(Revised 1/28/2002)



#### **Features**

- Dual Outputs (Independently Regulated)
- Power-up/Down Sequencing
- Input Voltage Range: 18V to 36V
- 1500 VDC Isolation
- Temp Range: -40° to 100°C
- High Efficiency: 88%
- Fixed Frequency Operation
- Over-Current Protection (Both Outputs)

- Dual Logic On/Off Control
- Over-Temperature Shutdown
- Over-Voltage Protection (Coordinated Shutdown)
- Under-Voltage Lockout
- Input Differential EMI Filter
- Solderable Copper Case
- Safety Approvals (Pending): UL 60950 CSA 22.2 60950

### **Description**

The PT4680 Excalibur™ Series is a dual-output isolated DC/DC converter that combine state-of-theart power conversion technology with unparalleled flexibility. Operating from a (–24V) industry standard input bus, the PT4680 series provides up to 20 ADC of output current from two independently regulated voltages (each output 15 ADC max).

The PT4680 series is characterized with high efficiencies and ultra-fast transient response, and incorporates many features to facilitate system integration. These include a flexible "On/Off" enable control, output

current limit, over-temperature protection, and an input under-voltage lock-out. In addition, both output voltages are designed to meet the power-up/power -down sequencing requirements of popular DSP ICs.

The PT4680 series is housed in space-saving solderable copper case. The package does not require a heatsink and is available in both vertical and horizontal configurations, including surface mount. The 'N' configuration occupies less than 2 in² of PCB area.

## **Ordering Information**

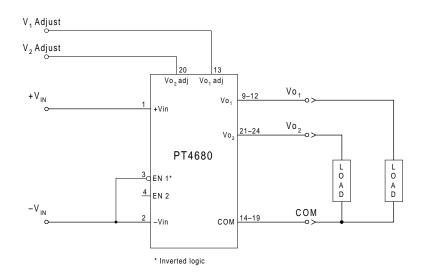
Pt. No.	Vo <sub>1</sub> /Vo <sub>2</sub>
PT4681□	= 5.0/3.3 Volts
PT4682□	= 3.3/2.5  Volts
PT4683□	= 3.3/1.8 Volts
PT4685□	= 3.3/1.5  Volts
PT4686□	= 2.5/1.8  Volts
PT4687□	= 5.0/1.8 Volts

## PT Series Suffix (PT1234x)

Case/Pin Configuration	Order Suffix	Package Code
Vertical	N	(EKD)
Horizontal	Α	(EKA)
SMD	С	(EKC)

(Reference the applicable package code drawing for the dimensions and PC board layout)

## **Typical Application**



#### **Pin-Out Information**

Pin	Function
1	+Vin
2	-Vin
3	EN 1
4	EN 2
5	TEMP
6	AUX
7	Do Not Connect
8	Do Not Connect
9	+Vo <sub>1</sub>

Note: Shaded functions indicate signals that are referenced to the input (-Vin) potential.

Pin	Function
10	$+Vo_1$
11	+Vo <sub>1</sub>
12	+Vo <sub>1</sub>
13	Vo <sub>1</sub> Adjust
14	COM
15	COM
16	COM
17	COM
18	COM

Pin	Function
19	COM
20	Vo <sub>2</sub> Adjust
21	+Vo,
22	+Vo,
23	+Vo,
24	+Vo,
25	Do Not Connect
26	Do Not Connect

## **On/Off Logic**

Pin 3	Pin 4	Output Status
1	×	Off
0	1	On
×	0	Off

#### Notes

 $\begin{array}{l} Logic \ 1 = Open \ collector \\ Logic \ 0 = -Vin \ (pin \ 2) \ potential \end{array}$ 

For positive Enable function, connect pin 3 to pin 2 and use pin 4.

For negative Enable function, leave pin 4 open and use pin 3.

**Pin Descriptions** 

**+Vin:** The positive input supply for the module with respect to  $-V_{in}$ . If powering the module from a -24V telecom central office supply, this input is connected to the primary system ground.

**-Vin:** The negative input supply for the module, and the 0VDC reference for the EN 1, EN 2, TEMP, and AUX signals. When the module is powered from a +24V supply, this input is connected to the 24V Return.

**EN 1:** The negative logic input that enables the module output. This pin is TTL compatible and referenced to  $-V_{in}$ . A logic '0' at this pin enables the module's outputs. A logic '1' or high impedance disables the module's outputs. If not used, the pin must be connected to  $-V_{in}$ .

**EN 2:** The positive logic input that enables the module output. This pin is TTL compatible and referenced to  $-V_{in}$ . A logic '1' or high impedance enables the module's outputs. If not used, the pin should be left open circuit.

