

VOLTAGE REGULATOR FOR CAR RADIO APPLICATIONS

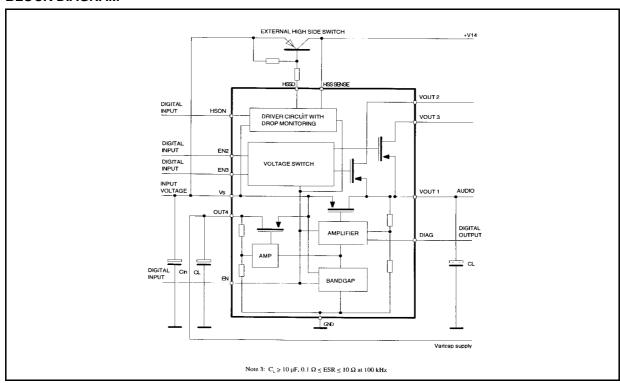
- OPERATION SUPPLY VOLTAGE 9.5 TO 28V
- PEAK SUPPLY OVERVOLTAGE PULSE UP TO 52V (VS. GND)
- FIXED OUTPUT VOLTAGE OF 8.6 ± 0.2V AND OUTPUT CURRENT UP TO 150mA
- VERY LOW DROP OUTPUT STAGE WITH LOW OUTPUT VOLTAGE DIAGNOSTIC
- OUTPUT SHORT CIRCUIT AND THERMAL OVERLOAD PROTECTION
- VERY LOW STANDBY CURRENT (DEVICE DISABLED)
- TWO INTERNAL OUTPUT VOLTAGE SWITCHES WITH LOW DROPOUT FOR AM/FM SWITCHING
- DRIVER CIRCUIT FOR EXTERNAL HIGH SIDE SWITCH WITH DROP MONITORING
- ENABLE INPUT TO SWITCH ON THE DE-VICE
- ADDITIONAL 10V VERY LOW DROP REGU-LATOR TO SUPPLY THE VARICAP



DESCRIPTION

The L4952 is a monolithic voltage regulator in an advanced BCD60II process with high efficient p-channel transistor. This device is optimized for Car Radio applications to obtain optimum performance and supply system integration with high flexibility and minimum peripheral components.

BLOCK DIAGRAM



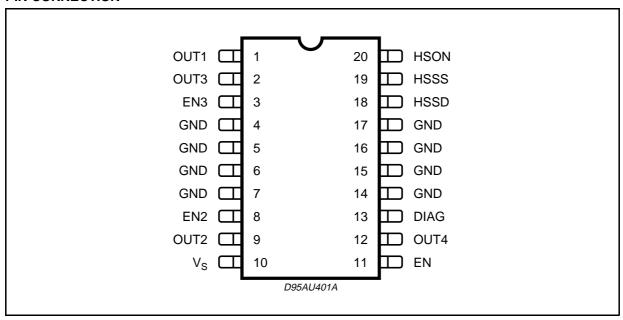
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ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter | Value | Unit |
|--------------------|---|-------------------------|------|
| V_s | DC supply voltage | -2.2 +28 | V |
| V_{st} | Transient operating supply voltage for t = 400ms | -2.2 +52 | V |
| $V_{01,2,3,4}$ | Output voltages | -0.3 +12 | V |
| V_{EN} | Input voltage of EN | -0.3 +18 | V |
| V_{IND} | Input voltage for EN2, EN3, HSON | -18 +18 | V |
| V_s - V_{IND} | Max Voltage Difference between V _s and logic inputs EN2, EN3, HSON | 52 | V |
| V_{DIAG} | Diagnostic output | -0.3 +20 | V |
| I _{o1} | Output current out1, out4 | short circuit protected | mA |
| $I_{02,3}$ | Output current out2,3 (Note 1) | internally limited | mA |
| I _{oDIAG} | Output current Diagnostic | 10 | mA |
| I _{HSSD} | Driver current for external pnp High side switch | internally limited | mA |
| TJ | Operating junction temperature | -40 +150 | °C |
| Po | Power dissipation with on board heat sink 2cm ² | 1 | W |
| V _{ESD} | ESD voltage capability (MIL 883 C) | ±2000 | V |

Note: Current limiter of OUT1 will also protect OUT2 and OUT3 as long as OUT1 is not reverse supplied. Output capacitors up to 100μF between OUT and GND will not harm this protection.

