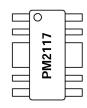
PM 2117

DATA SHEET

Single Supply RFIC Power Amplifier 2400 to 2500 MHz Operation

Features

- 29 dBm Output Power
- 40% Efficiency
- Single 3V to 5V Supply Class A Operation
- Linear Class AB Operation (Requires -V_{G2})*
- 50 Ω Matched Input, Simple External Output Match
- Unconditionally Stable
- NEW PM-SOP™ Package (No bottom side contact)



PM-SOPTM Plastic Package

Plastic Package L= .196" x W= .079" x H=.042"

Applications

- FHSS or DSSS ISM Band Transmitters
- RFID Interrogators
- MMDS Band Transmitters

Description

The PM2117 is a two-stage high efficiency GaAs RFIC power amplifier developed for applications in the 2400 MHz ISM band. The input and interstage matching networks are included on chip. A simple series line, shunt capacitor output matching network keeps the additional component count to just 9 parts, including all bypass and blocking capacitors. In single supply mode both stages are inherently biased for Class A operation: the current will remain constant, or increase somewhat under input power back-off. Operating from a single 5 volt supply in Class A operation the PM2117 can produce over 800 mW saturated output power with 40% typical power added efficiency and approximately 400 mW from a single 3 V supply with a typical supply current of less than 300 mA

* NOTE: For linear Class AB operation, supplying a negative 0.3 volts to $V_{\rm G2}$ will reduce current consumption when in an idle or backed off power mode. (See Application Note 2117AB for more details on linear operation.)

Electrical Characteristics

Typical Specifications for V_{DD} = 5.0V T_A = +25°C as tested in 50 Ω system, using matching circuit on page 3. **Minimum and Maximum Specifications are Guaranteed over Frequency and Temperature**

Parameter Symbol **Conditions** Min Typ Max Units Frequency Range 2400 2500 MHz Power Output $P_{\text{SAT}} \\$ $P_{\rm IN}\!=6~dBm$ 27.5 29 dBm Power Added Efficiency $P_{IN} = 6 dBm$ 30 40 % η Small Signal Gain 26 dBG Input Return Loss S_{11} 14 dB $P_{1dB} \\$ Output Power at 1dB Comp. 27 dBm Operating Drain Current $P_{OUT} = P_{SAT}$ 430 500 mΑ I_{DD} Stability Factor k With or without match 1.5 Thermal Resistance 35 $\theta_{\rm IC}$ Junction to GND lead °C/W

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Absolute Maximum Ratings

Characteristics	Symbol	Value	Units
Drain Voltage	$V_{\mathrm{DD1,2}}$	+7.0	V
Bias Current	I_{DS}	900	mA
RF Input Power	$P_{\rm IN}$	+15.0	dBm
Power Dissipation	P_{DISS}	1.9	W
Load VSWR	VSWR	10:1	
Operating Temperature	T_{OP}	-40 to +85	°C
Junction Temperature	T_{J}	150	°C
Storage Temperature Range	T_{STG}	-65 to +150	°C

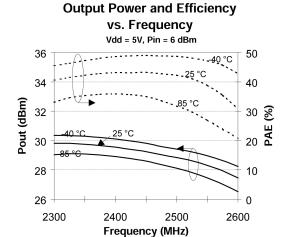
RF_{IN} 2 PM2117 RF_{OUT}/V_{DD2} RF_{OUT}/V_{DD2} 3, 6, 8, 10

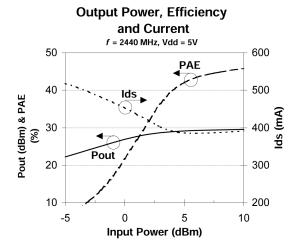
Pin-Out

Caution: Operating beyond the specified rating for any of these parameters may cause permanent damage to device.

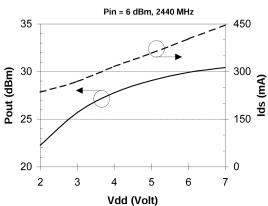
Typical Performance Characteristics

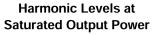
(Obtained using the suggested matching circuit shown on page 3.)

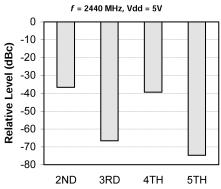




Output Power & Supply Current Vs Vdd

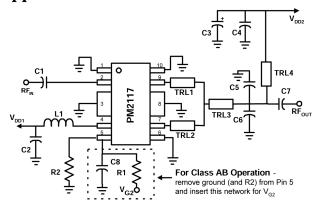






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Application Information



Suggested Matching Circuit for 2.4 to 2.5 GHz Operation.

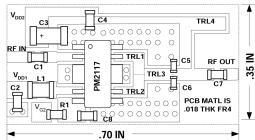
Value / Type Size Part 1.2 pF NPO C5, C6 0402 C2 1000 pF X7R 0603 C1, C7 5.6 pF NPO 0603 100 pF NPO 0603 C4 C3 6.8 μF TANT 1206 L1 12 nH 0802 R2 50Ω 0603 TRL1, 2 $\theta = 15^{\circ}, Z_0 = 60 \Omega$ @2.44 GHz TRL3 $\theta = 12^{\circ}, Z_0 = 50 \Omega$ @2.44 GHz TRL4 $\theta=90^{\circ},\,Z_0=105\;\Omega$ @2.44 GHz

