700MHz, DIFFERENTIAL-TO-3.3V LVPECL ZERO DELAY CLOCK GENERATOR

GENERAL DESCRIPTION



The ICS8735-21 is a highly versatile 1:1 Differential-to-3.3V LVPECL clock generator and a member of the HiPerClockS™ family of High Performance Clock Solutions from ICS. The CLK, nCLK pair can accept most standard differential

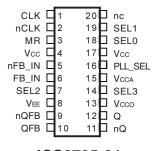
input levels. The ICS8735-21 has a fully integrated PLL and can be configured as zero delay buffer, multiplier or divider, and has an output frequency range of 31.25MHz to 700MHz. The reference divider, feedback divider and output divider are each programmable, thereby allowing for the following output-to-input frequency ratios: 8:1, 4:1, 2:1, 1:1, 1:2, 1:4, 1:8. The external feedback allows the device to achieve "zero delay" between the input clock and the output clocks. The PLL_SEL pin can be used to bypass the PLL for system test and debug purposes. In bypass mode, the reference clock is routed around the PLL and into the internal output dividers.

FEATURES

- 1 differential 3.3V LVPECL output pair,
 1 differential feedback output pair
- Differential CLK, nCLK input pair
- CLK, nCLK pair can accept the following differential input levels: LVDS, LVPECL, LVHSTL, SSTL, HCSL
- Output frequency range: 31.25MHz 700MHz
- Input frequency range: 31.25MHz 700MHz
- VCO range: 250MHz 700MHz
- Programmable dividers allow for the following output-to-input frequency ratios: 8:1, 4:1, 2:1, 1:1, 1:2, 1:4, 1:8
- External feedback for "zero delay" clock regeneration with configurable frequencies
- Cycle-to-cycle jitter: 25ps (maximum)
- Static phase offset: 50ps ± 100ps
- 3.3V supply voltage
- 0°C to 70°C ambient operating temperature
- Industrial temperature information available upon request

BLOCK DIAGRAM

PIN ASSIGNMENT



ICS8735-21

20-Lead, 300-MIL SOIC
7.5mm x 12.8mm x 2.3mm body package
M Package
Top View

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TABLE 1. PIN DESCRIPTIONS

Number	Name	Ty	/ре	Description
1	CLK	Input	Pulldown	Non-inverting differential clock input.
2	nCLK	Input	Pullup	Inverting differential clock input.
3	MR	Input	Pulldown	Master reset. Resets the output divider.
4, 17	V _{cc}	Power		Positive supply pins. Connect to 3.3V.
5	nFB_IN	Input	Pullup	Feedback input to phase detector for regenerating clocks with "zero delay". Connect to pin 9.
6	FB_IN	Input	Pulldown	Feedback input to phase detector for regenerating clocks with "zero delay". Connect to pin 10.
7	SEL2	Input	Pulldown	Determines output divider values in Table 3. LVCMOS interface levels.
8	$V_{\sf EE}$	Power		Negative supply pin. Connect to ground.
9, 10	nQFB, QFB	Output		Differential feedback outputs. LVPECL interface levels.
11, 12	nQ, Q	Output		Differential clock outputs. LVPECL interface levels.
13	V _{cco}	Power		Output supply pin. Connect to 3.3V.
14	SEL3	Input	Pulldown	Determines output divider values in Table 3. LVCMOS interface levels.
15	V _{CCA}	Power		Analog supply pin. Connect to 3.3V.
16	PLL_SEL	Input	Pullup	Selects between the PLL and reference clock as the input to the dividers. When LOW, selects reference clock. When HIGH, selects PLL. LVCMOS interface levels.
18	SEL0	Input	Pulldown	Determines output divider values in Table 3. LVCMOS interface levels.
19	SEL1	Input	Pulldown	Determines output divider values in Table 3. LVCMOS interface levels.
20	nc	Unused		No connect.

