

# Programmable Quad Supply Monitor with Adjustable Reset and Watchdog Timers

## FEATURES

- Simultaneously Monitors Four Supplies
- 16 User Selectable Combinations of 5V, 3.3V, 3V, 2.5V, 1.8V, 1.5V and/or  $\pm$ Adjustable Voltage Thresholds
- Guaranteed Threshold Accuracy:  $\pm 1.5\%$  of Monitored Voltage Over Temperature
- Low Supply Current: 43 $\mu$ A Typ
- Adjustable Reset Time
- Adjustable Watchdog Time
- Open-Drain  $\overline{\text{RST}}$  Output (LTC2901-1)
- Push-Pull  $\overline{\text{RST}}$  Output (LTC2901-2)
- Individual Nondelayed Monitor Output for Each Supply
- Power Supply Glitch Immunity
- Guaranteed  $\overline{\text{RST}}$  for  $V_{\text{CC}} \geq 1\text{V}$

## APPLICATIONS


- Desktop and Notebook Computers
- Multivoltage Systems
- Telecom Equipment
- Portable Battery-Powered Equipment
- Network Servers

## DESCRIPTION

The LTC<sup>®</sup>2901 is a programmable supply monitor for systems with up to four supply voltages. One of 16 preset or adjustable voltage monitor combinations can be selected using an external resistor divider connected to the program pin. The preset voltage thresholds are accurate to  $\pm 1.5\%$  over temperature. All four voltage comparator outputs are connected to separate pins for individual supply monitoring.

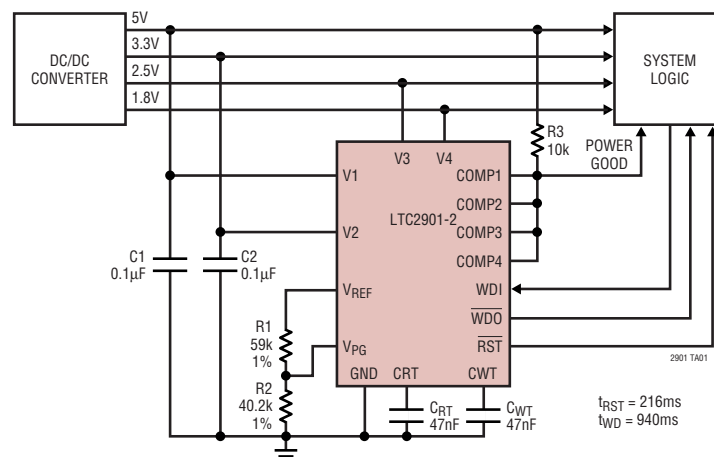
The reset and watchdog delay times are adjustable using external capacitors. Tight voltage threshold accuracy and glitch immunity ensure reliable reset operation without false triggering. The  $\overline{\text{RST}}$  output is guaranteed to be in the correct state for  $V_{\text{CC}}$  down to 1V. The LTC2901-1 features an open-drain  $\overline{\text{RST}}$  output, while the LTC2901-2 has a push-pull  $\overline{\text{RST}}$  output.

The 43 $\mu$ A supply current makes the LTC2901 ideal for power conscious systems and it may be configured to monitor less than four inputs. The parts are available in the 16-lead narrow SSOP package.

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## TYPICAL APPLICATION

Quadruple Supply Monitor (5V, 3.3V, 2.5V, 1.8V)





# LTC2901

## ABSOLUTE MAXIMUM RATINGS

(Notes 1, 2, 3)

V1, V2, V3, V4, V <sub>PG</sub>	–0.3V to 7V
RST (LTC2901-1)	–0.3V to 7V
RST (LTC2901-2)	–0.3V to (V2 + 0.3V)
COMPX	–0.3V to 7V
CWT, WDI, WDO	–0.3V to 7V
V <sub>REF</sub> , CRT	–0.3V to (V <sub>CC</sub> + 0.3V)
Reference Load Current (I <sub>VREF</sub> )	±1mA
V4 Input Current (–ADJ Mode)	–1mA
Operating Temperature Range	
LTC2901-1C/LTC2901-2C	0°C to 70°C
LTC2901-1I/LTC2901-2I	–40°C to 85°C
Storage Temperature Range	–65°C to 150°C
Lead Temperature (Soldering, 10 sec)	300°C

## PACKAGE/ORDER INFORMATION

	ORDER PART NUMBER	
	LTC2901-1CGN LTC2901-2CGN LTC2901-1IGN LTC2901-2IGN	
	GN16 PART MARKING	
	29011 29012 29011I 29012I	

Consult LTC Marketing for parts specified with wider operating temperature ranges.

## ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T<sub>A</sub> = 25°C. V<sub>CC</sub> = 5V, unless otherwise noted. (Note 3)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V <sub>RT50</sub>	5V, 5% Reset Threshold	V1 Input Threshold ●	4.600	4.675	4.750	V
V <sub>RT33</sub>	3.3V, 5% Reset Threshold	V1, V2 Input Threshold ●	3.036	3.086	3.135	V
V <sub>RT30</sub>	3V, 5% Reset Threshold	V2 Input Threshold ●	2.760	2.805	2.850	V
V <sub>RT25</sub>	2.5V, 5% Reset Threshold	V2, V3 Input Threshold ●	2.300	2.338	2.375	V
V <sub>RT18</sub>	1.8V, 5% Reset Threshold	V3, V4 Input Threshold ●	1.656	1.683	1.710	V
V <sub>RT15</sub>	1.5V, 5% Reset Threshold	V3, V4 Input Threshold ●	1.380	1.403	1.425	V
V <sub>RTA</sub>	ADJ Reset Threshold	V3, V4 Input Threshold ●	0.492	0.500	0.508	V
V <sub>RTAN</sub>	–ADJ Reset Threshold	V4 Input Threshold ●	–18	0	18	mV
V <sub>CC</sub>	Minimum Internal Operating Voltage	RST, COMPX in Correct Logic State; V <sub>CC</sub> Rising Prior to Program ●			1	V
V <sub>CCMINP</sub>	Minimum Required for Programming	V <sub>CC</sub> Rising ●			2.42	V
V <sub>CCMINC</sub>	Minimum Required for Comparators	V <sub>CC</sub> Falling ●			2.32	V
V <sub>REF</sub>	Reference Voltage	V <sub>CC</sub> ≥ 2.3V, I <sub>VREF</sub> = ±1mA, C <sub>REF</sub> ≤ 1000pF ●	1.192	1.210	1.228	V
V <sub>PG</sub>	Programming Voltage Range	V <sub>CC</sub> ≥ V <sub>CCMINP</sub> ●	0		V <sub>REF</sub>	V
I <sub>VPG</sub>	V <sub>PG</sub> Input Current	V <sub>PG</sub> = V <sub>REF</sub> ●			±20	nA
I <sub>V1</sub>	V1 Input Current	V1 = 5V, I <sub>VREF</sub> = 12μA, (Note 4) ●		43	75	μA
I <sub>V2</sub>	V2 Input Current	V2 = 3.3V ●		0.8	2	μA
I <sub>V3</sub>	V3 Input Current	V3 = 2.5V ●		0.52	1.2	μA
		V3 = 0.55V (ADJ Mode) ●	–15		15	nA
I <sub>V4</sub>	V4 Input Current	V4 = 1.8V ●		0.34	0.8	μA
		V4 = 0.55V (ADJ Mode) ●	–15		15	nA
		V4 = –0.05V (–ADJ Mode) ●	–15		15	nA
I <sub>CRT(UP)</sub>	CRT Pull-Up Current	V <sub>CRT</sub> = 0V ●	–1.4	–2	–2.6	μA
I <sub>CRT(DN)</sub>	CRT Pull-Down Current	V <sub>CRT</sub> = 1.3V ●	10	20	30	μA
t <sub>RST</sub>	Reset Time-Out Period	C <sub>RT</sub> = 1500pF ●	5	7	9	ms

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**ELECTRICAL CHARACTERISTICS**

