



INA2141

Dual, Low Power, $G = 10, 100$ INSTRUMENTATION AMPLIFIER

FEATURES

- **LOW OFFSET VOLTAGE:** $50\mu\text{V}$ max
- **LOW DRIFT:** $0.5\mu\text{V}/^\circ\text{C}$ max
- **EXCELLENT GAIN ACCURACY:**
 $\pm 0.05\%$ max at $G = 10$
- **LOW INPUT BIAS CURRENT:** 5nA max
- **HIGH CMR:** 117dB min ($G = 100$)
- **INPUTS PROTECTED TO $\pm 40\text{V}$**
- **WIDE SUPPLY RANGE:** $\pm 2.25\text{V}$ to $\pm 18\text{V}$
- **LOW QUIESCENT CURRENT:** $750\mu\text{A}/\text{IA}$
- **16-PIN PLASTIC DIP, SOL-16**

APPLICATIONS

- **SENSOR AMPLIFIER**
THERMOCOUPLE, RTD, BRIDGE
- **MEDICAL INSTRUMENTATION**
- **MULTIPLE CHANNEL SYSTEMS**

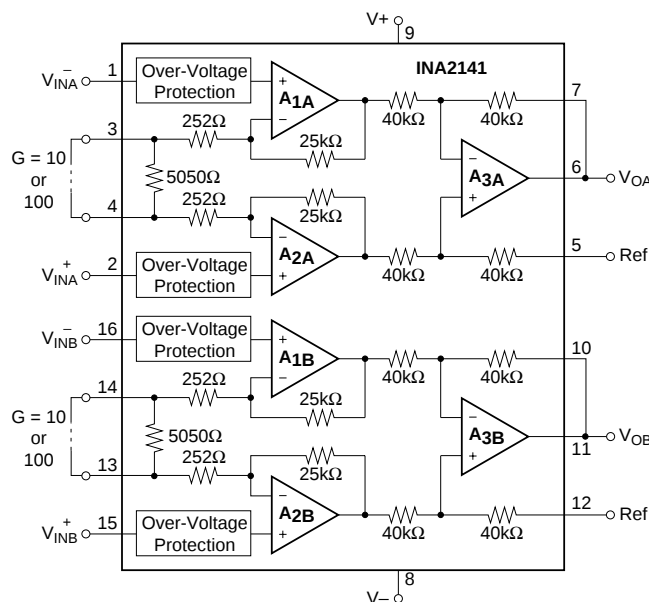
DESCRIPTION

The INA2141 is a low power, dual instrumentation amplifier offering excellent accuracy. Its versatile 3-op amp design and small size make it ideal for a wide range of applications. Current-feedback input circuitry provides wide bandwidth even at high gain (200kHz at $G = 100$).

Simple pin connections set an accurate gain of 10 or 100V/V without external resistors. Internal input protection can withstand up to $\pm 40\text{V}$ without damage.

The INA2141 is laser trimmed for very low offset voltage ($50\mu\text{V}$), drift ($0.5\mu\text{V}/^\circ\text{C}$) and high common-mode rejection (117dB at $G = 100$). It operates with power supplies as low as $\pm 2.25\text{V}$, and quiescent current is only $750\mu\text{A}$ per amplifier—ideal for battery operated systems.

Packages are 16-pin plastic DIP, and SOL-16 surface-mount, specified for the -40°C to $+85^\circ\text{C}$ temperature range.



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SPECIFICATIONS

At $T_A = +25^{\circ}\text{C}$, $V_S = \pm 15\text{V}$, and $R_L = 10\text{k}\Omega$, unless otherwise noted.