**TEMP:** This pin produces an output signal that tracks the module's metal case temperature. The output voltage is referenced to  $-V_{in}$  and rises approximately  $10 \text{mV/}^{\circ}\text{C}$  from an intital value of 0.1 VDC at -40 °C ( $V_{Temp} = 0.5 + 0.01 \cdot T_{Case}$ ). The signal is available whenever the module is supplied with a valid input

voltage, and is independent of the enable logic status. (Note: A load impedance of less than  $1M\Omega$  will adversly affect the module's over-temperature shutdown threshold. Use a high-impedance input when monitoring this signal.)

**AUX:** Produces a regulated output voltage of  $11.6V \pm 5\%$ , which is referenced to  $-V_{in}$ . The current drawn from the pin must be limited to 10mA. The voltage may be used to indicate the output status of the module to a primary referenced circuit, or power a low-current amplifer.

**Vo<sub>1</sub>:** The higher regulated output voltage, which is referenced to the COM node.

**Vo2:** The lower regulated output voltage, which is referenced to the COM node.

**COM:** The secondary return reference for the module's two regulated output voltages. It is DC isolated from the input supply pins.

**Vo<sub>1</sub> Adjust:** Using a single resistor, this pin allows Vo<sub>1</sub> to be adjusted higher or lower than the preset value. If not used, this pin should be left open circuit.

**Vo<sub>2</sub> Adjust:** Using a single resistor, this pin allows Vo<sub>2</sub> to be adjusted higher or lower than the preset value. If not used, this pin should be left open circuit.

## **Specifications** (Unless otherwise stated, $T_a$ =25°C, $V_{in}$ =24V, & $Io_1$ = $Io_2$ =10A)

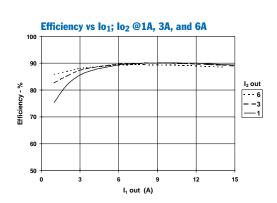
					PT4680 SERIES			
Characteristics	Symbols	Conditions		Min	Тур	Max	Units	
Output Current	Io <sub>1</sub> Io <sub>2</sub>		Vo <sub>1</sub> Vo <sub>2</sub>	0	_	15 15	A	
	$Io_1+Io_2$	Total (both outputs)		0	_	20 (1)	A	
Input Voltage Range	V <sub>in</sub>			18	24	36	V	
Set Point Voltage Tolerance	Votol			_	±1	±2	$%V_{o}$	
Temperature Variation	$\Delta \text{Reg}_{\text{temp}}$	-40 to +100°C Case, Io <sub>1</sub> =Io <sub>2</sub> =0A		_	±0.5	_	$%V_{o}$	
Line Regulation	ΔReg <sub>line</sub>	Over V <sub>in</sub> range with Io <sub>1</sub> =Io <sub>2</sub> =5A		_	±5	±10	mV	
Load Regulation	$\Delta Regload$	1A ≤Io <sub>1</sub> ≤15A, Io <sub>2</sub> =1A 1A ≤Io <sub>2</sub> ≤15A, Io <sub>1</sub> =1A	$\Delta Vo_1 \\ \Delta Vo_2$	_	±2 ±2	±10 ±10	mV	
Cross Regulation	$\Delta Reg_{cross}$	1A ≤I <sub>O2</sub> ≤15A, I <sub>O1</sub> =1A 1A ≤I <sub>O1</sub> ≤15A, I <sub>O2</sub> =1A	$\Delta Vo_1 \\ \Delta Vo_2$	_	±2 ±2	±10 ±5	mV	
Total Output Variation	$\Delta V_{ m o}$ tol	Includes set-point, line load, -40°C to +100°C case	$\Delta Vo_1 \\ \Delta Vo_2$	_	±2 ±2	±3 ±3	$%V_{o}$	
Efficiency	η	Io <sub>1</sub> =1o <sub>2</sub> =10A	PT4681 PT4682 PT4683 PT4685 PT4686 PT4687		88 87 87 86 85 86		%	
V <sub>o</sub> Ripple (pk-pk)	V <sub>r</sub>	Io <sub>1</sub> =Io <sub>2</sub> =5A, 20MHz bandwidth	V <sub>o</sub> =5V V <sub>o</sub> <5V	_	_	75 50	$\mathrm{mV}_{\mathrm{pp}}$	
Transient Response	t <sub>tr</sub>	1A/µs load step from 50% to 100% I <sub>c</sub> (either output)	1A/µs load step from 50% to 100% I <sub>o</sub> max			100	μSec %V <sub>o</sub>	
Current Limit	I <sub>lim</sub>	Each output with other unloaded		15.5	18	_	A	
Output Rise Time	Von	At turn-on to within 90% of Vo		_	5	10	mSec	
Output Over-Voltage Protection	OVP	Either output; shutdown and latch of	f	_	125 (2)	_	$%V_{o}$	
Switching Frequency	$f_s$			280	_	320	kHz	
Under-Voltage-Lockout	UVLO	Rising Falling			17 16	18 —	V	
Internal Input Capacitance	Cin			_	2	_	μF	
On/Off Control Input High Voltage Input Low Voltage	V <sub>IH</sub> V <sub>IL</sub>	Referenced to -Vin		3.5	_	— 0.8 (3)	V	
Input Low Current	${ m I}_{ m IL}$			_	0.5	_	mA	
Quiescent Current	I <sub>in</sub> standby	Pins 2, 3, & 4 connected		_	3	5	mA	
External Output Capacitance	$C_{out}$	Per each output		0	_	5,000	μF	
Primary/Secondary Isolation	V <sub>iso</sub> C <sub>iso</sub> R <sub>iso</sub>			$\frac{1500}{10}$	1500 —		V pF MΩ	
Temperature Sense	$ m V_{temp}$	Output voltage at temperatures:-	−40°C 100°C	_	0.1 (4) 1.5 (4)	_	V	
Over-Temperature Shutdown	OTP	Case temperature (auto restart)		_	110	_	°C	
Operating Temperature Range	Ta	Over V <sub>in</sub> range		-40	_	+85 (5)	°C	
Storage Temperature	T <sub>s</sub>			-40		+125	°C	
Mechanical Shock	_	Per Mil-STD-883D, Method 2002.3		_	500	_	G's	
Mechanical Vibration Per Mil-STD-883D, 20–2,000Hz		Vertical Horizontal			10 (6) 20 (6)		G's	
Weight		_	·	_	90	_	grams	
Flammability		Materials meet UL 94V-0						