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ .  $V_{CC} = 5\text{V}$ , unless otherwise noted. (Note 3)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
$t_{UV}$	$V_X$ Undervoltage Detect to $\overline{\text{RST}}$ or COMPX	$V_X$ Less Than Reset Threshold $V_{\text{RTX}}$ by More Than 1%			150		$\mu\text{s}$
$V_{OL}$	Output Voltage Low $\overline{\text{RST}}$ , COMPX	$I_{\text{SINK}} = 2.5\text{mA}$ ; $V_1 = 3\text{V}$ , $V_2 = 3\text{V}$ ; $V_3, V_4 = 0\text{V}$ ; $V_{\text{PG}} = 0\text{V}$	●		0.15	0.4	V
		$I_{\text{SINK}} = 100\mu\text{A}$ ; $V_2 = 1\text{V}$ ; $V_1, V_3, V_4 = 0\text{V}$	●		0.05	0.3	V
		$I_{\text{SINK}} = 100\mu\text{A}$ ; $V_1 = 1\text{V}$ ; $V_2, V_3, V_4 = 0\text{V}$	●		0.05	0.3	V
$V_{OH}$	Output Voltage High $\overline{\text{RST}}$ , $\overline{\text{WDO}}$ , COMPX (Note 5)	$I_{\text{SOURCE}} = 1\mu\text{A}$	●	$V_2 - 1$			V
$V_{OL}$	Output Voltage Low $\overline{\text{WDO}}$	$I_{\text{SINK}} = 2.5\text{mA}$ ; $V_1 = 5\text{V}$ , $V_2 = 3.3\text{V}$ ; $V_3, V_4 = 1\text{V}$ ; $V_{\text{PG}} = 0\text{V}$	●		0.15	0.4	V
$V_{OH}$	Output Voltage High $\overline{\text{RST}}$ (LTC2901-2) (Note 6)	$I_{\text{SOURCE}} = 200\mu\text{A}$	●	$0.8 \cdot V_2$			V
$I_{\text{CWT(UP)}}$	CWT Pull-Up Current	$V_{\text{CWT}} = 0\text{V}$	●	-1.4	-2	-2.6	$\mu\text{A}$
$I_{\text{CWT(DN)}}$	CWT Pull-Down Current	$V_{\text{CWT}} = 1.3\text{V}$	●	10	20	30	$\mu\text{A}$
$t_{\text{WD}}$	Watchdog Time-Out Period	$C_{\text{WT}} = 1500\text{pF}$	●	20	30	40	ms
$V_{\text{IH}}$	WDI Input Threshold High	$V_{\text{CC}} = 3.3\text{V}$ to $5.5\text{V}$	●			1.6	V
$V_{\text{IL}}$	WDI Input Threshold Low	$V_{\text{CC}} = 3.3\text{V}$ to $5.5\text{V}$	●	0.4			V
$t_{\text{WP}}$	WDI Input Pulse Width	$V_{\text{CC}} = 3.3\text{V}$	●	150			ns
$I_{\text{WDI}}$	WDI Pull-Up Current	$V_{\text{WDI}} = 0\text{V}$			-10		$\mu\text{A}$

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

**Note 2:** All currents into pins are positive, all voltages are referenced to GND unless otherwise noted.

**Note 3:** The greater of  $V_1$ ,  $V_2$  is the internal supply voltage ( $V_{\text{CC}}$ ).

**Note 4:** Under static no-fault conditions,  $V_1$  will necessarily supply quiescent current. If at any time  $V_2$  is larger than  $V_1$ ,  $V_2$  must be capable of supplying

the quiescent current, programming (transient) current and reference load current.

**Note 5:** The output pins  $\overline{\text{RST}}$ ,  $\overline{\text{WDO}}$  and COMPX have internal pull-ups to  $V_2$  of typically  $6\mu\text{A}$ . However, external pull-up resistors may be used when faster rise times are required or for  $V_{\text{OH}}$  voltages greater than  $V_2$ .

**Note 6:** The push-pull  $\overline{\text{RST}}$  output pin on the LTC2901-2 is *actively* pulled up to  $V_2$ .



## TEST CIRCUITS

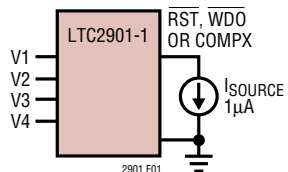


Figure 1.  $\overline{\text{RST}}$ ,  $\overline{\text{WDO}}$ ,  $\text{COMPX}$   $V_{OH}$  Test

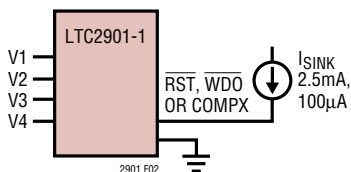


Figure 2.  $\overline{\text{RST}}$ ,  $\overline{\text{WDO}}$ ,  $\text{COMPX}$   $V_{OL}$  Test

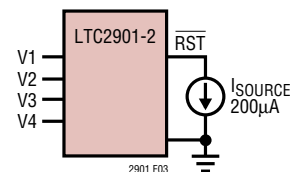
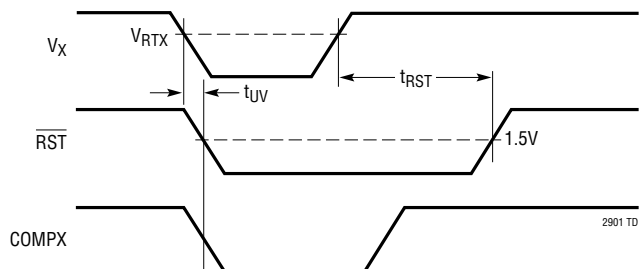


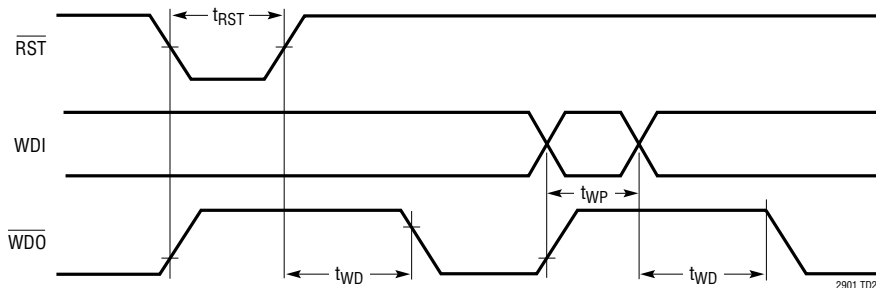
Figure 3. Active Pull-Up  $\overline{\text{RST}}$   $V_{OH}$  Test

## TIMING DIAGRAMS

### $V_X$ Monitor Timing



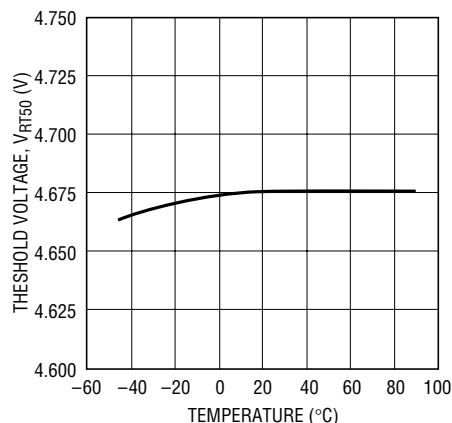
### Watchdog Timing





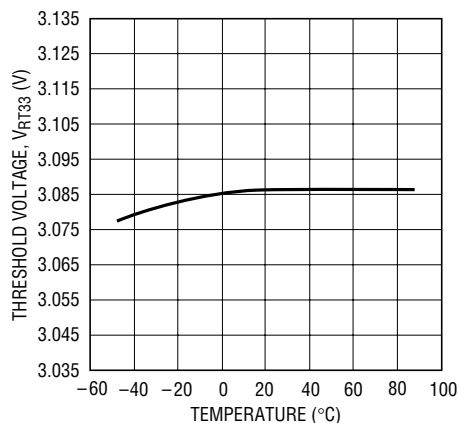
# TYPICAL PERFORMANCE CHARACTERISTICS

**5V Threshold Voltage  
vs Temperature**



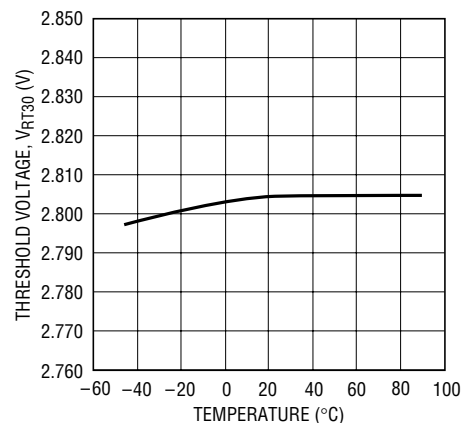
2901 G01

**3.3V Threshold Voltage  
vs Temperature**



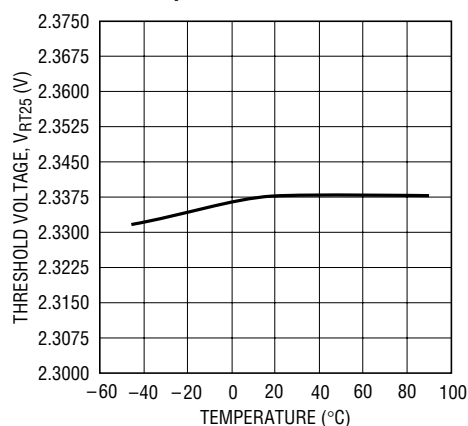
2901 G02

**3V Threshold Voltage  
vs Temperature**



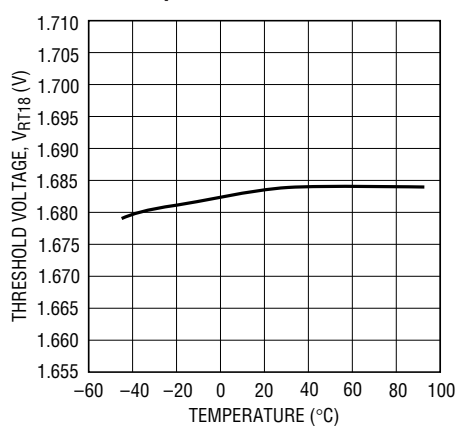
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**2.5V Threshold Voltage  
vs Temperature**



