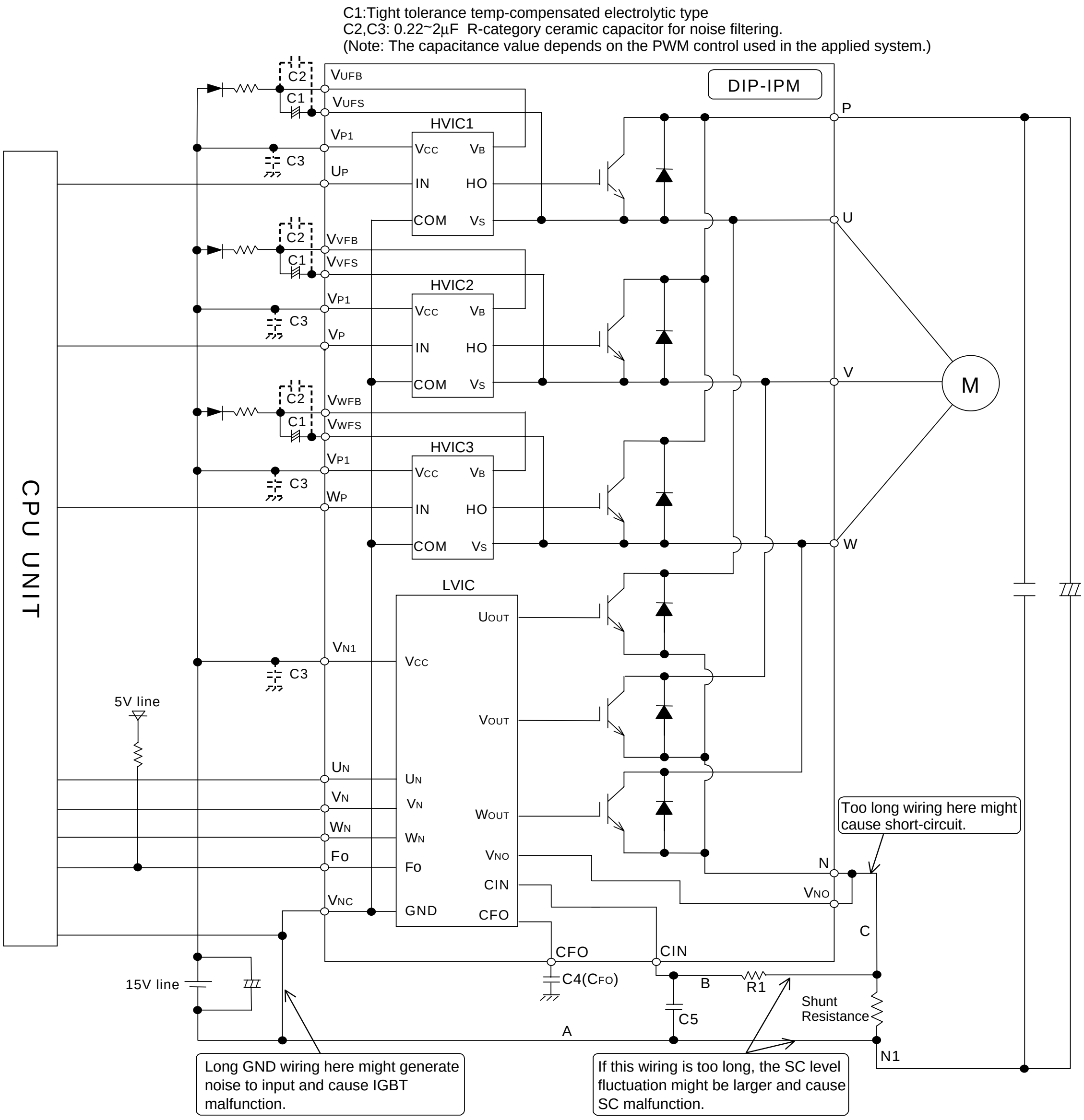


Fig. 8 TYPICAL DIP-IPM APPLICATION CIRCUIT EXAMPLE



Note 1 : To prevent the input signals oscillation, the wiring of each input should be as short as possible. (Less than 2cm)
2 : By virtue of integrating an application specific type HVIC inside the module, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.
3 : F_O output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 10kΩ resistor.
4 : F_O output pulse width is determined by the external capacitor between CFO and V_{NC} terminals (C_{FO}). (Example : C_{FO} = 22 nF → t_{FO} = 1.8 ms (typ.))
5 : The logic of input signal is high-active. The DIP-IPM input signal section integrates a 2.5kΩ (min) pull-down resistor. Therefore, when using external filtering resistor, care must be taken to satisfy the turn-on threshold voltage requirement.
6 : To prevent malfunction of protection, the wiring of A, B, C should be as short as possible.
7 : Please set the R_1C_5 time constant in the range 1.5~2µs.
8 : Each capacitor should be located as nearby the pins of the DIP-IPM as possible.
9 : To prevent surge destruction, the wiring between the smoothing capacitor and the P&N1 pins should be as short as possible. Approximately a 0.1~0.22µF snubber capacitor between the P&N1 pins is recommended.
10 : The terminal V_{NO} should be connected with the terminal N outside.