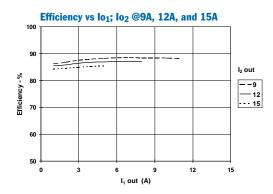
**Notes:** (1) The sum-total current from Vo<sub>1</sub> & Vo<sub>2</sub> must not exceed 20ADC.

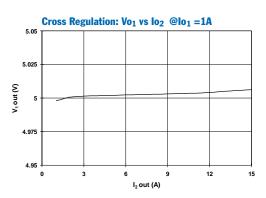
- The sum-value current from vol [2] Volumes to the excess the module's sensitivity to over-voltage detection. For more information, see the application note on output voltage adjustment.
   The EN<sub>1</sub> and EN<sub>2</sub> control inputs (pins 3 & 4) have internal pull-ups and may be controlled with an open-collector (or open-drain) transistor. Both inputs are diode protected and can be connected to +V<sub>in</sub>. The maximum open-circuit voltage is 5.4V.
   Voltage output at "TEMP" pin is defined by the equation:-V<sub>TEMP</sub> = 0.5 + 0.01 T, where T is in °C. See pin descriptions for more information.

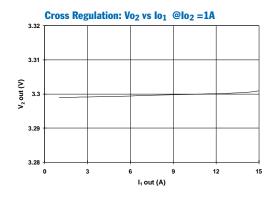
- (5) See SOA curves or consult the factory for the appropriate derating.
  (6) The case pins on the through-holed package types (suffixes N & A) must be soldered. For more information see the applicable package outline drawing.

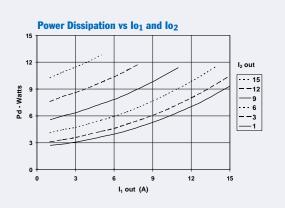
## **PT4681 (Vo<sub>1</sub>/Vo<sub>2</sub> =5.0V/3.3V); V\_{in} =24V (See Notes A & B)**

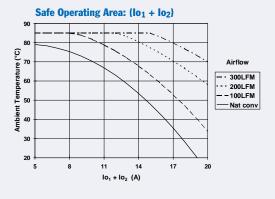






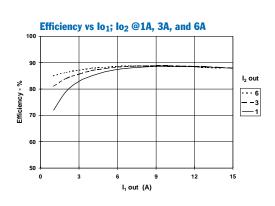


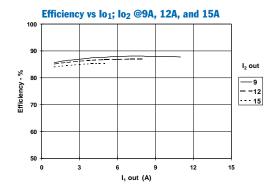


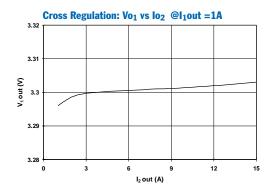


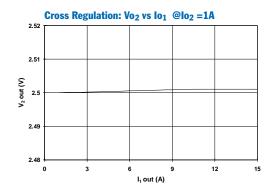
**Note A:** All Characteristic data in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the converter. **Note B:** SOA curves represent operating conditions at which internal components are at or below manufacturer's maximum rated operating temperatures.