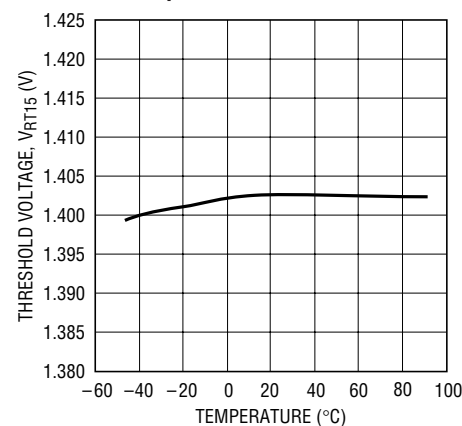
2901 G04

**1.8V Threshold Voltage  
vs Temperature**



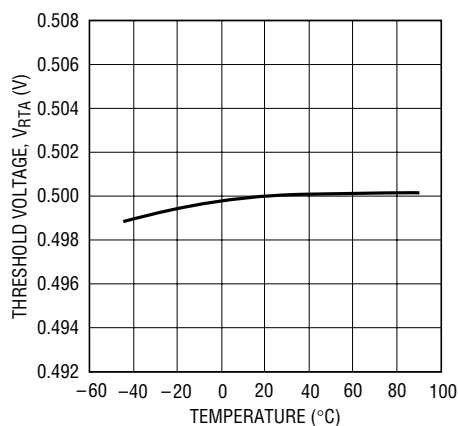
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**1.5V Threshold Voltage  
vs Temperature**



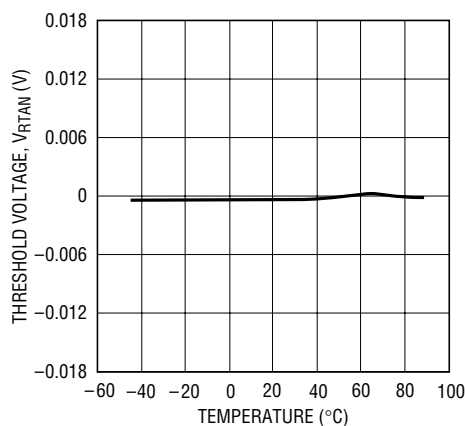
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**ADJ Threshold Voltage  
vs Temperature**



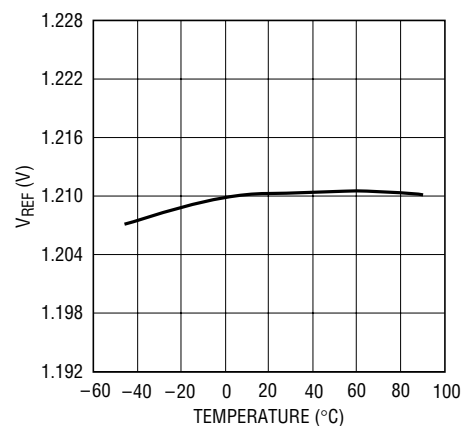
2901 G07

**-ADJ Threshold Voltage  
vs Temperature**



2901 G08

**$V_{REF}$  vs Temperature**

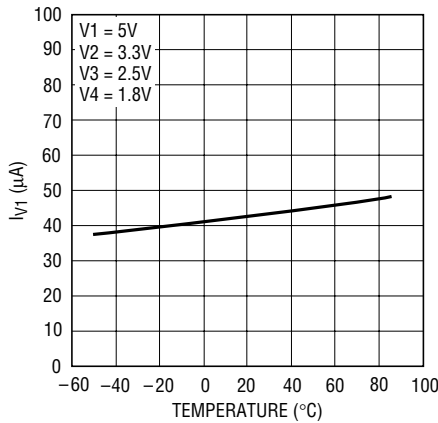
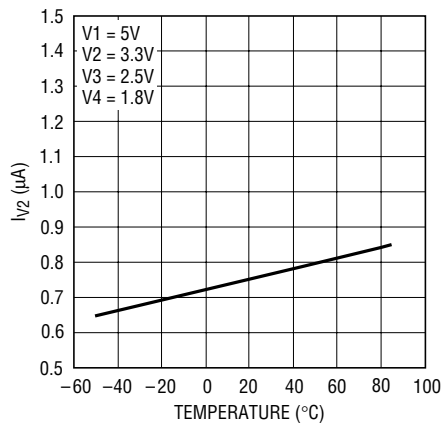
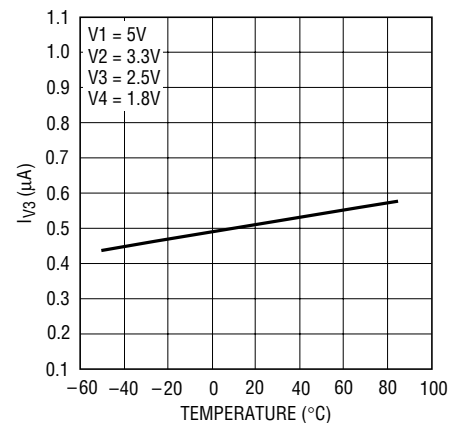
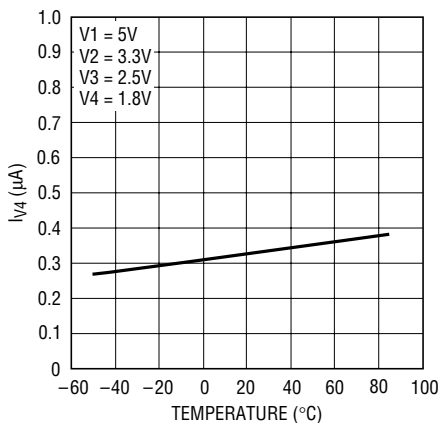


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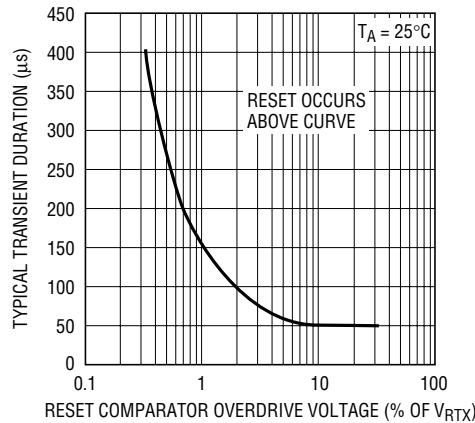
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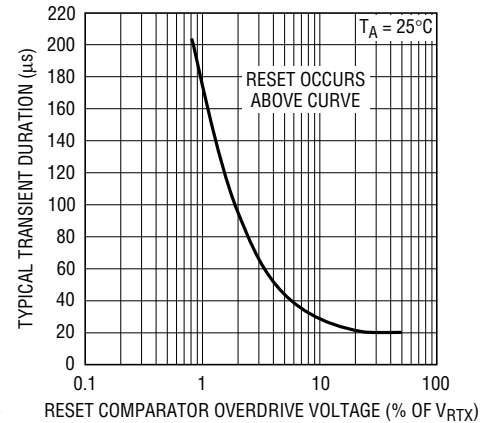
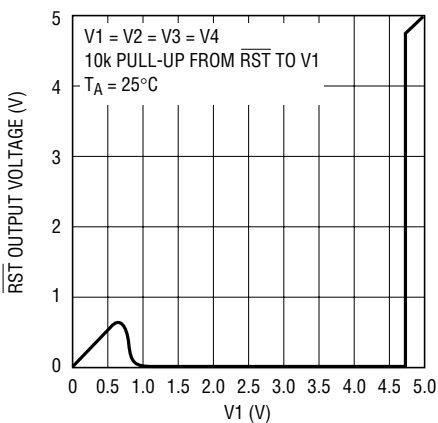
## TYPICAL PERFORMANCE CHARACTERISTICS

 $I_{V1}$  vs Temperature $I_{V2}$  vs Temperature $I_{V3}$  vs Temperature $I_{V4}$  vs Temperature

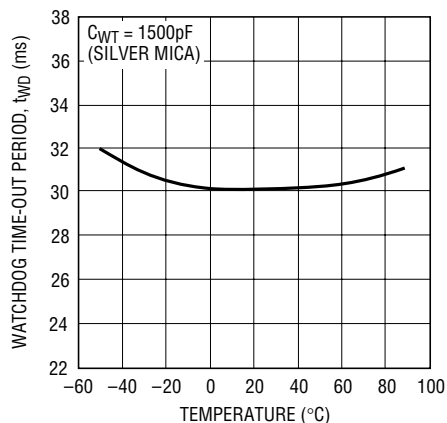
Typical Transient Duration vs Comparator Overdrive (V1, V2)



