March 2000



LM2765 Switched Capacitor Voltage Converter General Description Features

The LM2765 CMOS charge-pump voltage converter operates as a voltage doubler for an input voltage in the range of +1.8V to +5.5V. Two low cost capacitors and a diode are used in this circuit to provide up to 20 mA of output current.

The LM2765 operates at 50 kHz switching frequency to reduce output resistance and voltage ripple. With an operating current of only 130 μ A (operating efficiency greater than 90% with most loads) and 0.1 μ A typical shutdown current, the LM2765 provides ideal performance for battery powered systems. The device is manufactured in a SOT-23-6 package.

- Doubles Input Supply Voltage
- SOT23-6 Package
- 20Ω Typical Output Impedance
- 90% Typical Conversion Efficiency at 20 mA
- 0.1µA Typical Shutdown Current

Applications

- Cellular Phones
- Pagers
- PDAs
- Operational Amplifier Power Supplies
- Interface Power Supplies
- Handheld Instruments

Basic Application Circuits Voltage Doubler D, V_{IN} (1.8V to 5.5V) V_{OUT} V+ $V_{OUT} = 2V_{IN}$ LM2765 C_2 CAP+ SD CAP-GND DS101281-1 **Connection Diagram** 6-Lead SOT (M6) 1 ഗ DS101281-22 ы С Actual Size П DS101281-13 **Top View With Package Marking Ordering Information** Order Number Packado Supplied as Packago

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		Number	Marking	
	LM2765M6	MA06A	S15B (Note 1)	Tape and Reel (1000 units/reel)
	LM2765M6X	MA06A	S15B (Note 1)	Tape and Reel (3000 units/reel)
1				

Note 1: The small physical size of the SOT-23 package does not allow for the full part number marking. Devices will be marked with the designation shown in the column Package Marking.

LM2765

Pin Description

Pin	Name	Function	
1	V+	Power supply positive voltage input.	
2	GND	Power supply ground input.	
3	CAP-	Connect this pin to the negative terminal of the charge-pump capacitor.	
4	SD	Shutdown control pin, tie this pin to ground in normal operation.	
5	V _{OUT}	Positive voltage output.	
6	CAP+	Connect this pin to the positive terminal of the charge-pump capacitor.	

Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage (V+ to GND, or \	/+ to V _{OUT})	5.8V
SD	(GND – 0.3V) to (V+ +
		0.3V)
V _{OUT} Continuous Output Current	i i	40 mA
Output Short-Circuit Duration to	GND (Note 3)	1 sec.
Continuous Power		600 mW
Dissipation ($T_A = 25^{\circ}C$)(Note 4)		
T _{JMax} (Note 4)		150°C

Operating Ratings

θ _{JA} (Note 4)	210°C/W
Junction Temperature Range	–40° to 100°C
Ambient Temperature Range	–40° to 85°C
Storage Temperature Range	–65°C to 150°C
Lead Temp. (Soldering, 10 seconds)	240°C
ESD Rating (Note 5) Human body model Machine model	2kV 200V

Electrical Characteristics

Limits in standard typeface are for $T_J = 25^{\circ}C$, and limits in **boldface** type apply over the full operating temperature range. Unless otherwise specified: V+ = 5V, $C_1 = C_2 = 3.3 \ \mu$ F. (Note 6)

Symbol	Parameter	Condition	Min	Тур	Max	Units	
V+	Supply Voltage		1.8		5.5	V	
IQ	Supply Current	No Load		130	450	μΑ	
I _{SD}	Shutdown Supply Current			0.1	0.5		
		$T_A = 85^{\circ}C$		0.2		- μΑ	
V _{SD}	Shutdown Pin Input Voltage	Shutdown Mode	2.0			V	
		Normal Operation			0.6	7 °	
IL	Output Current	$2.5V \le V_{IN} \le 5.5V$	20				
		$1.8V \le V_{IN} < 2.5V$	10			— mA	
R _{OUT}	Output Resistance (Note 7)	$I_L = 20 \text{ mA}$		20	40	Ω	
f _{osc}	Oscillator Frequency	(Note 8)	40	100	200	kHz	
f _{SW}	Switching Frequency	(Note 8)	20	50	100	kHz	
P _{EFF}	Power Efficiency	$I_L = 20 \text{ mA to GND}$		92		%	
VOEFF	Voltage Conversion Efficiency	No Load		99.96		%	

Note 2: Absolute maximum ratings indicate limits beyond which damage to the device may occur. Electrical specifications do not apply when operating the device beyond its rated operating conditions.