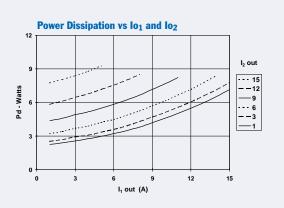
## PT4682 (Vo<sub>1</sub>/Vo<sub>2</sub> =3.3V/2.5V); V<sub>in</sub> =24V (See Notes A & B)

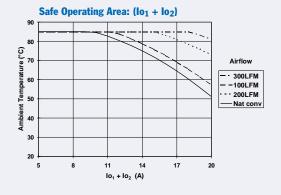


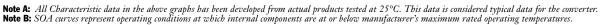




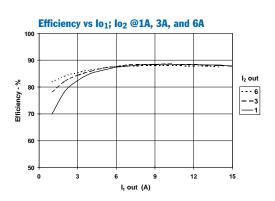




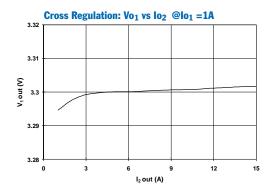


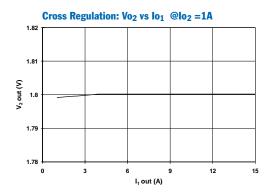


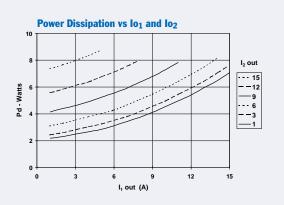
## **PT4683 (Vo<sub>1</sub>/Vo<sub>2</sub> =3.3V/1.8V); V**<sub>in</sub> =24V (See Notes A & B)

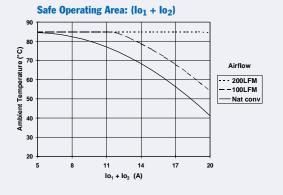






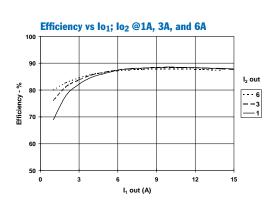


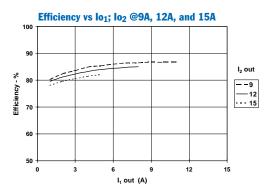


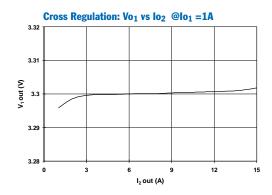


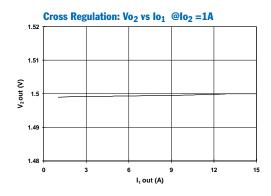
**Note A:** All Characteristic data in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the converter. **Note B:** SOA curves represent operating conditions at which internal components are at or below manufacturer's maximum rated operating temperatures.

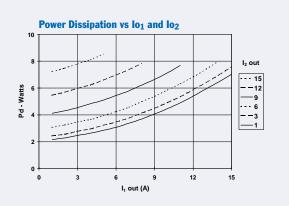
## PT4685 (Vo<sub>1</sub>/Vo<sub>2</sub> =3.3V/1.5V); V<sub>in</sub> =24V

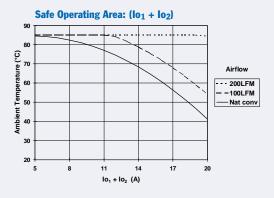






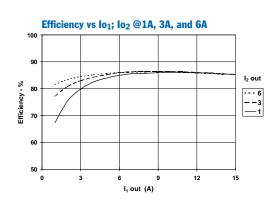


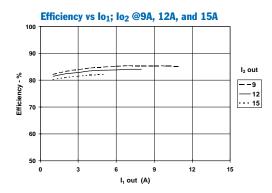


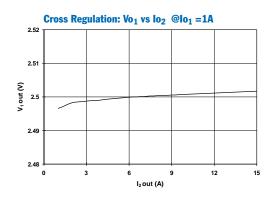


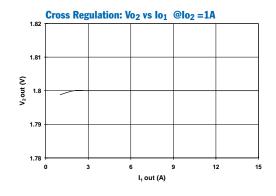
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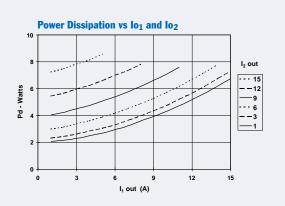
## PT4686 (Vo<sub>1</sub>/Vo<sub>2</sub> =2.5V/1.8V); V<sub>in</sub> =24V