PIN CONNECTION



THERMAL DATA

| Symbol | Parameter | Value | Unit |
|-----------------------|---|-------|------|
| $R_{\text{th j-amb}}$ | Thermal Resistance Junction to Ambient with on board heatsink 2cm^2 | 60 | °C/W |
| T _{JSD} | Thermal shutdown junction temperature | >150 | °C |

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ELECTRICAL CHARACTERISTICS (V_s = 14V; T_J = 25°C unless otherwise specified)

| Symbol | Parameter Test Condition | | Min. | Тур. | Max. | Unit |
|-------------------------|--|--|------|-------|------|------|
| Is | V _s Quiescent current | no Load | | 1.6 | 5 | mA |
| I _{ssb} | V _s standby current | EN = 0V | | 0 | 20 | μΑ |
| V _{INLEN} | Low Voltage EN | | -0.3 | | 0.5 | V |
| V _{INHEN} | High Voltage EN | V _S = 18V | 3.5 | | 18 | V |
| V _{IN L EN2,3} | EN2, EN3, Low Voltage | | -18 | | 1.05 | V |
| V _{IN H EN2,3} | EN2, EN3, High Voltage | V _S = 18V | 1.45 | | 18 | V |
| V _{IN L HSON} | HSON Low Voltage | | -18 | | 1.0 | V |
| V _{IN H HSON} | HSON High Voltage | V _S = 18V | 2.0 | | 18 | V |
| I _{INH} | Current EN2, EN3, HSON | $V_{In} = 5V$ | | 30 | 70 | μΑ |
| I _{INL} | Current EN2, EN3, HSON | $V_{In} = 0V$ | -10 | | 10 | μΑ |
| I _{IN ON/OFF} | Current High EN | $V_{In} = 5V$ | -10 | 100 | 200 | μΑ |
| Voltage Re | gulator (OUT 1) | | | | | |
| V _{o1} | Output voltage | no load | 8.4 | 8.6 | 8.8 | V |
| V _{DP1} | Dropout voltage | $V_{s} = 8.4V, V_{DP1} = V_{s} - V_{out1}$ | | | | |
| | | $I_{o1}=0.15A, I_{02,3}=0mA$ | | 0.16 | 0.6 | V |
| | | I_{o1} = 0.1A, $I_{02,3}$ = 0mA | | 0.11 | 0.4 | V |
| | | I _{o1} = 0.05A, I _{02,3} = 0mA | | 0.06 | 0.2 | V |
| SVRR | Supply Voltage ripple rejection | f = 100Hz, $C = 10μF ESR = 5Ω$ | 60 | 70 | | dB |
| V_{oLo} | Load regulation | 10mA < I _o < 150 mA | | 30 | 60 | mV |
| I _{oLim1} | Current limits | V _{OUT} = 8V | 150 | 300 | 800 | mA |
| Voltage Re | gulator (OUT 4) | | | | | |
| V _{o4} | Output voltage | no load | 9.5 | 10.0 | 10.5 | V |
| V_{DP4} | Dropout voltage | $I = 3mA V_s = 8.6V$ | | 0.14 | 0.3 | V |
| | | $I = 1.5 \text{mA V}_s = 8.6 \text{V}$ | | 0.075 | 0.15 | V |
| SVRR | Supply Voltage ripple rejection | f = 100Hz C = 10 μF ESR = 5 Ω | 30 | 60 | | dB |
| V_{o4Lo} | Load regulation | 0.5mA < I < 3 mA | | 100 | | mV |
| I _{o4Lim} | Current limits | V _{OUT} = 8V | 8 | 60 | | mA |
| Diagnostic | Output | | | | | |
| V_{LDIAG} | Output Diagnostic Low voltage | R_{DIAG} to 5 V = 10K Ω | | | 0.4 | V |
| ΔV_{out} | Output voltage drop before diagnostic activation | $V_S > 6V I_{O1} = 100 \text{mA}$ $I_{O2} = I_{O3} = 0 \text{mA}$ | | 30 | 200 | mV |
| Voltage Sw | ritches Vout2/3 | | | | | |
| V_{DP2} | Dropout V _{out2} | $I_{o2} = 50 \text{mA}, En_2 = H$ | | 0.25 | 0.5 | V |
| | | I ₀₂ = 25mA, En ₂ = H | | 0.125 | 0.25 | V |
| V _{DP3} | Dropout V _{out3} | I _{o3} = 75mA, En ₃ = L | | 0.25 | 0.5 | V |
| | | I _{o3} = 40mA, En ₃ = L | | 0.14 | 0.28 | V |
| High Side S | Switch Driver | • | | • | | |
| V _{DPHS} | Low Drop Voltage | I _{out} = 50mA, HSON = H | | 0.26 | 0.8 | V |
| V _{DPHS} | Drop detection threshold | $V_{Dth} = V_S - V_{14}$, HSON = H | 0.7 | 1 | 1.3 | V |