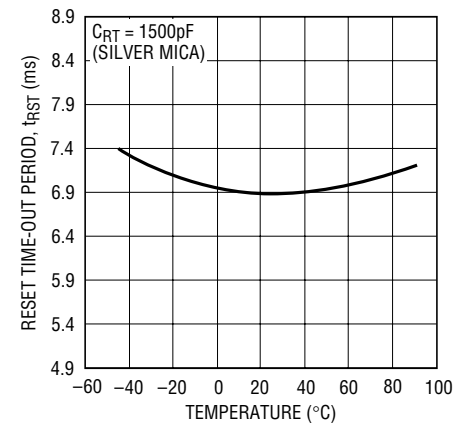
Typical Transient Duration vs Comparator Overdrive (V3, V4)

 $\overline{\text{RST}}$  Output Voltage vs V1,  $V_{PG} = 0V$ 

Watchdog Time-Out Period vs Temperature



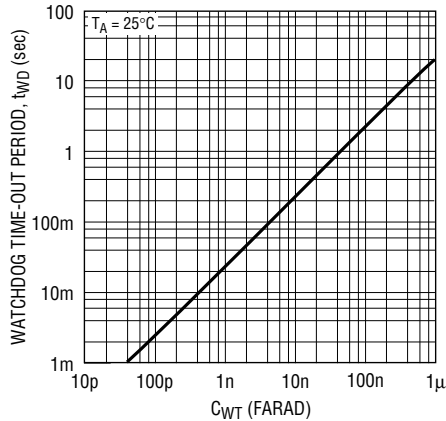
Reset Time-Out Period vs Temperature





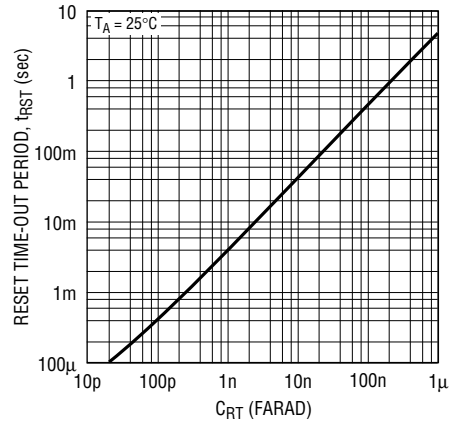
# TYPICAL PERFORMANCE CHARACTERISTICS

**Watchdog Time-Out Period vs Capacitance**



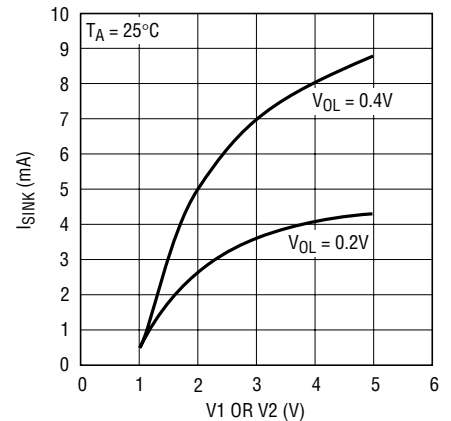
2901 G17

**Reset Time-Out Period vs Capacitance**



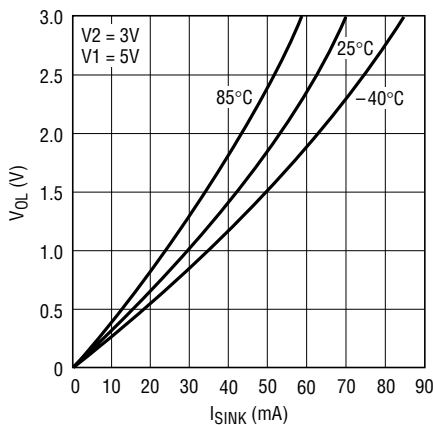
2901 G18

**RST, WDO, COMPM  $I_{SINK}$  vs Supply Voltage**



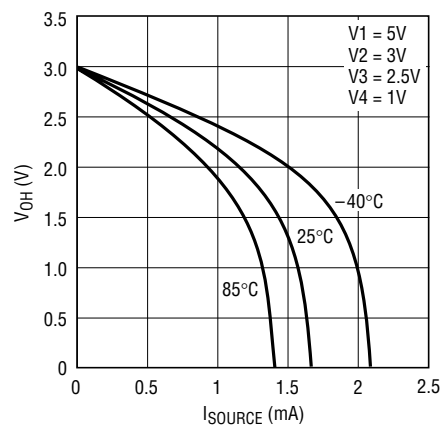
2901 G19

**RST, WDO, COMPM Voltage Output Low vs Output Sink Current**



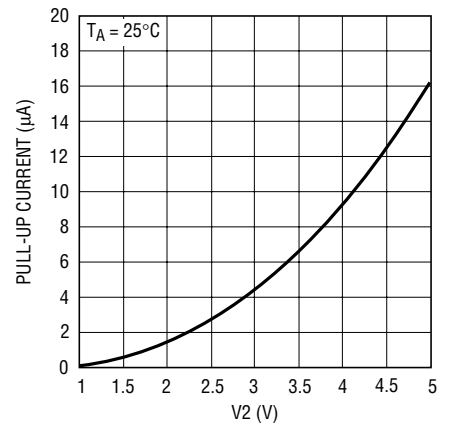
2901 G20

**RST High Level Output Voltage vs Output Source Current (LTC2901-2)**



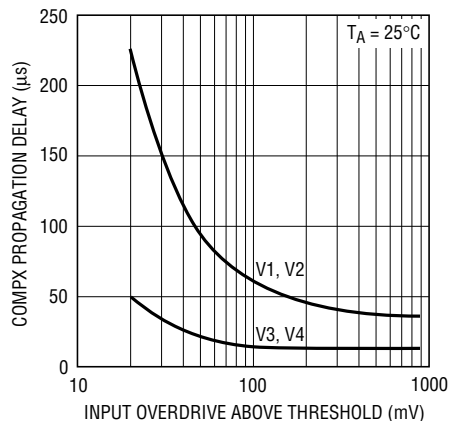
2901 G21

**COMPM and WDO Pull-Up Current vs  $V_2$  (COMPM and WDO Held at 0V)**



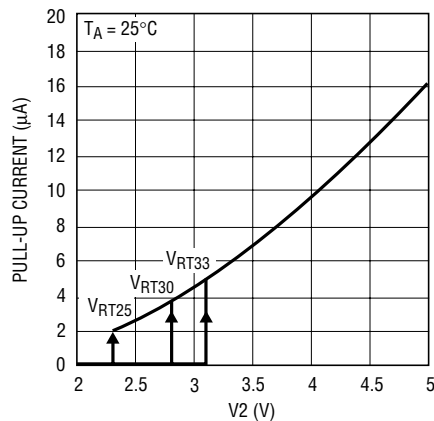
2901 G24

**COMPM Propagation Delay vs Input Overdrive Above Threshold**



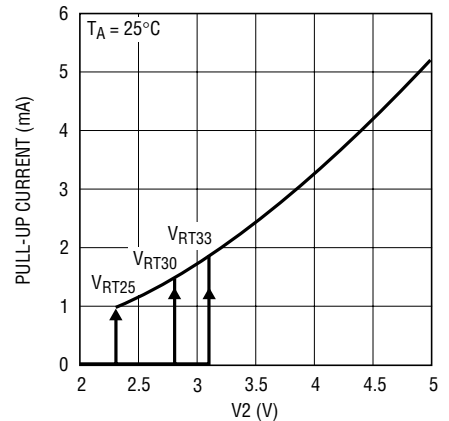
2901 G25

**RST Pull-Up Current vs  $V_2$  (LTC2901-1)**



2901 G26

**RST Pull-Up Current vs  $V_2$  (LTC2901-2)**



2901 G27

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## PIN FUNCTIONS

**COMP3 (Pin 1):** Comparator Output 3. Nondelayed, active high logic output with weak pull-up to V2. Pulls high when V3 is above reset threshold. May be pulled greater than V2 using external pull-up.

**COMP1 (Pin 2):** Comparator Output 1. Nondelayed, active high logic output with weak pull-up to V2. Pulls high when V1 is above reset threshold. May be pulled greater than V2 using external pull-up.

**V3 (Pin 3):** Voltage Input 3. Select from 2.5V, 1.8V, 1.5V or ADJ. See Table 1 for details.

**V1 (Pin 4):** Voltage Input 1. Select from 5V or 3.3V. See Table 1 for details. The greater of (V1, V2) is also  $V_{CC}$  for the device. Bypass this pin to ground with a 0.1 $\mu$ F (or greater) capacitor.

**CRT (Pin 5):** Reset Delay Time Programming Pin. Attach an external capacitor ( $C_{RT}$ ) to GND to set a reset delay time of 4.6ms/nF. Leaving the pin open generates a minimum delay of approximately 50 $\mu$ s. A 47nF capacitor will generate a 216ms reset delay time.

**RST (Pin 6):** Reset Logic Output. Active low with weak pull-up to V2 (LTC2901-1) or active pull-up to V2 (LTC2901-2). Pulls low when any voltage input is below the reset threshold and held low for the programmed delay time after all voltage inputs are above threshold. May be pulled above V2 using an external pull-up (LTC2901-1 only).