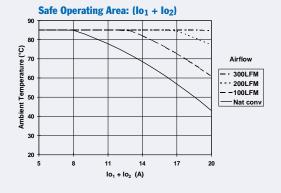






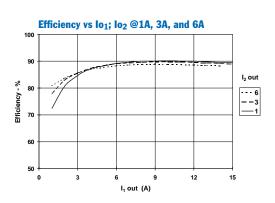


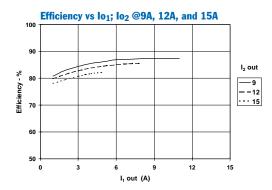


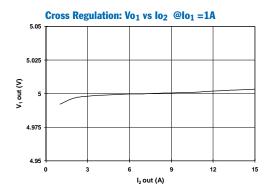


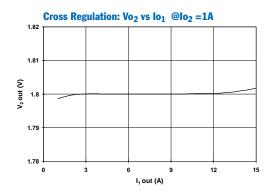
**Note A:** All Characteristic data in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the converter. **Note B:** SOA curves represent operating conditions at which internal components are at or below manufacturer's maximum rated operating temperatures.

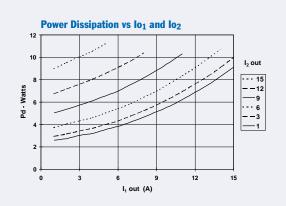
## PT4687 (Vo<sub>1</sub>/Vo<sub>2</sub> =5V/1.8V); V<sub>in</sub> =24V

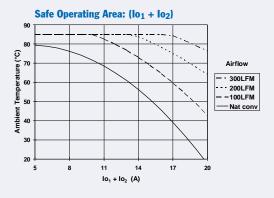












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PT4660 & PT4680 Series

# Operating Features & System Considerations for the PT4660/PT4680 Dual-Output DC/DC Converters

#### **Over-Current Protection**

The dual-outputs of the PT4660 and PT4680 series of DC/DC converters have independent output voltage regulation and current limit control. Applying a load current in excess of the current limit threshold at either output will cause the respective output voltage to drop. However, the voltage at Vo<sub>2</sub> is derived from Vo<sub>1</sub>. Therefore a current limit fault on Vo<sub>1</sub> will also cause Vo<sub>2</sub> to drop. Conversely, a current limit fault applied to Vo<sub>2</sub> will only cause Vo<sub>2</sub> voltage to drop, and Vo<sub>1</sub> will remain in regulation.

The current limit circuitry incorporates a limited amount of foldback. The fault current flowing into an absolute short circuit is therefore slightly less than the current limit threshold. Recovery from a current limit fault is automatic and the converter will not be damaged by a continuous short circuit at either output.

### **Output Over-Voltage Protection**

Each output is monitored for over voltage (OV). For fail safe operation and redundancy, the OV fault detection circuitry uses a separate reference to the voltage regulation circuits. The OV threshold is fixed, and set nominally 25% higher than the set-point output voltage. If either output exceeds the threshold, the converter is shutdown and must be actively reset. The OV protection circuit can be reset by momentarily turning the converter off. This is accomplished by either cycling one of the output enable control pins (EN1 or EN2), or by removing the input power to the converter. Note: If Vo1 or Vo2 is adjusted to a higher voltage, the margin between the respective steady-state output voltage and its OV threshold is reduced. This can make the module sensitive to OV fault detection, that may result from random noise and load transients.

## **Over-Temperature Protection**

The PT4660/80 DC/DC converters have an internal temperature sensor, which monitors the temperature of the module's metal case. If the case temperature exceeds a nominal 115°C the converter will shut down. The converter will automatically restart when the sensed temperature returns to about 100°C. The analog voltage generated by the sensor is also made available at the 'TEMP' output (pin 5), and can be monitored by the host system for diagnostic purposes. Consult the 'Pin Descriptions' section of the data sheet for further information on this feature.

#### **Under-Voltage Lock-Out**

The Under-Voltage Lock-Out (UVLO) circuit prevents operation of the converter whenever the input voltage to the module is insufficient to maintain output regulation. The UVLO has approximately 2V of hysterisis. This is to prevent oscillation with a slowly changing input voltage. Below the UVLO threshold the module is off and the enable control inputs, EN1 and EN2 are inoperative.

### **Primary-Secondary Isolation**

The PT4460/80 series of DC/DC converters incorporate electrical isolation between the input terminals (primary) and the output terminals (secondary). All converters are production tested to a withstand voltage of 1500VDC. The isolation complies with UL60950 and EN60950, and the requirements for operational isolation. This allows the converter to be configured for either a positive or negative input voltage source.

The regulation control circuitry for these modules is located on the secondary (output) side of the isolation barrier. Control signals are passed between the primary and secondary sides of the converter via a proprietory magnetic coupling scheme. This eliminates the use of opto-couplers. The data sheet 'Pin Descriptions' and 'Pin-Out Information' provides guidance as to which reference (primary or secondary) that must be used for each of the external control signals.