Note 3: V_{OUT} may be shorted to GND for one second without damage. However, shorting V_{OUT} to V+ may damage the device and should be avoided. Also, for temperatures above 85°C, OUT must not be shorted to GND or V+, or device may be damaged.

Note 4: The maximum allowable power dissipation is calculated by using $P_{DMax} = (T_{JMax} - T_A)/\theta_{JA}$, where T_{JMax} is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction-to-ambient thermal resistance of the specified package.

Note 5: The human body model is a 100pF capacitor discharged through a 1.5kΩ resistor into each pin. The machine model is a 200pF capacitor discharged directly into each pin.

Note 6: In the test circuit, capacitors C_1 and C_2 are 3.3 μ F, 0.3 Ω maximum ESR capacitors. Capacitors with higher ESR will increase output resistance, reduce output voltage and efficiency.

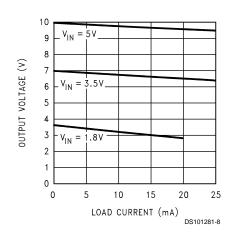
Note 7: Specified output resistance includes internal switch resistance and capacitor ESR. See the details in the application information for positive voltage doubler.

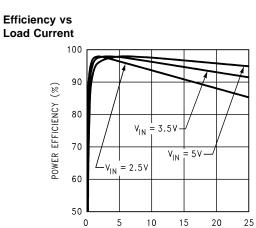
Note 8: The output switches operate at one half of the oscillator frequency, f_{OSC} = $2f_{SW}.$

LM2765 **Test Circuit** D₁ $V_{\rm IN}$ V+ VOUT V_{OUT} LM2765 6 SD CAP+ CIN ROUT CAP-GND * $\rm C_{IN},~C_1$, and $\rm C_2$ are 3.3 $\mu\rm F$ OS-CON capacitors. DS101281-3 FIGURE 1. LM2765 Test Circuit **Typical Performance Characteristics** (Circuit of Figure 1, $V_{IN} = 5V$, $T_A = 25^{\circ}C$ unless otherwise specified) Supply Current vs **Output Resistance vs** Supply Voltage Capacitance 200 140 180 120 160 SUPPLY CURRENT (μ A) V_{IN} = 1.8V OUTPUT RESISTANCE (U) 140 100 120 80 100 $V_{IN} = 3.5V$ 60 80 60 40 40 20 20 Źγ¦ = 5 V 0 0 2 2.5 3 3.5 4 4.5 5 1.5 5.5 20 25 0 5 10 15 30 35 SUPPLY VOLTAGE (V) CAPACITANCE (μ F) DS101281-4 DS101281-5 Output Resistance vs **Output Resistance vs** Supply Voltage Temperature 40 60 35 50 OUTPUT RESISTANCE (Ω) OUTPUT RESISTANCE (Ω) 30 V_{IN} = 1.8V 40 25 20 30 15 20 10 V_{IN} 5٧ 10 5 0 0 1.5 2 2.5 3 3.5 4 4.5 5 5.5 -25 -50 0 25 50 75 100 SUPPLY VOLTAGE (V) TEMPERATURE (°C) DS101281-6 DS101281-7

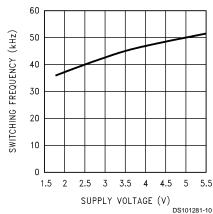
Typical Performance Characteristics (Circuit of Figure 1, $V_{IN} = 5V$, $T_A = 25^{\circ}C$ unless otherwise specified) (Continued)

Output Voltage vs Load Current

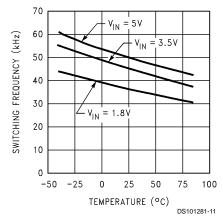




Switching Frequency vs Supply Voltage



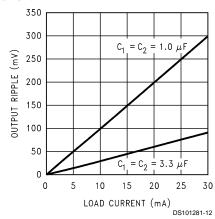
Switching Frequency vs Temperature



LOAD CURRENT (mA)

DS101281-9

Output Ripple vs Load Current



Circuit Description

The LM2765 contains four large CMOS switches which are switched in a sequence to double the input supply voltage. Energy transfer and storage are provided by external capacitors. Figure 2 illustrates the voltage conversion scheme. When S_2 and S_4 are closed, C_1 charges to the supply voltage V+. During this time interval, switches S_1 and S_3 are open. In the next time interval, S_2 and S_4 are open; at the same time, S_1 and S_3 are closed, the sum of the input voltage V+ and the voltage across C_1 gives the 2V+ output voltage when there is no load. The output voltage drop when a load is added is determined by the parasitic resistance ($R_{ds(on)}$ of the MOSFET switches and the ESR of the capacitors) and the charge transfer loss between capacitors. Details will be discussed in the following application information section.