Required only for Class AB Operation:					
R1	10 Ω	0603			
C8	1000 pF X7R	0603			

The PM2117 is internally matched on the input and interstage, so only output matching is required. A series line, shunt C network is required to

match the output. Pins 7 and 9 provide the RF output. Separate transmission lines (TRL1 and TRL2) on each output provide a small impedance transformation before the main series transmission line (TRL3). Two parallel capacitors, C5 and C6 complete the series-L, shunt-C matching network. Two 0402-size capacitors are used for the equivalent shunt C to reduce parasitic inductance effects within the capacitors. L1 is required for supply line isolation to avoid excess gain at low frequencies. For Class AB Operation R1 and C8 replaces the direct connection to ground as shown above. $V_{\rm G2}$ = -0.3 Vdc. (See Application Note 2117AB)

PCB Layout (PCB material = 0.018" FR4, ε_r = 4)



The placement and size (0402) of C5 and C6 are critical with respect to the IC package and the TRL's. Care should be taken to produce trace dimensions to achieve the TRL impedance's and phase lengths shown above. Dashed lines in this layout represent reference planes. All other components and traces may be moved to accommodate layout constraints. Pins 3 and 8 provide the source grounding for the FET's. Via holes (.010" dia.) should be placed in close proximity to the lead edges to minimize ground inductance.

Note: R1 and C8 are for Class AB operation. For single supply Class A operation, remove R1 and replace C8 with R2. There is no V_{G2} in Class A operation.

Biasing

The PM2117 requires no negative bias for Class A operation because both FET's operate at 0 volts $V_{\rm GS}$. At high input RF drive levels (> 3 dBm), efficient PA operation is possible because the output current swings between zero and $I_{\rm max}$, which is roughly 200% above $I_{\rm DSS}$. Under saturated operation, the operating DC current is 1/3 $I_{\rm max}$. If the RF drive is backed off, the drain current rises to $I_{\rm DSS}$. As such, "small-signal" operation will cause the current draw to increase by about 50%, to about 600 mA. This will not cause damage to the device for a short duration. Class AB operation is suggested and is accomplished by applying a negative 0.3 volts ($V_{\rm G2}$) and follow the schematic above. At 5 V Class A operation, the maximum duty cycle for small-signal operation is limited to 60% in order to remain below the 150 °C maximum recommended average junction temperature. Continuous operation is possible at 3V. For saturated operation, the drain efficiency increases to allow 100% duty cycle up to 5V.

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PM2117 Small-Signal S-Parameters					(V_{DD} = 5V, Class A Operation)					
	S.	11	S	21	S	12	S	22	MAG	k-Factor
Freq.	MAG	ANG	MAG	ANG	MAG	ANG	MAG	ANG	dB	
500	0.820	153.5	2.959	-0.6	0.0024	29.3	0.612	-159.2	16.3	14.91
1000	0.789	126.3	1.737	-136.7	0.0037	-24.9	0.791	177.0	13.3	10.39
1500	0.733	96.9	0.889	-127.7	0.0034	-102.1	0.706	169.7	5.3	37.88
2000	0.606	54.5	6.292	167.4	0.0012	104.0	0.648	172.6	20.3	23.98
2200	0.432	23.5	11.070	108.6	0.0007	-65.1	0.734	177.3	25.0	24.55
2300	0.285	9.4	12.201	71.1	0.0061	-77.4	0.791	174.0	26.4	2.27
2350	0.226	7.0	11.674	55.5	0.0064	-92.9	0.806	171.9	26.1	2.43
2400	0.168	7.3	11.386	37.9	0.0068	-112.5	0.816	169.7	26.0	2.00
2450	0.124	9.8	11.238	22.9	0.0070	-125.3	0.821	167.6	26.0	2.17
2500	0.100	17.6	10.380	9.6	0.0061	-133.4	0.818	165.7	25.2	2.57
2550	0.080	29.3	10.134	-5.0	0.0054	-143.7	0.815	163.8	24.9	3.14
2600	0.071	34.7	9.307	-15.8	0.0058	-152.0	0.808	162.6	24.0	3.18
3000	0.038	-165.5	6.245	-97.6	0.0076	160.5	0.749	156.1	19.5	4.67
4000	0.329	175.6	2.871	103.9	0.0116	9.8	0.649	146.5	12.0	7.66
5000	0.482	145.3	1.960	-41.6	0.0142	-99.4	0.599	126.7	8.9	8.89
6000	0.510	110.1	1.410	177.4	0.0212	143.1	0.521	98.6	5.7	9.00

Thermal Considerations

The fused leads (pins 3, 8) on the PM2117 provide most of the thermal path from the device junction. In the absence of other heat conducting structures, we recommend 3 sq-in of continuous copper ground plane per watt of dissipated power (Note: $P_{DISS} = Duty\ Cycle \times (P_{DC} - P_{RF})$). 2 oz. copper is recommended for the ground plane layer, with a minimum of 20 via holes (.010 diameter) to the component side in close proximity to the ground leads.

Pin Connections

Pin Number	Function
1	GND
2	$\mathrm{RF}_{\mathrm{IN}}$
3	GND
4	$ m V_{DD1}$
5	GND (ClassA)
5	V _{G2} (Class AB)
6	GND
7	RF_{OUT}/V_{DD2}
8	GND
9	RF_{OUT}/V_{DD2}
10	GND

Part Number Marking:

The PM2117 shall be marked as follows:

Model Number Code: "117" = PM2117

Lot Date Code: "Y" = Last Digit of Year

"WW" = Week During Year

$PM\text{-}SOP^{TM} \ Outline \ Drawing$