#### **Fuse Recommendations**

If desired an input fuse may be added to protect against the application of a reverse input voltage.



## Using the On/Off Enable Controls on the PT4660 and PT4680 Series of DC/DC Converters

The PT4660 (48V input) and PT4680 (24V input) series of 75-W dual-output DC/DC converters incorporates both positive and negative logic *Output Enable* controls. EN1 (pin 3) is the positive enable input, and EN2 (pin 4) is the negative enable input. Both inputs are TTL logic compatible, and are electrically referenced to -V<sub>in</sub> (pin 2) on the primary (input) side of the converter. A pull-up resistor is not required, but may be added if desired. Adding a pull-up resistor from either input, up to +V<sub>in</sub>, will not damage the converter.

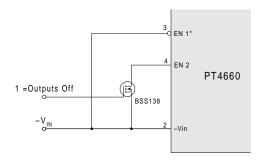
## **Automatic (UVLO) Power-Up**

Connecting EN1 (pin 3) to  $-V_{in}$  (pin 2) and leaving EN2 (pin 4) open-circuit configures the converter for automatic power up. (See data sheet "Typical Application"). The converter control circuitry incorporates an "Under Voltage Lockout" (UVLO) function, which disables the converter until the minimum specified input voltage is present at  $\pm V_{in}$ . (See data sheet Specifications). The UVLO circuitry ensures a clean transition during power-up and power-down, allowing the converter to tolerate a slow-rising input voltage. For most applications EN1 and EN2, can be configured for automatic power-up.

#### **Positive Output Enable (Negative Inhibit)**

To configure the converter for a positive enable function, connect EN1 (pin 3) to  $-V_{\rm in}$  (pin 2), and apply the system On/Off control signal to EN2 (pin 4). In this configuration, a logic '0' ( $-V_{\rm in}$  potential) applied to pin 4 disables the converter outputs. An example of this configuration is detailed in Figure 1.

Figure 1; Positive Enable Configuration

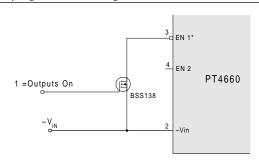


## **Negative Output Enable (Positive Inhibit)**

To configure the converter for a negative enable function, EN2 (pin 4) is left open circuit, and the system On/Off control signal is applied to EN1 (pin 3). A logic '0' (-V<sub>in</sub> potential) must then be applied to pin 3 in order to

enable the outputs of the converter. An example of this configuration is detailed in Figure 2. <u>Note</u>: The converter will only produce and output voltage if a valid input voltage is applied to  $\pm V_{in}$ .

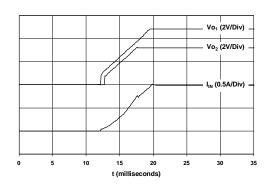
Figure 2; Negative Enable Configuration



## **On/Off Output Voltage Sequencing**

The output voltages from the PT4660 series of DC/DC converters are independantly regulated, and are internally sequenced to meet the power-up requirements of popular microprocessor and DSP chipsets. Figure 3 shows the waveforms from a PT4661 after the converter is enabled at t=0s. During power-up, the Vo1 and Vo2 voltage waveforms typically track within 0.4V prior to Vo2 reaching regulation. The waveforms were measured with a 5-Adc resistive load at each output, and with a 48-VDC input source applied. The converter typically produces a fully regulated output within 25ms. The actual turn-on time will vary slightly with input voltage, but the power-up sequence is independent of the load at either output.

Figure 3; Vo<sub>1</sub>, Vo<sub>2</sub> Power-Up Sequence



During turn-off, both outputs drop rapidly due to the discharging effect of actively switched rectifiers. The voltage at  $Vo_1$  remains higher than  $Vo_2$  during this period. The discharge time is typically 100 $\mu$ s, but will vary with the amount of external load capacitance.

#### PT4660 & PT4680 Series

# Adjusting the Output Voltage of the PT4660 and PT4680 Dual Output Voltage DC/DC Converters

The output voltages  $Vo_1$  and  $Vo_2$  from the PT4680 (24V Bus) and PT4660 (48V Bus) series of DC/DC converters can be independantly adjusted higher or lower than the factory trimmed pre-set voltage by up to  $\pm 10\%$ . The adjustment requires the addition of a single external resistor<sup>1</sup>. Table 1 gives the adjustment range of  $Vo_1$  and  $Vo_2$  for each model in the series as  $V_a$ (min) and  $V_a$ (max).