NOTE: Pullup and Pulldown refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance				4	pF
R _{PULLUP}	Input Pullup Resistor			51		ΚΩ
R _{PULLDOWN}	Input Pulldown Resistor			51		ΚΩ

TABLE 3A. CONTROL INPUT FUNCTION TABLE

			Inputs		Outputs PLL_SEL = 1 PLL Enable Mode
SEL3	SEL2	SEL1	SEL0	Reference Frequency Range (MHz)*	Q, nQ; QFB, nQFB
0	0	0	0	250 - 700	÷ 1
0	0	0	1	125 - 350	÷ 1
0	0	1	0	62.5 - 175	÷ 1
0	0	1	1	31.25 - 87.5	÷ 1
0	1	0	0	250 - 700	÷ 2
0	1	0	1	125 - 350	÷ 2
0	1	1	0	62.5 - 175	÷ 2
0	1	1	1	250 - 700	÷ 4
1	0	0	0	125 - 350	÷ 4
1	0	0	1	250 - 700	÷ 8
1	0	1	0	125 - 350	x 2
1	0	1	1	62.5 - 175	x 2
1	1	0	0	31.25 - 87.5	x 2
1	1	0	1	62.5 - 175	x 4
1	1	1	0	31.25 - 87.5	x 4
1	1	1	1	31.25 - 87.5	x 8

*NOTE: VCO frequency range for all configurations above is 250 to 700MHz.

TABLE 3B. PLL BYPASS FUNCTION TABLE

	Inp	Outputs PLL_SEL = 0 PLL Bypass Mode		
SEL3	SEL2	SEL1	SEL0	Q, nQ; QFB, nQFB
0	0	0	0	÷ 4
0	0	0	1	÷ 4
0	0	1	0	÷ 4
0	0	1	1	÷ 8
0	1	0	0	÷ 8
0	1	0	1	÷ 8
0	1	1	0	÷ 16
0	1	1	1	÷ 16
1	0	0	0	÷ 32
1	0	0	1	÷ 64
1	0	1	0	÷ 2
1	0	1	1	÷ 2
1	1	0	0	÷ 4
1	1	0	1	÷ 1
1	1	1	0	÷ 2
1	1	1	1	÷ 1

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ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V_{CCx} 4.6V

 $\begin{array}{ll} \text{Inputs, V}_{\text{I}} & -0.5\text{V to V}_{\text{CC}} + 0.5\text{ V} \\ \text{Outputs, V}_{\text{O}} & -0.5\text{V to V}_{\text{CCO}} + 0.5\text{V} \\ \text{Package Thermal Impedance, } \theta_{\text{JA}} & 46.2^{\circ}\text{C/W (0 Ifpm)} \\ \text{Storage Temperature, T}_{\text{STG}} & -65^{\circ}\text{C to } 150^{\circ}\text{C} \end{array}$

Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Table 4A. Power Supply DC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{cc}	Positive Supply Voltage		3.135	3.3	3.465	V
V _{CCA}	Analog Supply Voltage		3.135	3.3	3.465	V
V _{cco}	Output Supply Voltage		3.135	3.3	3.465	V
I _{EE}	Power Supply Current				150	mA
I _{CCA}	Analog Supply Current				15	mA

Table 4B. LVCMOS/LVTTL DC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V _{IH}	Input High Voltage			2		V _{cc} + 0.3	V
V _{IL}	Input Low Voltage			-0.3		0.8	V
I _{IH}	Input High Current	SEL0, SEL1, SEL2, SEL3, MR	$V_{CC} = V_{IN} = 3.465V$			150	μΑ
IH IH		PLL_SEL	$V_{CC} = V_{IN} = 3.465V$			5	μΑ
I _{II}	Input Low Current	SEL0, SEL1, SEL2, SEL3, MR	$V_{CC} = 3.465V, V_{IN} = 0V$	-5			μΑ
I IL		PLL_SEL	$V_{CC} = 3.465V, V_{IN} = 0V$	-150			μΑ

Table 4C. Differential DC Characteristics, $V_{cc} = V_{cca} = V_{cco} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
	Input High Current	CLK, FB_IN	$V_{CC} = V_{IN} = 3.465V$			150	μΑ
I'IH	Input High Current	nCLK, nFB_IN	$V_{CC} = V_{IN} = 3.465V$			5	μΑ
	Innut Low Current	CLK, FB_IN	$V_{CC} = 3.465V, V_{IN} = 0V$	-5			μΑ
I _{IL}	Input Low Current	nCLK, nFB_IN	$V_{CC} = 3.465V, V_{IN} = 0V$	-150			μΑ
V _{PP}	Peak-to-Peak Input	Voltage		0.15		1.3	V
V _{CMR}	Common Mode Inp	ut Voltage; NOTE 1, 2		V _{EE} + 0.5		V _{cc} - 0.85	V

NOTE 1: Common mode voltage is defined as $V_{\rm IH}$.

NOTE 2: For single ended applications, the maximum input voltage for CLK, nCLK is V_{cc} + 0.3V.