**WDO (Pin 7):** Watchdog Output. Active low logic output with weak pull-up to V2. May be pulled greater than V2 using external pull-up. The watchdog output pulls low if the watchdog timer is allowed to time out and remains low until set high by the next WDI transition or anytime RST is low. The watchdog timer is enabled when RST is high.

**WDI (Pin 8):** Watchdog Input. A logic input whose rising or falling edge must occur on this pin (while RST is high) within the selected watchdog time-out period, prohibiting a high-to-low transition on the WDO pin. The watchdog time-out period is set by the value of the capacitor that is attached to the CWT pin. A rising or falling edge on the WDI pin clears the voltage on the  $C_{WT}$  capacitor, preventing WDO from going low. When disabling the watchdog function, tie CWT to GND.

**CWT (Pin 9):** Watchdog Time-Out Programming Pin. Attach a capacitor ( $C_{WT}$ ) between CWT and GND to set a watchdog time-out period of 20ms/nF. Leaving the pin open generates a minimum time-out of approximately 200 $\mu$ s. A 47nF capacitor will generate a 940ms watchdog time-out period.

**GND (Pin 10):** Ground.

**V<sub>PG</sub> (Pin 11):** Voltage Threshold Combination Select Input. Connect to an external 1% resistive divider between  $V_{REF}$  and GND to select 1 of 16 combinations of preset and/or  $\pm$ adjustable voltage thresholds (see Table 1). Do not add capacitance on the V<sub>PG</sub> pin.

**V<sub>REF</sub> (Pin 12):** Buffered Reference Voltage. A 1.210V nominal reference used for programming voltage ( $V_{PG}$ ) and for the offset of negative adjustable applications. The buffered reference can source and sink up to 1mA. The reference can drive a bypass capacitor of up to 1000pF without oscillation.

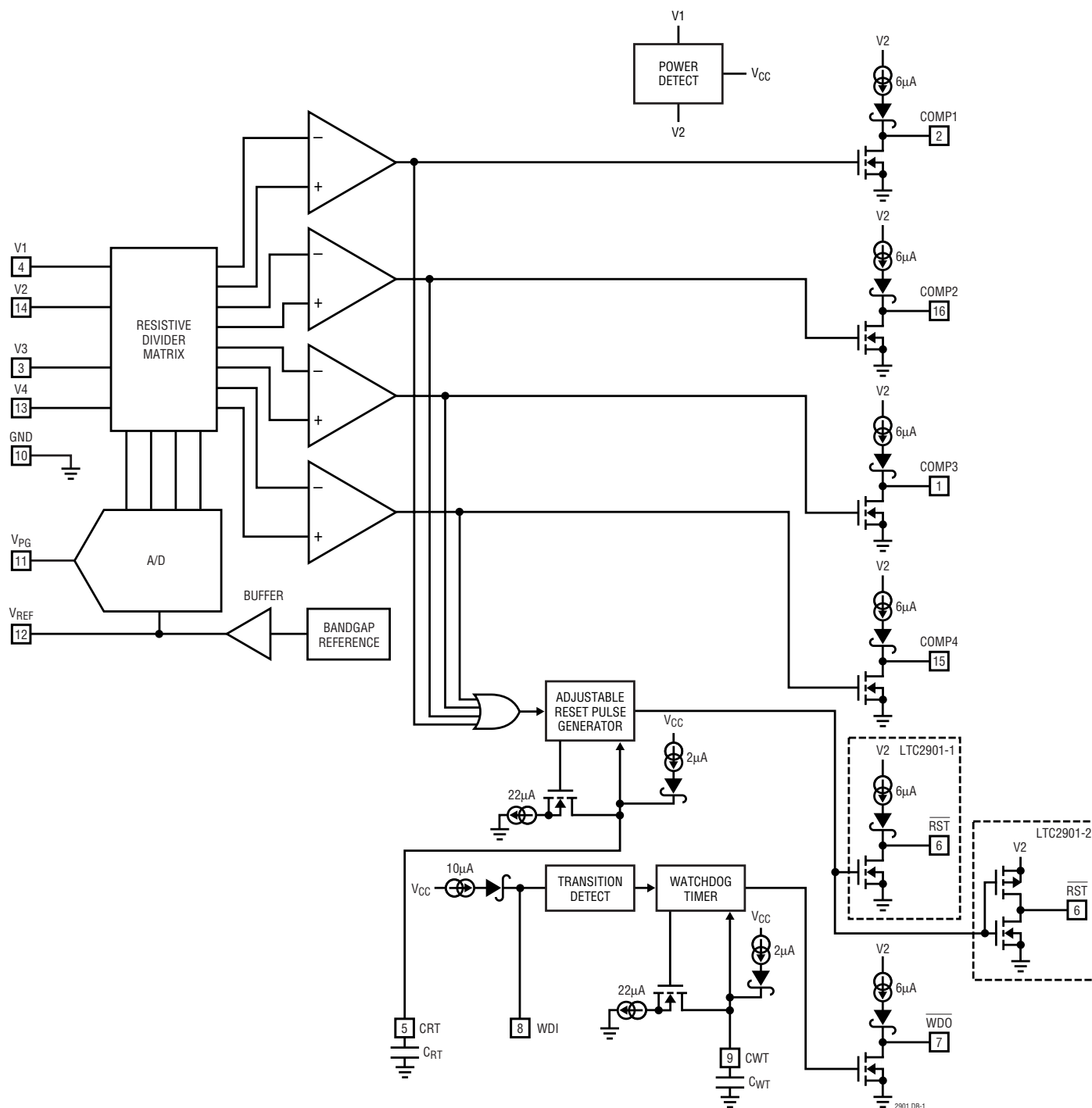
**V4 (Pin 13):** Voltage Input 4. Select from 1.8V, 1.5V, ADJ or –ADJ. See Table 1 for details.

**V2 (Pin 14):** Voltage Input 2. Select from 3.3V, 3V or 2.5V. See Table 1 for details. The greater of (V1, V2) is also  $V_{CC}$  for device. Bypass this pin to ground with a 0.1 $\mu$ F (or greater) capacitor. All logic outputs (COMP1, COMP2, COMP3, COMP4, RST, WDO) are weakly pulled up to V2 (LTC2901-1). RST is actively pulled up to V2 in the LTC2901-2.

**COMP4 (Pin 15):** Comparator Output 4. Nondelayed, active high logic output with weak pull-up to V2. Pulls high when V4 is above reset threshold. May be pulled greater than V2 using external pull-up.

**COMP2 (Pin 16):** Comparator Output 2. Nondelayed, active high logic output with weak pull-up to V2. Pulls high when V2 is above reset threshold. May be pulled greater than V2 using external pull-up.







## APPLICATIONS INFORMATION

### Power-Up

The greater of V1, V2 is the internal supply voltage ( $V_{CC}$ ). On power-up,  $V_{CC}$  will power the drive circuits for the  $\overline{RST}$  and the COMPX pins. This ensures that the  $\overline{RST}$  and COMPX outputs will be low as soon as V1 or V2 reaches 1V. The  $\overline{RST}$  and COMPX outputs will remain low until the part is programmed. After programming, if any one of the  $V_X$  inputs is below its programmed threshold,  $\overline{RST}$  will be a logic low. Once all the  $V_X$  inputs rise above their thresholds, an internal timer is started and  $\overline{RST}$  is released after the programmed delay time. If  $V_{CC} < (V3 - 1)$  and  $V_{CC} < 2.4V$ , the V3 input impedance will be low ( $1k\Omega$  typ).

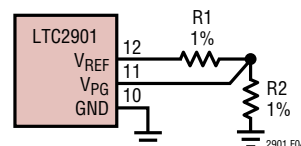
### Monitor Programming

The LTC2901 input voltage combination is selected by placing the recommended resistive divider from  $V_{REF}$  to GND and connecting the tap point to  $V_{PG}$ , as shown in Figure 4. Table 1 offers recommended 1% resistor values for the various modes. The last column in Table 1 specifies optimum  $V_{PG}/V_{REF}$  ratios ( $\pm 0.01$ ) to be used when programming with a ratiometric DAC.

During power-up, once V1 or V2 reaches 2.4V max, the monitor enters a programming period of approximately

**Table 1. Voltage Threshold Programming**

MODE	V1 (V)	V2 (V)	V3 (V)	V4 (V)	R1 (k $\Omega$ )	R2 (k $\Omega$ )	$V_{PG}/V_{REF}$
0	5.0	3.3	ADJ	ADJ	Open	Short	0.000
1	5.0	3.3	ADJ	–ADJ	93.1	9.53	0.094
2	3.3	2.5	ADJ	ADJ	86.6	16.2	0.156
3	3.3	2.5	ADJ	–ADJ	78.7	22.1	0.219
4	3.3	2.5	1.5	ADJ	71.5	28.0	0.281
5	5.0	3.3	2.5	ADJ	66.5	34.8	0.344
6	5.0	3.3	2.5	1.8	59.0	40.2	0.406
7	5.0	3.3	2.5	1.5	53.6	47.5	0.469
8	5.0	3.0	2.5	ADJ	47.5	53.6	0.531
9	5.0	3.0	ADJ	ADJ	40.2	59.0	0.594
10	3.3	2.5	1.8	1.5	34.8	66.5	0.656
11	3.3	2.5	1.8	ADJ	28.0	71.5	0.719
12	3.3	2.5	1.8	–ADJ	22.1	78.7	0.781
13	5.0	3.3	1.8	–ADJ	16.2	86.6	0.844
14	5.0	3.3	1.8	ADJ	9.53	93.1	0.906
15	5.0	3.0	1.8	ADJ	Short	Open	1.000



**Figure 4. Monitor Programming**

150 $\mu$ s during which the voltage on the  $V_{PG}$  pin is sampled and the monitor is configured to the desired input combination. Do not add capacitance to the  $V_{PG}$  pin. Immediately after programming, the comparators are enabled and supply monitoring will begin.