**Vo<sub>1</sub> Adjust Down:** Add a resistor  $(R_1)$ , between pin 13  $(V_1 \text{ Adj})$  and pin 12  $(V_0)^2$ .

**Vo<sub>1</sub> Adjust Up:** To increase the output, add a resistor  $R_2$  between pin 13 ( $V_1$  Adj) and pin 14 (COM) <sup>2</sup>, <sup>4</sup>.

**Vo<sub>2</sub> Adjust Down:** Add a resistor (R<sub>3</sub>) between pin 20 (V<sub>2</sub> Adj) and pin 21 (Vo<sub>2</sub>) <sup>2</sup>.

**Vo<sub>2</sub> Adjust Up:** Add a resistor R<sub>4</sub> between pin 20 (V<sub>2</sub> Adj) and pins 19 (COM) <sup>2</sup>, <sup>4</sup>.

Refer to Figure 1 and Table 2 for both the placement and value of the required resistor.

## Notes:

- 1. Adjust resistors are not required if  $Vo_1$  and  $Vo_2$  are to remain at their respective nominal set-point voltage. In this case,  $V_1$  Adj (pin 13) and  $V_2$  Adj (pin 20) are left open-circuit
- 2. Use only a single 1% resistor in either the (R<sub>1</sub>) or R<sub>2</sub> location to adjust Vo<sub>1</sub>, and in the (R<sub>3</sub>) or R<sub>4</sub> location to adjust Vo<sub>2</sub>. Place the resistor as close to the DC/DC/DC converter as possible.

- 3. Vo<sub>2</sub> must always be at least 0.3V lower than Vo<sub>1</sub>.
- 4. The over-voltage protection threshold is fixed, and is set nominally 25% above the set-point output voltage. Adjusting Vo<sub>1</sub> or Vo<sub>2</sub> higher will reduce the voltage margin between the respective steady-state output voltage and its over-voltage (OV) protection threshold. This could make the module sensitive to OV fault detection, as a result of random noise and load transients.

<u>Note</u>: An OV fault is a latched condition that shuts down both outputs of the converter. The fault can only be cleared by cycling one of the Enable control pins  $(EN_1^* / EN_2)$ , or by momentarily removing the input power to the module.

 Never connect capacitors to either the Vo<sub>1</sub> Adjust or Vo<sub>2</sub> Adjust pins. Any capacitance added to these control pins will affect the stability of the respective regulated output.

The adjust up and adjust down resistor values can also be calculated using the following formulas. Be sure to select the correct formula parameter from Table 1 for the output and model being adjusted.

$$(R_1) \text{ or } (R_3) = \frac{K_0 (V_a - V_r)}{V_r (V_0 - V_a)} - R_s \quad k\Omega$$

$$R_2 \text{ or } R_4 = \frac{K_0}{V_0 - V_0} - R_s \quad k\Omega$$

Where:  $V_0$  = Original output voltage, ( $V_{01}$  or  $V_{02}$ )

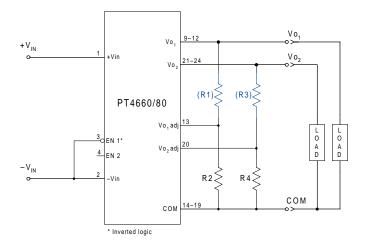
V<sub>a</sub> = Adjusted output voltage

 $V_r$  = The reference voltage from Table 1

K<sub>o</sub> = The multiplier constant in Table 1

 $R_s$  = The series resistance from Table 1

Figure 1



## PT4660 & PT4680 Series

**Table 1; ADJUSTMENT RANGE AND FORMULA PARAMETERS** 

Vo <sub>1</sub> Bus					Vo <sub>2</sub> Bus (2)				
24V Bus Pt.#	PT4681/7	PT4682/3/5	PT4688	PT4686	PT4681	PT4682	PT4683/6/7	PT4685	PT4688
48V Bus Pt.#	PT4661/7	PT4662/3/5	PT4668	PT4666	PT4661	PT4662	PT4663/6/7	PT4665	PT4668
Adj. Resistor	(R1)/R2	(R1)/R2	(R1)/R2	(R1)/R2	(R3)/R4	(R3)/R4	(R3)/R4	(R3)/R4	(R3)/R4
V <sub>o</sub> (nom)	5.0V	3.3V	3.3V	2.5V	3.3V	2.5V	1.8V	1.5V	1.2V
Va(min)	4.5V	2.97V	2.97V	2.25V	2.97V	2.25V	1.62V	1.35	1.08
Va(max)	5.5V	3.63V	3.63V	2.75V	3.63V	2.75V	1.98V	1.65	1.32
Vr	2.5V	1.65V	2.5V	1.25	1.5V	1.5V	1.5V	TBD	0.6V
K <sub>o</sub> (V·kΩ)	1.248	8.234	10.96	6.24	1.8	2.0	1.9	TBD	0.726
R <sub>s</sub> (kΩ)	20.0	20.0	4.99	20.0	4.99	3.32	3.32	TBD	4.22