FUNCTIONAL DESCRIPTION

The L4952 is a monolithic voltage regulator with an output voltage of typically 8.6V and a maximum output current of 150mA. It is a device for audio applications in carradios.

The device contain a precision Bandgap reference, a output amplifier with overcurrent protection, two voltage switches, a driver for an external

pnp switch, a discharge circuitry for theft detection and a overtemperature protection.

For automatic PCB assembly the package is SO12+4+4. To use the maximal current of 150mA a small copper area of 2cm² as heat sink on board (Rth=60K/W) is necessary.

The device has a very low quiescent current in standby mode. If the digital Input EN is Low the device is switched off (stand by mode) and if EN

Figure 1: Application Circuit Diagram

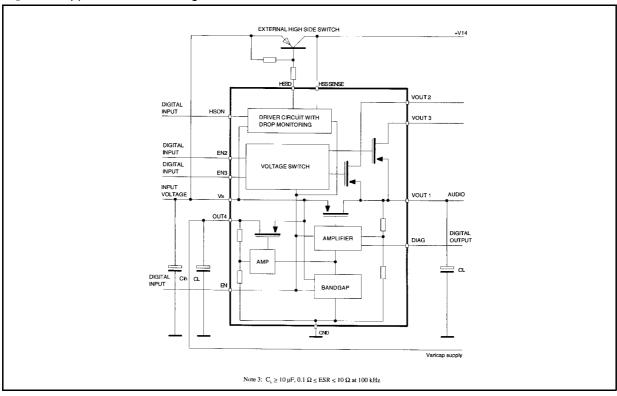


Figure 2.

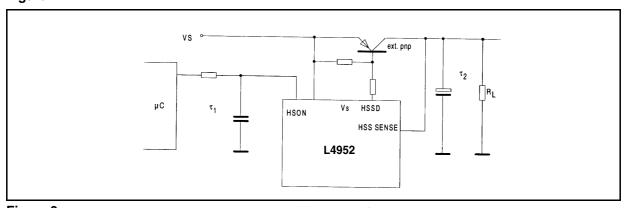
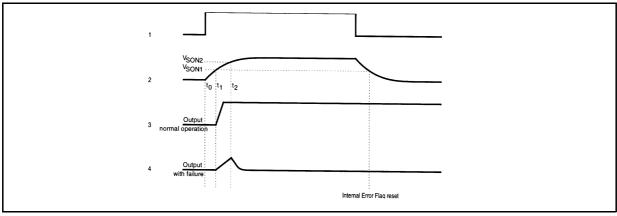


Figure 3.