### Supply Monitoring

The LTC2901 is a low power, high accuracy programmable quad supply monitoring circuit with four nondelayed monitor outputs, a common reset output and a watchdog timer. Watchdog and reset timing are both adjustable using external capacitors. Single pin programming selects 1 of 16 input voltage monitor combinations. All four voltage inputs must be above predetermined thresholds for the reset not to be invoked. The LTC2901 will assert the reset and comparator outputs during power-up, power-down and brownout conditions on any one of the voltage inputs.

The inverting inputs on the V3 and/or V4 comparators are set to 0.5V when the positive adjustable modes are selected (Figure 5). The tap point on an external resistive divider, connected between the positive voltage being sensed and ground, is connected to the high impedance noninverting inputs (V3, V4). The trip voltage is calculated from:

$$V_{TRIP} = 0.5V \left( 1 + \frac{R3}{R4} \right)$$

In the negative adjustable mode, the noninverting input on the V4 comparator is connected to ground (Figure 6). The tap point on an external resistive divider, connected between the negative voltage being sensed and the  $V_{REF}$  pin, is connected to the high impedance inverting input (V4).  $V_{REF}$  provides the necessary level shift required to operate at ground. The trip voltage is calculated from:

$$V_{TRIP} = -V_{REF} \left( \frac{R3}{R4} \right); V_{REF} = 1.210V \text{ Nominal}$$



## APPLICATIONS INFORMATION

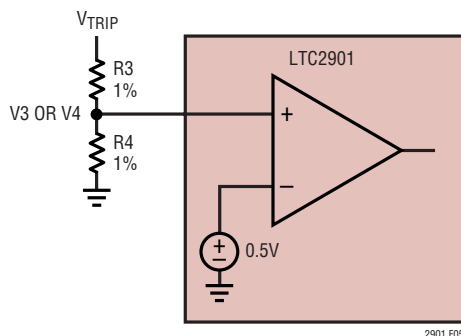


Figure 5. Setting the Positive Adjustable Trip Point

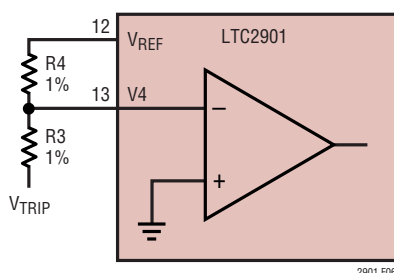


Figure 6. Setting the Negative Adjustable Trip Point

In a negative adjustable application, the minimum value for R4 is limited by the sourcing capability of  $V_{REF}$  ( $\pm 1\text{mA}$ ). With no other load on  $V_{REF}$ , R4 (minimum) is:

$$1.21\text{V} \div 1\text{mA} = 1.21\text{k}\Omega.$$

Tables 2 and 3 offer suggested 1% resistor values for various adjustable applications.

Although all four supply monitor comparators have built-in glitch immunity, bypass capacitors on V1 and V2 are recommended because the greater of V1 or V2 is also the  $V_{CC}$  for the device. Filter capacitors on the V3 and V4 inputs are allowed.

### Power-Down

On power-down, once any of the  $V_X$  inputs drop below their threshold,  $\overline{\text{RST}}$  and  $\text{COMPX}$  are held at a logic low. A logic low of 0.4V is guaranteed until both V1 and V2 drop below 1V. If the bandgap reference becomes invalid ( $V_{CC} < 2\text{V}$  typ), the part will reprogram once  $V_{CC}$  rises above 2.4V max.

Table 2. Suggested 1% Resistor Values for the ADJ Inputs

$V_{\text{SUPPLY}}$ (V)	$V_{\text{TRIP}}$ (V)	R3 (k $\Omega$ )	R4 (k $\Omega$ )
12	11.25	2150	100
10	9.4	1780	100
8	7.5	1400	100
7.5	7	1300	100
6	5.6	1020	100
5	4.725	845	100
3.3	3.055	511	100
3	2.82	464	100
2.5	2.325	365	100
1.8	1.685	237	100
1.5	1.410	182	100
1.2	1.120	124	100
1	0.933	86.6	100
0.9	0.840	68.1	100

Table 3. Suggested 1% Resistor Values for the -ADJ Input

$V_{\text{SUPPLY}}$ (V)	$V_{\text{TRIP}}$ (V)	R3 (k $\Omega$ )	R4 (k $\Omega$ )
-2	-1.87	187	121
-5	-4.64	464	121
-5.2	-4.87	487	121
-10	-9.31	931	121
-12	-11.30	1130	121

### Monitor Output Rise and Fall Time Estimation

All of the outputs ( $\overline{\text{RST}}$ ,  $\text{COMPX}$ ,  $\overline{\text{WDO}}$ ) have strong pull-down capability. If the external load capacitance ( $C_{\text{LOAD}}$ ) for a particular output is known, output fall time (10% to 90%) is estimated using:

$$t_{\text{FALL}} \approx 2.2 \cdot R_{\text{PD}} \cdot C_{\text{LOAD}}$$

where  $R_{\text{PD}}$  is the on-resistance of the internal pull-down transistor. The typical performance curve ( $V_{\text{OL}}$  vs  $I_{\text{SINK}}$ ) demonstrates that the pull-down current is somewhat linear versus output voltage. Using the 25°C curve,  $R_{\text{PD}}$  is estimated to be approximately 40 $\Omega$ . Assuming a 150pF load capacitance, the fall time is about 13.2ns.

Although the outputs are considered to be “open-drain,” they do have a weak pull-up capability (see  $\text{COMPX}$  or  $\overline{\text{RST}}$



## APPLICATIONS INFORMATION

Pull-Up Current vs V2 curve). Output rise time (10% to 90%) is estimated using:

$$t_{RISE} \approx 2.2 \cdot R_{PU} \cdot C_{LOAD}$$

where  $R_{PU}$  is the on-resistance of the pull-up transistor. The on-resistance as a function of the V2 voltage at room temperature is estimated using:

$$R_{PU} = \frac{6 \cdot 10^5}{V2 - 1} \Omega$$

with  $V2 = 3.3V$ ,  $R_{PU}$  is about 260k. Using 150pF for load capacitance, the rise time is 86μs. If the output needs to pull up faster and/or to a higher voltage, a smaller external pull-up resistor may be used. Using a 10k pull-up resistor, the rise time is reduced to 3.3μs for a 150pF load capacitance.

The LTC2901-2 has an active pull-up to V2 on the  $\overline{RST}$  output. The typical performance curve ( $\overline{RST}$  Pull-Up Current vs V2 curve) demonstrates that the pull-up current is somewhat linear versus the V2 voltage and  $R_{PU}$  is estimated to be approximately 625Ω. A 150pF load capacitance makes the rise time about 206ns.

### Watchdog Timer

The watchdog circuit typically monitors a μP's activity. The μP is required to change the logic state of the WDI pin on a periodic basis in order to clear the watchdog timer and prevent the  $\overline{WDO}$  pin from going low. Whenever  $\overline{RST}$  is low, the watchdog timer is cleared and  $\overline{WDO}$  is set high. The watchdog timer is started when  $\overline{RST}$  pulls high. Subsequent edges received on the WDI pin will clear the watchdog timer. The timer will continue to run until the watchdog timer times out. Once the watchdog timer times out, internal circuitry will bring the  $\overline{WDO}$  pin low.  $\overline{WDO}$  will remain low and the watchdog timer will remain cleared until the next edge is received on the WDI pin or until  $\overline{RST}$  goes low.

To disable the watchdog timer, simply ground the CWT pin (Pin 9). With CWT held at ground, any reset event will force  $\overline{WDO}$  high indefinitely. It is safe to leave the WDI pin (Pin 8) unconnected because the weak internal pull-up (10μA typ) will pull WDI high. Tying WDI to V1 or ground is also allowed, but grounding the WDI pin will force the pull-up current to be drawn continuously.

### Selecting the Reset Timing Capacitor

The reset time-out period is adjustable in order to accommodate a variety of microprocessor applications. The reset time-out period,  $t_{RST}$ , is adjusted by connecting a capacitor,  $C_{RT}$ , between the CRT pin and ground. The value of this capacitor is determined by:

$$C_{RT} = t_{RST} \cdot 217 \cdot 10^{-9}$$

with  $C_{RT}$  in Farads and  $t_{RST}$  in seconds. The  $C_{RT}$  value per millisecond of delay can also be expressed as  $C_{RT}/ms = 217$  (pF/ms).