Table 2A; ADJUSTMENT RESISTOR VALUES, Vo1

24V Bus Pt.#	PT4681/7		PT4682/3/5	PT4688		PT4686
48V Bus Pt.#	PT4661/7		PT4662/3/5	PT4668		PT4666
Adj. Resistor	(R1)/R2		(R1)/R2	(R1)/R2		(R1)/R2
Vo(nom)	5.0V		3.3V	3.3V		2.5V
V <sub>a</sub> (req'd)		Va(req'd)			$V_a(req'd)$	
5.5	5.0kΩ	3.6	$7.4 \mathrm{k}\Omega$	31.5kΩ	2.75	5.0kΩ
5.4	11.2kΩ	3.54	14.3kΩ	40.7kΩ	2.7	11.2kΩ
5.3	21.6kΩ	3.48	$25.7 \mathrm{k}\Omega$	55.9kΩ	2.65	21.6kΩ
5.2	42.4kΩ	3.42	48.6kΩ	86.3kΩ	2.6	42.4kΩ
5.1	105.0kΩ	3.36	$117.0 \mathrm{k}\Omega$	$178.0 \mathrm{k}\Omega$	2.55	105.0kΩ
5.0		3.3			2.5	
4.9	(99.8)kΩ	3.24	(112.0kΩ)	(49.1kΩ)	2.45	(99.8kΩ)
4.8	(37.4)kΩ	3.18	$(43.6k\Omega)$	$(19.9k\Omega)$	2.4	(37.4kΩ)
4.7	(16.6)kΩ	3.12	$(20.8k\Omega)$	$(10.1k\Omega)$	2.35	(16.6kΩ)
4.6	$(6.2)$ k $\Omega$	3.06	$(9.3k\Omega)$	(5.2kΩ)	2.3	(6.2kΩ)
4.5	(0.0)	3.0	(2.5kΩ)	(2.3kΩ)	2.25	$(0.0k\Omega)$

 $R_1/R_3 = (Blue), R_2/R_4 = Black$ 

Table 2B; ADJUSTMENT RESISTOR VALUES, Vo2

24V Bus Pt.#	PT4681	PT4682		PT4683/6/7	PT4685	PT4688
48V Bus Pt.#	PT4661	PT4662		PT4663/6/7	PT4665	PT4668
Adj. Resistor	(R3)/R4	(R3)/R4		(R3)/R4	(R3)/R4	(R3)/R4
V <sub>o</sub> (nom)	3.3V	2.5V		1.8V	1.5V	1.2V
V <sub>a</sub> (req'd)			V <sub>a</sub> (req'd)			
3.6	$1.0$ k $\Omega$		1.95	9.4kΩ		
3.54	2.5kΩ		1.9	$15.7 \mathrm{k}\Omega$		
3.48	$5.0$ k $\Omega$		1.85	34.7kΩ		
3.42	$10.0$ k $\Omega$		1.8			
3.36	$25.0$ k $\Omega$		1.75	$(3.0)$ k $\Omega$		
3.3			1.7			
3.24	$(29.8)$ k $\Omega$		1.65		TBD	
3.18	$(11.8)$ k $\Omega$		1.6		TBD	
3.12	$(5.8)$ k $\Omega$		1.55		TBD	
3.06	$(2.8)$ k $\Omega$		1.5			
3.0	$(1.0)$ k $\Omega$		1.45		(TBD)	
2.75		$4.7 \mathrm{k}\Omega$	1.4		(TBD)	
2.7		$6.7 \mathrm{k}\Omega$	1.35		(TBD)	
2.65		$10.0 \mathrm{k}\Omega$	1.3			$3.0 \mathrm{k}\Omega$
2.6		$16.7 \mathrm{k}\Omega$	1.275			$5.5 \mathrm{k}\Omega$
2.55		$36.7 \mathrm{k}\Omega$	1.25			10.3kΩ
2.5			1.225			$24.8 \mathrm{k}\Omega$
2.45		$(22.0)$ k $\Omega$	1.2			
2.4		$(8.7)$ k $\Omega$	1.175			$(23.6)$ k $\Omega$
2.35		$(4.2)$ k $\Omega$	1.15			$(9.1)$ k $\Omega$
2.3		$(2.0)$ k $\Omega$	1.125			$(4.3)$ k $\Omega$
2.25		$(0.7)$ k $\Omega$	1.1			$(1.8)$ k $\Omega$

 $R_1/R_3 = (Blue), R_2/R_4 = Black$ 



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