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is High the device is switched on. The diagnostic circuitry detect the low drop voltage.

In this case the DIAG output is going low and can mute the power output stage to avoid noise on the loudspeaker. The two internal switches can switch the stabilized output voltage with P-MOS Transistors to one of the outputs 2 and 3 with low drop. This is useful to switch the AM and FM circuitry on or off.

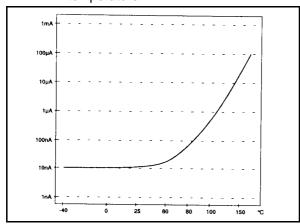
To control it there are two digital inputs En_2 and En_3 one for each switch. EN_2 is High active and EN_3 is Low active. It's possible to drive the AM/FM switch with one digital line (EN_2 and EN_3 together).

The driver for the external High side switch can switch on and off the external pnp transistor. The drop detection circuitry avoid the damage of the external power pnp transistor.

To supply the varicaps and the PLL-opamp of the car radio a second very low drop 10 V regulator is available. This regulator in dropout has a typical resistance of 50Ω .

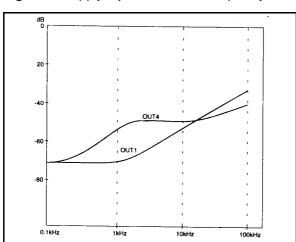
Typical Characteristics (Note 4)

Figure 3: Stand by consumption versus temperature



Supply voltage rejection (C = $10\mu F$, ESR = 4.7Ω , Load at OUT4 = $10k\Omega$)

Figure 5: Supply rejection versus Frequency.



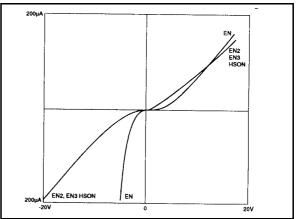
Function of the external High side switch driver

Fig 2 shows the principle circuitry of the external High side switch. Fig. 3 shows the switch on/off phase of the external High side switch. At the time t0 the microcontroller switches on (curve 1 = output signal of the microcontroller).

The signal on the HSON pin of the L4952 is shown on curve 2. At t1 the external Power pnp is switched on. At t2 ($V_{HSON} = V_{SON2}$) the internal comparator compares the drop of the external pnp. In case of normal operation the drop is smaller than V_{Dth} and no failure will be detected (curve 3).

In case of failure (that mean a higher drop than V_{Dth}) the external power pnp will be switched off (curve 4). If an error is detected it will be stored in the internal error flip-flop. The external pnp can only be switched on again after having turned HSON off ($V_{HSON} < V_{SON1}$) again.

Figure 4: Logic input currents versus input voltage



OUT2 propagation delay (Load = 100Ω)

Figure 6: OUT2 propagation delay (Load = 100Ω)

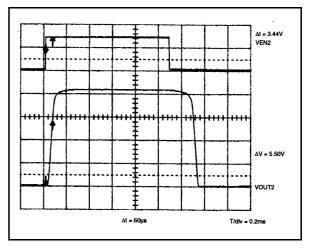


Figure 7: OUT3 propagation delay (Load = 100Ω)

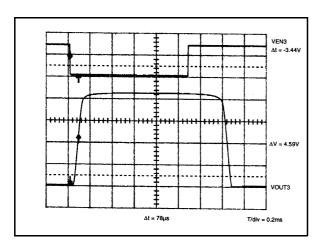


Figure 9: (C = 10μ F, ESR = 4.7Ω Load current 0mA and 117mA)

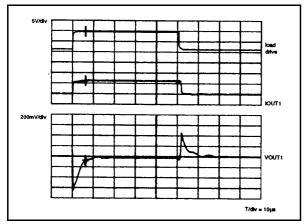
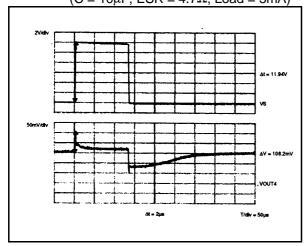