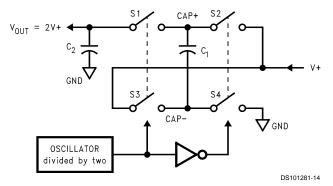


FIGURE 2. Voltage Doubling Principle

Application Information

Positive Voltage Doubler

The main application of the LM2765 is to double the input voltage. The range of the input supply voltage is 1.8V to 5.5V.

The output characteristics of this circuit can be approximated by an ideal voltage source in series with a resistance. The voltage source equals 2V+. The output resistance R_{out} is a function of the ON resistance of the internal MOSFET switches, the oscillator frequency, and the capacitance and ESR of C₁ and C₂. Since the switching current charging and discharging C₁ is approximately twice as the output current, the effect of the ESR of the pumping capacitor C₁ will be multiplied by four in the output resistance. The output capacitor C₂ is charging and discharging at a current approximately equal to the output current, therefore, its ESR only counts once in the output resistance. A good approximation of R_{out} is:

$$R_{OUT} \simeq 2R_{SW} + \frac{2}{f_{OSC} \times C_1} + 4ESR_{C1} + ESR_{C2}$$

where R_{SW} is the sum of the ON resistance of the internal MOSFET switches shown in Figure 2. R_{SW} is typically 8Ω for the LM2765.

The peak-to-peak output voltage ripple is determined by the oscillator frequency as well as the capacitance and ESR of the output capacitor C_2 :

$$V_{RIPPLE} = \frac{I_{L}}{f_{OSC} \times C_{2}} + 2 \times I_{L} \times ESR_{C2}$$

High capacitance, low ESR capacitors can reduce both the output resistance and the voltage ripple.

The Schottky diode D_1 is only needed to protect the device from turning-on its own parasitic diode and potentially latching-up. During start-up, D_1 will also quickly charge up the output capacitor to $V_{\rm IN}$ minus the diode drop thereby decreasing the start-up time. Therefore, the Schottky diode D_1 should have enough current carrying capability to charge the output capacitor at start-up, as well as a low forward voltage to prevent the internal parasitic diode from turning-on. A Schottky diode like 1N5817 can be used for most applications. If the input voltage ramp is less than 10V/ms, a smaller Schottky diode like MBR0520LT1 can be used to reduce the circuit size.

Shutdown Mode

A shutdown (SD) pin is available to disable the device and reduce the quiescent current to 0.1 μ A. In normal operating mode, the SD pin is connected to ground. The device can be brought into the shutdown mode by applying to the SD pin a voltage greater than 40% of the V+ pin voltage.

Capacitor Selection

As discussed in the *Positive Voltage Doubler* section, the output resistance and ripple voltage are dependent on the capacitance and ESR values of the external capacitors. The output voltage drop is the load current times the output resistance, and the power efficiency is

$$\eta = \frac{P_{OUT}}{P_{IN}} = \frac{I_{L}^{2} R_{L}}{I_{I}^{2} R_{I} + I_{I}^{2} R_{OUT} + I_{O} (V+)}$$

Where $I_Q(V+)$ is the quiescent power loss of the IC device, and $I_L^2 R_{out}$ is the conversion loss associated with the switch on-resistance, the two external capacitors and their ESRs.

The selection of capacitors is based on the specifications of the dropout voltage (which equals $I_{out} R_{out}$), the output voltage ripple, and the converter efficiency. Low ESR capacitors (*Table 1*) are recommended to maximize efficiency, reduce the output voltage drop and voltage ripple.

Phone	Website	Capacitor Type
(847)-843-7500	www.nichicon.com	PL & PF series, through-hole aluminum
		electrolytic
(843)-448-9411	www.avxcorp.com	TPS series, surface-mount tantalum
(207)-324-4140	www.vishay.com	593D, 594D, 595D series, surface-mount tantalum
(619)-661-6835	www.sanyovideo.com	OS-CON series, through-hole aluminum
		electrolytic
-	(847)-843-7500 (843)-448-9411 (207)-324-4140	(847)-843-7500 www.nichicon.com (843)-448-9411 www.avxcorp.com (207)-324-4140 www.vishay.com

TABLE 1. Low ESR Capacitor Manufacturers

Application Information (Continued)

TABLE 1. Low ESR Capacitor Manufacturers (Continued)

Manufacturer	Phone	Website	Capacitor Type
Murata	(800)-831-9172	www.murata.com	Ceramic chip capacitors
Taiyo Yuden	(800)-348-2496	www.t-yuden.com	Ceramic chip capacitors
Tokin	(408)-432-8020	www.tokin.com	Ceramic chip capacitors