Leaving the CRT pin unconnected will generate a minimum reset time-out of approximately 50μs. Maximum reset time-out is limited by the largest available low leakage capacitor. The accuracy of the time-out period will be affected by capacitor leakage (the nominal charging current is 2μA) and capacitor tolerance. A low leakage ceramic capacitor is recommended.

### Selecting the Watchdog Timing Capacitor

The watchdog time-out period is adjustable and can be optimized for software execution. The watchdog time-out period,  $t_{WD}$ , is adjusted by connecting a capacitor,  $C_{WT}$ , between the CWT pin and ground. Given a specified watchdog time-out period, the capacitor is determined by:

$$C_{WT} = t_{WD} \cdot 50 \cdot 10^{-9}$$

with  $C_{WT}$  in Farads and  $t_{WD}$  in seconds. The  $C_{WT}$  value per millisecond of delay can also be expressed as  $C_{WT}/ms = 50$  (pF/ms).

Leaving the CWT pin unconnected will generate a minimum watchdog time-out of approximately 200μs. Maximum time-out is limited by the largest available low leakage capacitor. The accuracy of the time-out period will be affected by capacitor leakage (the nominal charging current is 2μA) and capacitor tolerance. A low leakage ceramic capacitor is recommended.

### Monitoring Power Supply Controller Activity

Figure 7 demonstrates how the LTC2901 can be used to monitor switcher activity. The monitor is configured to supervise 3.3V, 2.5V, 1.8V and one adjustable input.



## APPLICATIONS INFORMATION

Because 2.5V does not exist in this application, the V2 input is tied to the V1 (3.3V) input. The feedback voltage on the LTC1772 (0.8V typ) is monitored with the adjustable input (V4). The  $\overline{\text{RST}}$  pin will go high 216ms ( $C_{\text{RT}} = 47\text{nF}$ ) after the 3.3V and 1.8V supplies and the feedback voltage are above threshold. Individual input status is available at the COMPX pins.

While the voltage monitors can detect low voltage or shorted inputs, the watchdog circuit can be used to detect an open circuit to the primary load. With the CWT pin unconnected, the watchdog time-out is approximately 200 $\mu\text{s}$ . At low load currents on the 1.8V supply, the LTC1772 will go into Burst Mode<sup>®</sup> operation. With an open-circuit load, the duty cycle at the gate of M1 will drop, and the pulse spacing will exceed the watchdog time-out period. The  $\overline{\text{WDO}}$  pin will go low indicating the low load condition. The  $\overline{\text{WDO}}$  pin will return high on the next pulse

to the gate of M1. The  $\overline{\text{WDO}}$  pin will remain high if the load is restored.

### Ensuring Reset Valid for $V_{\text{CC}}$ Down to 0V (LTC2901-2)

Some applications require the reset output ( $\overline{\text{RST}}$ ) to be valid with  $V_{\text{CC}}$  down to 0V. The LTC2901-2 is designed to handle this requirement with the addition of an external resistor from  $\overline{\text{RST}}$  to ground. The resistor will provide a path for stray charge and/or leakage currents, preventing the  $\overline{\text{RST}}$  output from floating to undetermined voltages when connected to high impedance (such as CMOS logic inputs). The resistor value should be small enough to provide effective pull-down without excessively loading the active pull-up circuitry. Too large a value may not pull down well enough. A 100k resistor from  $\overline{\text{RST}}$  to ground is satisfactory for most applications.

Burst Mode is a registered trademark of Linear Technology Corporation.

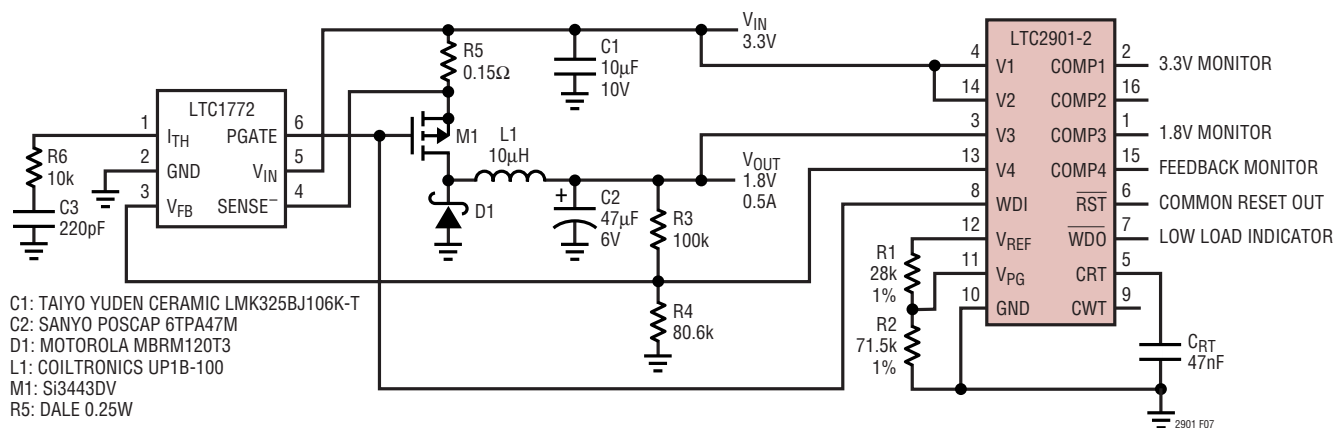
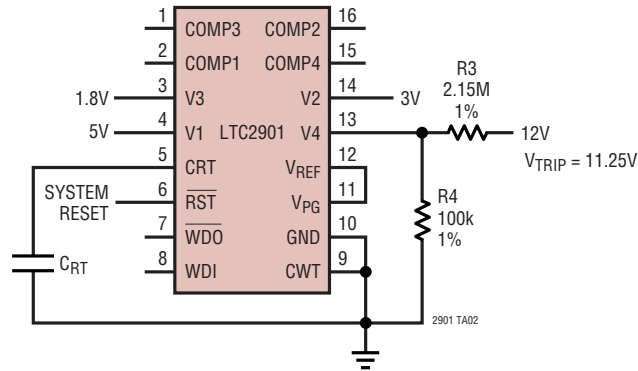


Figure 7. Monitor Input, Output, Feedback Voltage and Low Load Conditions on DC/DC Controller

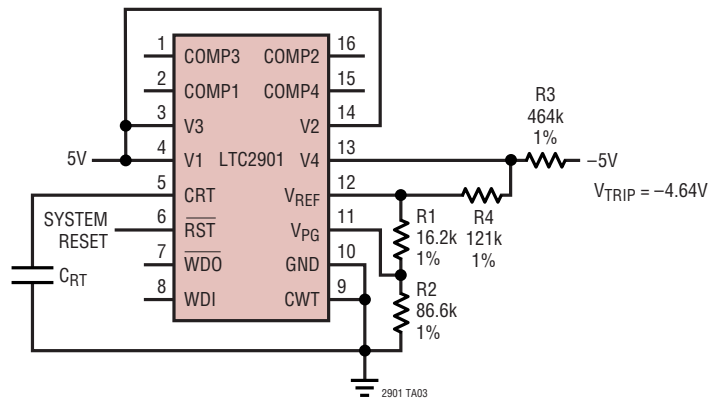


## TYPICAL APPLICATIONS

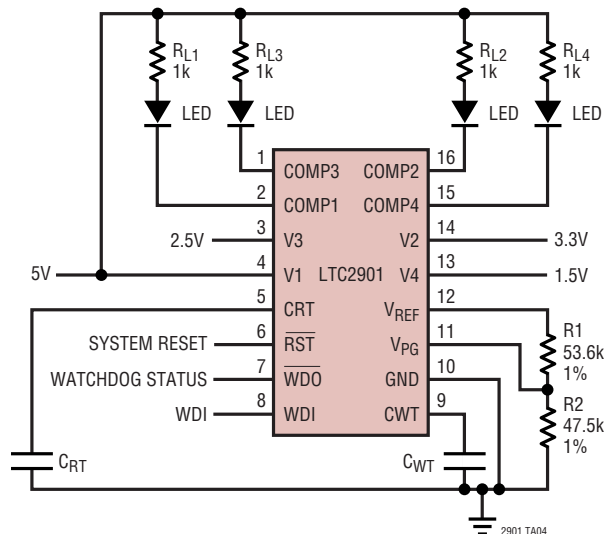
**Quad Supply Monitor with Watchdog Timer Disabled  
5V, 3V, 1.8V, 12V (ADJ)**