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Table 4D. LVPECL DC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{OH}	Output High Voltage; NOTE 1		V _{cco} - 1.4		V _{cco} - 1.0	V
V _{OL}	Output Low Voltage; NOTE 1		V _{cco} - 2.0		V _{cco} - 1.7	V
V _{SWING}	Peak-to-Peak Output Voltage Swing		0.6		0.9	V

NOTE 1: Outputs terminated with 50 $\!\Omega$ to V $_{\!\scriptscriptstyle CCO}$ - 2V.

Table 5. Input Frequency Characteristics, $V_{cc} = V_{cca} = V_{cco} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f	Input Fraguanay	CLK =CLK	PLL_SEL = 1	31.25		700	MHz
I _{IN}	Input Frequency	CLK, nCLK	PLL_SEL = 0			700	MHz

Table 6. AC Characteristics, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, Ta = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f _{MAX}	Output Frequency				700	MHz
t _{PD}	Propagation Delay; NOTE 1	PLL_SEL = 0V, f ≤ 700MHz	3.6	3.9	4.2	ns
tsk(o)	Output Skew; NOTE 4, 5	PLL_SEL = 0V			20	ps
t(Ø)	Static Phase Offset; NOTE 2, 5	PLL_SEL = 3.3V	-50	50	150	ps
tjit(cc)	Cycle-to-Cycle Jitter; NOTE 5, 6				25	ps
<i>t</i> jit(θ)	Phase Jitter; NOTE 3, 5, 6				±50	ps
t_	PLL Lock Time				1	ms
t _R	Output Rise Time	20% to 80% @ 50MHz	300		700	ps
t _F	Output Fall Time	20% to 80% @ 50MHz	300		700	ps
odc	Output Duty Cycle		47		53	%

All parameters measured at f_{MAX} unless noted otherwise.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as the time difference between the input reference clock and the average feedback input signal,

when the PLL is locked and the input reference frequency is stable.

NOTE 3: Phase jitter is dependent on the input source used.

NOTE 4: Defined as skew between outputs at the same supply voltage and with equal load conditions.

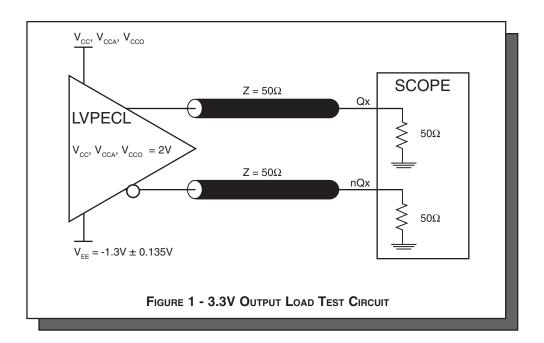
Measured at the output differential crosspoints.

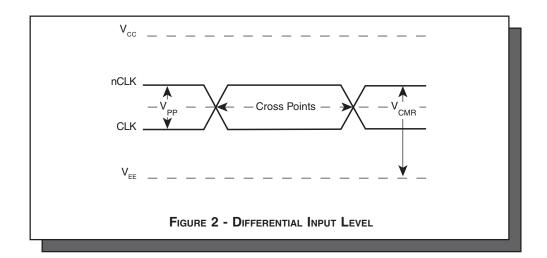
NOTE 5: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 6: Characterized at VCO frequency of 622MHz.