Other Applications

Paralleling Devices

Any number of LM2765s can be paralleled to reduce the output resistance. Each device must have its own pumping capacitor C_1 , while only one output capacitor C_{out} is needed as shown in Figure 3. The composite output resistance is:

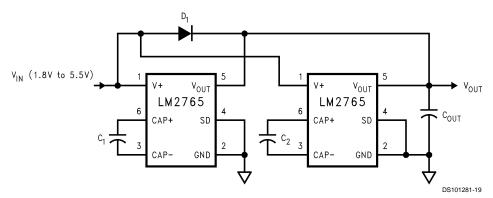


FIGURE 3. Lowering Output Resistance by Paralleling Devices

Cascading Devices

Cascading the LM2765s is an easy way to produce a greater voltage (A two-stage cascade circuit is shown in Figure 4). The effective output resistance is equal to the weighted sum of each individual device:

$$R_{out} = 1.5R_{out_1} + R_{out_2}$$

 $R_{OUT} = \frac{R_{OUT} \text{ of each LM2765}}{\text{Number of Devices}}$

Note that increasing the number of cascading stages is practically limited since it significantly reduces the efficiency, increases the output resistance and output voltage ripple.

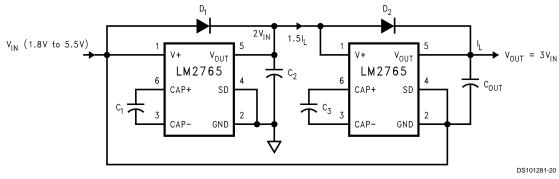


FIGURE 4. Increasing Output Voltage by Cascading Devices

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LM2765

Other Applications (Continued)

Regulating VOUT

It is possible to regulate the output of the LM2765 by use of a low dropout regulator (such as LP2980-5.0). The whole converter is depicted in Figure 5.

A different output voltage is possible by use of LP2980-3.3, LP2980-3.0, or LP2980-adj.

Note that the following conditions must be satisfied simultaneously for worst case design:

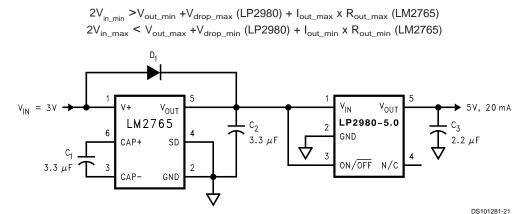
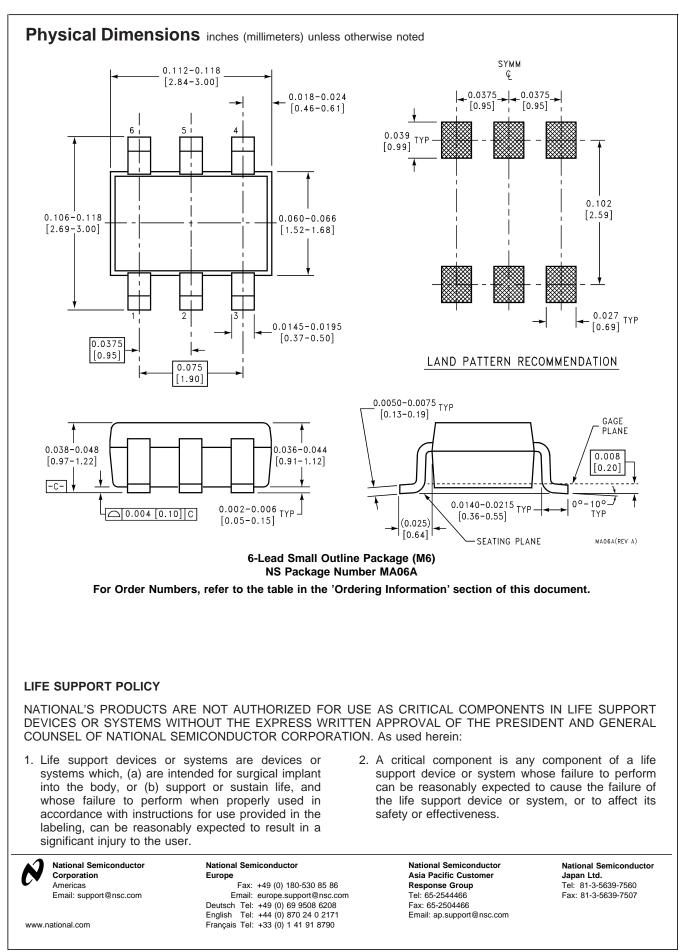


FIGURE 5. Generate a Regulated +5V from +3V Input Voltage



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