**5V, -5V Monitor with Watchdog Timer Disabled and Unused  
V2, V3 Inputs Pulled Above Trip Thresholds**



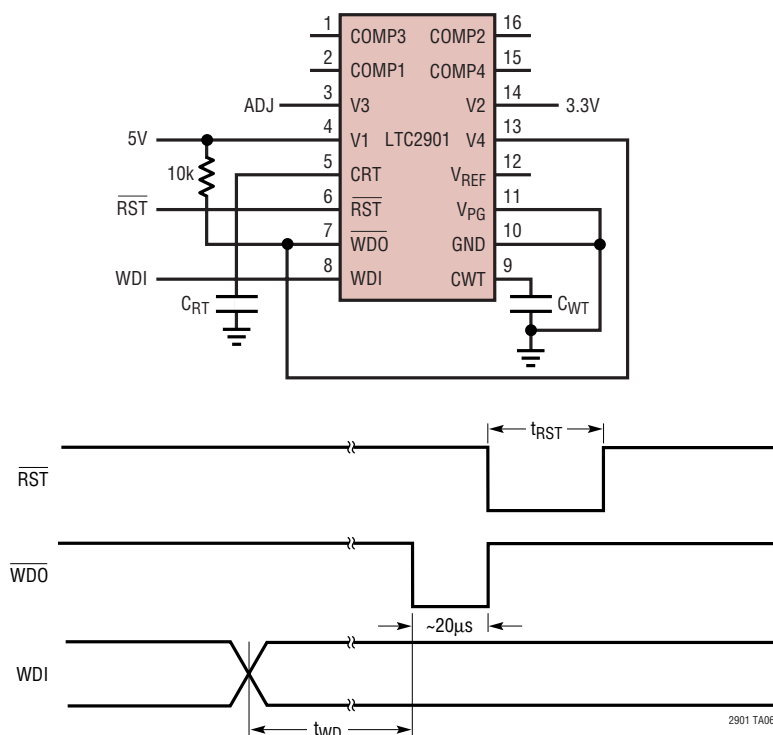
**Quad Supply Monitor with LED Undervoltage Indicators  
5V, 3.3V, 2.5V, 1.5V**





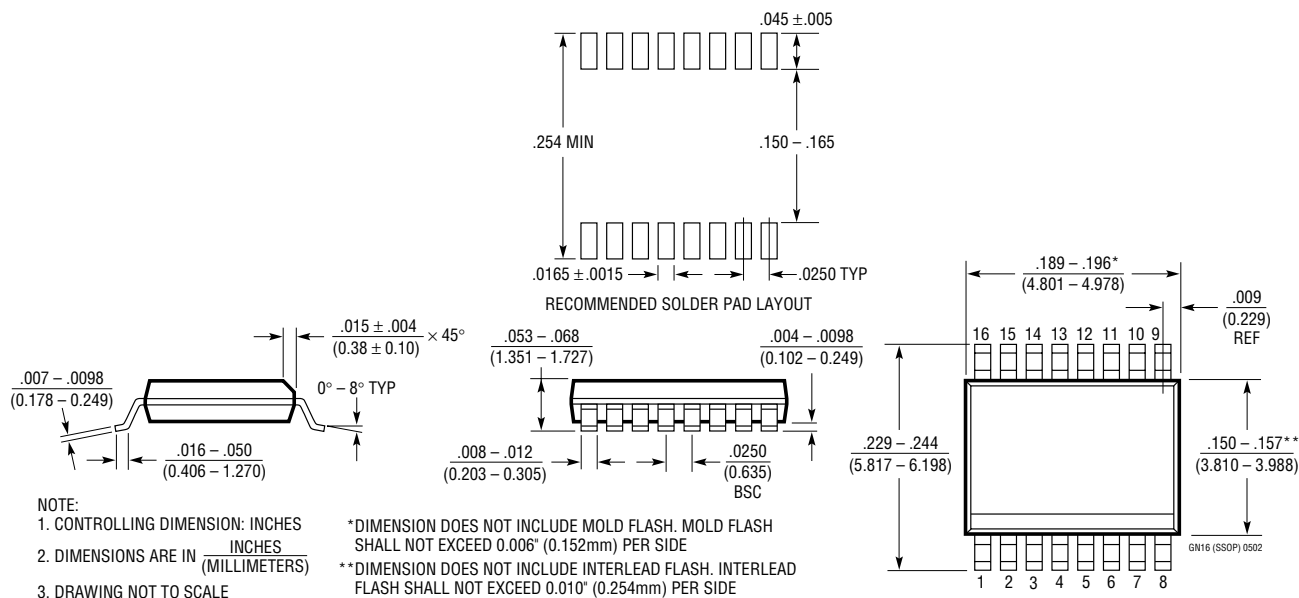
## TYPICAL APPLICATION

### Generate RESET Pulse Through Watchdog Timeout



## PACKAGE DESCRIPTION

### GN Package 16-Lead Plastic SSOP (Narrow .150 Inch) (Reference LTC DWG # 05-08-1641)

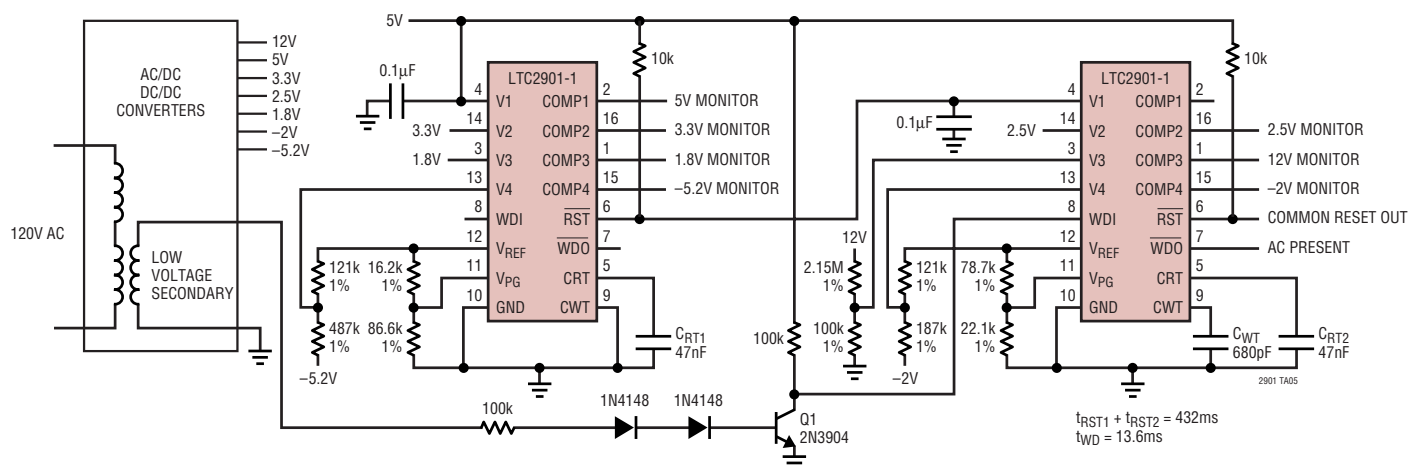


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## TYPICAL APPLICATION

Monitor Seven Supplies (12V, 5V, 3.3V, 2.5V, 1.8V, -2V, -5.2V) with Sequenced Reset and AC Present Indication



## RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LTC690	5V Supply Monitor, Watchdog Timer and Battery Backup	4.65V Threshold
LTC694-3.3	3.3V Supply Monitor, Watchdog Timer and Battery Backup	2.9V Threshold
LTC699	5V Supply Monitor and Watchdog Timer	4.65V Threshold
LTC1232	5V Supply Monitor, Watchdog Timer and Push-Button Reset	4.37V/4.62V Threshold
LTC1326	Micropower Precision Triple Supply Monitor for 5V, 3.3V and ADJ	4.725V, 3.118V, 1V Thresholds ( $\pm 0.75\%$ )
LTC1326-2.5	Micropower Precision Triple Supply Monitor for 2.5V, 3.3V and ADJ	2.363V, 3.118V, 1V Thresholds ( $\pm 0.75\%$ )
LTC1536	Precision Triple Supply Monitor for PCI Applications	Meets PCI $t_{FAIL}$ Timing Specifications
LTC1726-2.5	Micropower Triple Supply Monitor for 2.5V, 3.3V and ADJ	Adjustable $\overline{RESET}$ and Watchdog Time-Outs
LTC1726-5	Micropower Triple Supply Monitor for 5V, 3.3V and ADJ	Adjustable $\overline{RESET}$ and Watchdog Time-Outs
LTC1727-2.5/LTC1727-5	Micropower Triple Supply Monitors with Open-Drain Reset	Individual Monitor Outputs in MSOP
LTC1728-1.8/LTC1728-3.3	Micropower Triple Supply Monitor with Open-Drain Reset	5-Lead SOT-23 Package
LTC1728-2.5/LTC1728-5	Micropower Triple Supply Monitors with Open-Drain Reset	5-Lead SOT-23 Package
LTC1985-1.8	Micropower Triple Supply Monitor with Push-Pull Reset Output	5-Lead SOT-23 Package
LTC2900	Programmable Quad Supply Monitor	Adjustable Reset Timer, 10-Lead MSOP and 3mm $\times$ 3mm 10-Lead DFN
LTC2902	Programmable Quad Supply Monitor	Adjustable Reset Timer, Supply Tolerance and Margining Functions, 16-Lead Narrow SSOP