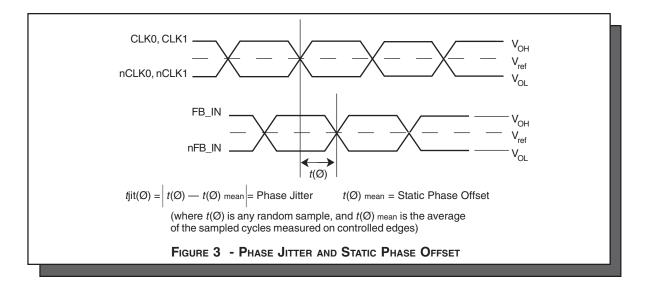


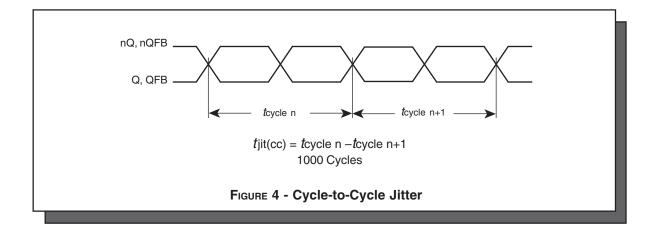
PARAMETER MEASUREMENT INFORMATION

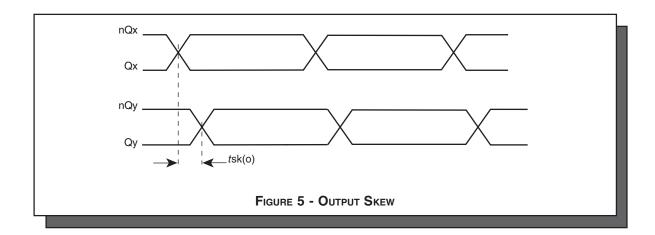




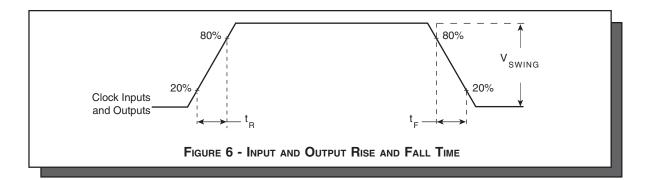
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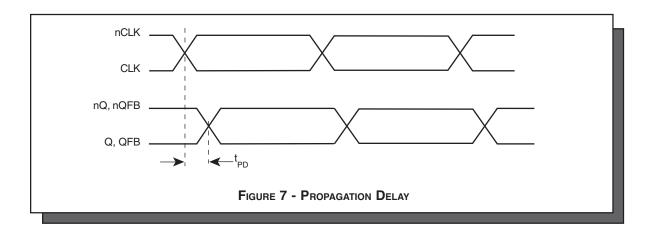


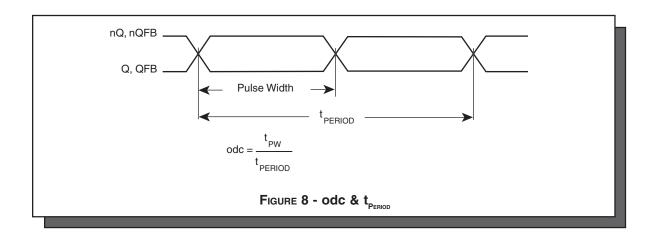




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APPLICATION INFORMATION

Power Supply Filtering Techniques

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS8735-21 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $V_{\rm CC}$, $V_{\rm CCA}$, and $V_{\rm CCO}$ should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. *Figure 9* illustrates how a 10Ω resistor along with a $10\mu F$ and a $.01\mu F$ bypass capacitor should be connected to each $V_{\rm CCA}$ pin.

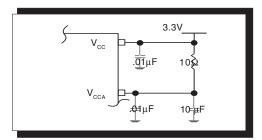
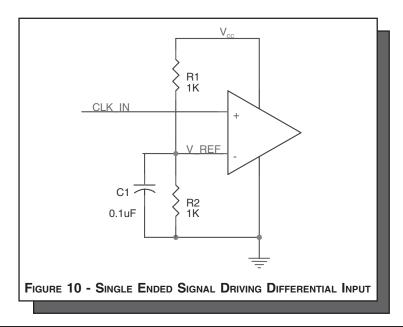


FIGURE 9. POWER SUPPLY FILTERING

WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 10 shows how the differential input can be wired to accept single ended levels. The reference voltage V_REF $_{\sim}$ V $_{cc}$ /2 is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio of R1 and R2 might need to be adjusted to position the V_REF in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and V $_{cc}$ = 3.3V, V_REF should be 1.25V and R2/R1 = 0.609.



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POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS8735-21. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS8735-21 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC MAX} * I_{EE MAX} = 3.465V * 150mA = 519.8mW
- Power (outputs)_{MAX} = 30.2mW/Loaded Output pair
 If all outputs are loaded, the total power is 2 * 30.2mW = 60.4mW

Total Power_{MAX} (3.465V, with all outputs switching) = 519.8mW + 60.4mW = 580.2mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS TM devices is 125°C.