Figure 11: OUT4 supply transient response (C = 10μ F, ESR = 4.7Ω , Load = 3mA)



Note 4Typical charcteristics shown by the Figures 3 to 12 are typical parameters. Depending on produc-

Figure 8: OUT1 supply transient rejection (C = $10\mu F$, ESR = 4.7Ω , Load at OUT1 = 100Ω , Load = $10k\Omega$)

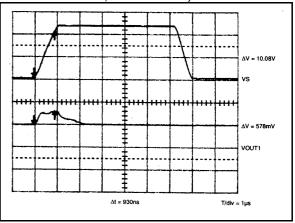


Figure 10: OUT1 voltage versus temperature

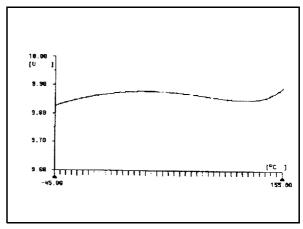
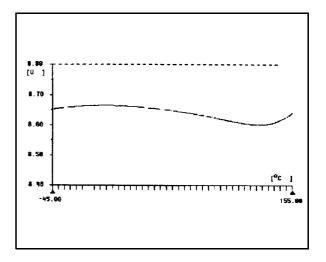


Figure 12: OUT4 voltage versus temperature

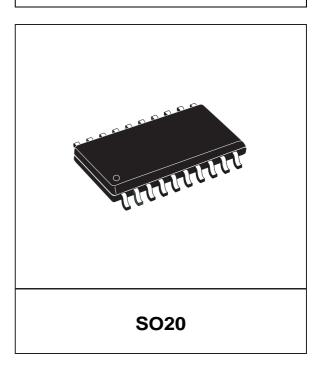


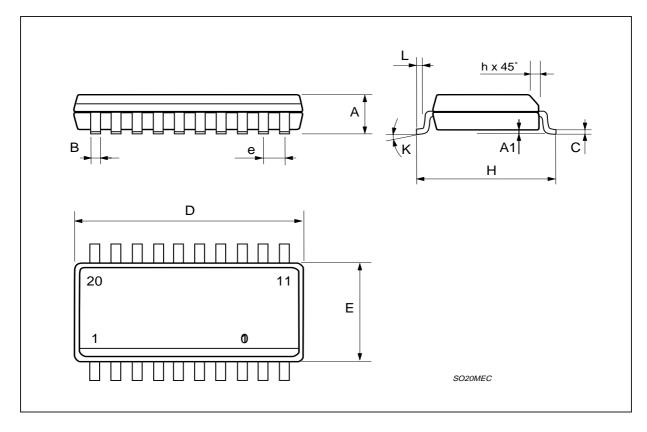
tion spread certain deviations may occure. For limits see pages 2 to 4.

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| DIM. | mm | | | inch | | |
|------|--------------------|------|-------|-------|-------|-------|
| | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| Α | 2.35 | | 2.65 | 0.093 | | 0.104 |
| A1 | 0.1 | | 0.3 | 0.004 | | 0.012 |
| В | 0.33 | | 0.51 | 0.013 | | 0.020 |
| С | 0.23 | | 0.32 | 0.009 | | 0.013 |
| D | 12.6 | | 13 | 0.496 | | 0.512 |
| E | 7.4 | | 7.6 | 0.291 | | 0.299 |
| е | | 1.27 | | | 0.050 | |
| Н | 10 | | 10.65 | 0.394 | | 0.419 |
| h | 0.25 | | 0.75 | 0.010 | | 0.030 |
| L | 0.4 | | 1.27 | 0.016 | | 0.050 |
| К | 0° (min.)8° (max.) | | | | | |

OUTLINE AND MECHANICAL DATA





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