The equation for Tj is as follows: $Tj = \theta_{JA} * Pd_total + T_A$

Tj = Junction Temperature

 θ_{JA} = junction-to-ambient thermal resistance

Pd_total = Total device power dissipation (example calculation is in section 1 above)

 $T_{\Delta} = Ambient Temperature$

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used . Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 39.7°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 70° C with all outputs switching is: 70° C + 0.580W * 39.7° C/W = 93° C. This is well below the limit of 125° C

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

Table 7. Thermal Resistance θ_{JA} for 20-pin SOIC, Forced Convection

0 200 500 Single-Layer PCB, JEDEC Standard Test Boards 83.2°C/W 65.7°C/W 57.5°C/W Multi-Layer PCB, JEDEC Standard Test Boards 46.2°C/W 39.7°C/W 36.8°C/W

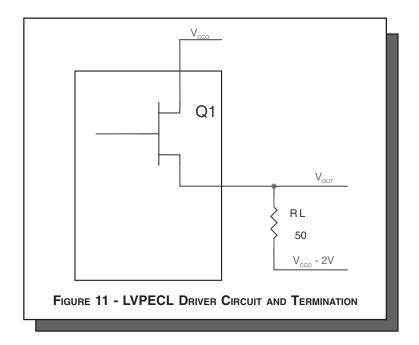
NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

 θ_{1A} by Velocity (Linear Feet per Minute)

3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 11.



To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{CC} - 2V.

• For logic high,
$$V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 1.0V$$

$$(V_{CCO_MAX} - V_{OH_MAX}) = 1.0V$$

For logic low,
$$V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.7V$$

$$(V_{CCO_MAX} - V_{OL_MAX}) = 1.7V$$

Pd_H is power dissipation when the output drives high. Pd_L is the power dissipation when the output drives low.

$$Pd_{-}H = [(V_{OH_MAX} - (V_{CCO_MAX} - 2V))/R_{_{L}}] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - (V_{CCO_MAX} - V_{OH_MAX}))/R_{_{L}}] * (V_{CCO_MAX} - V_{OH_MAX}) = (2V - 1V) * 1V = 20.0mW$$

$$Pd_L = [(V_{OL_MAX} - (V_{CCO_MAX} - 2V))/R_{L}] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - (V_{CCO_MAX} - V_{OL_MAX}))/R_{L}] * (V_{CCO_MAX} - V_{OL_MAX}) = (2V - 1.7V) * 1.7V = 10.2mW$$

Total Power Dissipation per output pair = Pd_H + Pd_L = 30.2mW

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RELIABILITY INFORMATION

Table 8. $\theta_{JA} \text{vs. A} \text{IR Flow Table}$

$\boldsymbol{\theta}_{\text{JA}}$ by Velocity (Linear Feet per Minute)

 O
 200
 500

 Single-Layer PCB, JEDEC Standard Test Boards
 83.2°C/W
 65.7°C/W
 57.5°C/W

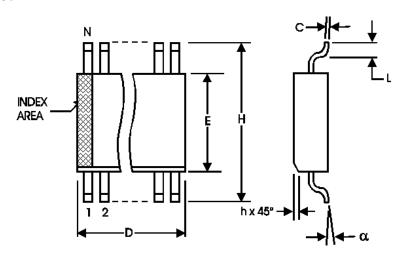
 Multi-Layer PCB, JEDEC Standard Test Boards
 46.2°C/W
 39.7°C/W
 36.8°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

TRANSISTOR COUNT

The transistor count for ICS8735-21 is: 2969

PACKAGE OUTLINE - M SUFFIX



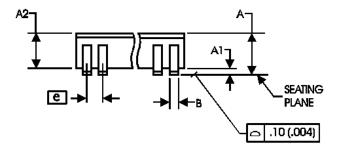


TABLE 9. PACKAGE DIMENSIONS

SYMBOL	Millim	neters
STINIBOL	Minimum	Maximum
N	2	0
Α		2.65
A1	0.10	
A2	2.05	2.55
В	0.33	0.51
С	0.18	0.32
D	12.60	13.00
E	7.40	7.60
е	1.27 E	BASIC
Н	10.00	10.65
h	0.25	0.75
L	0.40	1.27
α	0°	8°

Reference Document: JEDEC Publication 95, MS-013, MO-119



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TABLE 10. ORDERING INFORMATION

Part/Order Number	Marking	Package	Count	Temperature
ICS8735AM-21	ICS8735AM-21	20 Lead SOIC	38 per tube	0°C to 70°C
ICS8735AM-21T	ICS8735AM-21	20 Lead SOIC on Tape and Reel	1000	0°C to 70°C

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REVISION HISTORY SHEET				
Rev	Table	Page	Description of Change	Date
В		1	Revised Block Diagram.	
	6	5	Added Output Skew row at 20ps Max.	10/31/01
	6	5	Relabled PLL Reference Zero Delay to Static Phase Offset.	10/31/01
		7	Added Output Skew Diagram.	
В	3A	3	Added note at bottom of the table.	11/20/01
	6	5	Added Note 6.	11/20/01