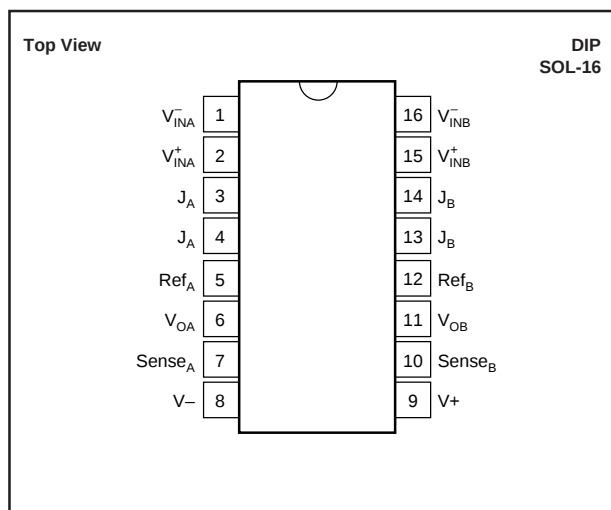
PARAMETER	CONDITIONS	INA2141P, U			INA2141PA, UA			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
INPUT								
Offset Voltage, RTI	$G = 100$		± 20	± 50		*	± 125	μV
vs Temperature	$G = 10$		± 50	± 100		*	± 250	μV
	$G = 100$		± 0.2	± 0.5		*	± 1.5	$\mu\text{V}/^{\circ}\text{C}$
	$G = 10^{(2)}$		± 0.5	± 2		*	± 5	$\mu\text{V}/^{\circ}\text{C}$
vs Power Supply	$V_S = \pm 2.25$ to $\pm 18\text{V}$, $G = 100$		± 1	± 2		*	± 5	$\mu\text{V/V}$
	$G = 10$		± 2	± 10		*	± 20	$\mu\text{V/V}$
Long-Term Stability	$G = 100$		0.2			*		$\mu\text{V/mo}$
	$G = 10$		0.5			*		$\mu\text{V/mo}$
Impedance, Differential			$10^{10} \parallel 2$			*		$\Omega \parallel \text{pF}$
Common-Mode			$10^{10} \parallel 9$			*		$\Omega \parallel \text{pF}$
Common-Mode Voltage Range ⁽¹⁾	$V_O = 0\text{V}$	$(V+) - 2$ $(V-) + 2$	$(V+) - 1.4$ $(V-) + 1.7$		*	*		V
Safe Input Voltage				± 40			*	V
Common-Mode Rejection	$V_{CM} = \pm 13\text{V}$, $\Delta R_S = 1\text{k}\Omega$							V
	$G = 100$	117	125		110	120		dB
	$G = 10$	97	106		93	100		dB
BIAS CURRENT								
vs Temperature			± 2	± 5		*	± 10	nA
Offset Current			± 30			*		$\text{pA}/^{\circ}\text{C}$
vs Temperature			± 1	± 5		*	± 10	nA
			± 30			*		$\text{pA}/^{\circ}\text{C}$
NOISE VOLTAGE, RTI								
$f = 10\text{Hz}$	$G = 100$, $R_S = 0\Omega$		10			*		$\text{nV}/\sqrt{\text{Hz}}$
$f = 100\text{Hz}$			8			*		$\text{nV}/\sqrt{\text{Hz}}$
$f = 1\text{kHz}$			8			*		$\text{nV}/\sqrt{\text{Hz}}$
$f_B = 0.1\text{Hz}$ to 10Hz			0.2			*		$\mu\text{Vp-p}$
$f = 10\text{Hz}$	$G = 10$, $R_S = 0\Omega$		22			*		$\text{nV}/\sqrt{\text{Hz}}$
$f = 100\text{Hz}$			13			*		$\text{nV}/\sqrt{\text{Hz}}$
$f = 1\text{kHz}$			12			*		$\text{nV}/\sqrt{\text{Hz}}$
$f_B = 0.1\text{Hz}$ to 10Hz			0.6			*		$\mu\text{Vp-p}$
Noise Current								
$f = 10\text{Hz}$			0.9			*		$\text{pA}/\sqrt{\text{Hz}}$
$f = 1\text{kHz}$			0.3			*		$\text{pA}/\sqrt{\text{Hz}}$
$f_B = 0.1\text{Hz}$ to 10Hz			30			*		pAp-p
GAIN								
Gain Error	$V_O = \pm 13.6\text{V}$, $G = 100$		± 0.03	± 0.075		*	± 0.15	%
	$G = 10$		± 0.01	± 0.05		*	± 0.15	%
Gain vs Temperature ⁽²⁾	$G = 10$, 100		± 2	± 10		*	*	$\text{ppm}/^{\circ}\text{C}$
Nonlinearity	$G = 100$		± 0.0005	± 0.002		*	± 0.004	% of FSR
	$G = 10$		± 0.0003	± 0.001		*	± 0.002	% of FSR
OUTPUT								
Voltage: Positive	$R_L = 10\text{k}\Omega$	$(V+) - 1.4$	$(V+) - 0.9$		*	*		V
Negative	$R_L = 10\text{k}\Omega$	$(V-) + 1.4$	$(V-) + 0.9$		*	*		V
Load Capacitance Stability			1000			*		pF
Short-Circuit Current			+6/-15			*		mA
FREQUENCY RESPONSE								
Bandwidth, -3dB	$G = 100$		200			*		kHz
	$G = 10$		1			*		MHz
Slew Rate	$V_O = \pm 10\text{V}$, $G = 10$		4			*		V/ μs
Settling Time, 0.01%	$V_O = \pm 5\text{V}$, $G = 100$		9			*		μs
	$G = 10$		7			*		μs
Overload Recovery	50% Overdrive		4			*		μs
POWER SUPPLY								
Voltage Range		± 2.25	± 15	± 18	*	*	*	V
Current, Total	$V_{IN} = 0\text{V}$		± 1.5	± 1.6		*	*	mA
TEMPERATURE RANGE								
Specification		-40		85	*		*	$^{\circ}\text{C}$
Operating		-40		125	*		*	$^{\circ}\text{C}$
θ_{JA}			80			*		$^{\circ}\text{C/W}$

* Specification same as INA2141P, U.

NOTE: (1) Input common-mode range varies with output voltage—see typical curves. (2) Guaranteed by wafer test.

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PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS

Supply Voltage	$\pm 18V$
Analog Input Voltage Range	$\pm 40V$
Output Short-Circuit (to ground)	Continuous
Operating Temperature	$-40^{\circ}C$ to $+125^{\circ}C$
Storage Temperature	$-40^{\circ}C$ to $+125^{\circ}C$
Junction Temperature	$+150^{\circ}C$
Lead Temperature (soldering, 10s)	$+300^{\circ}C$



ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

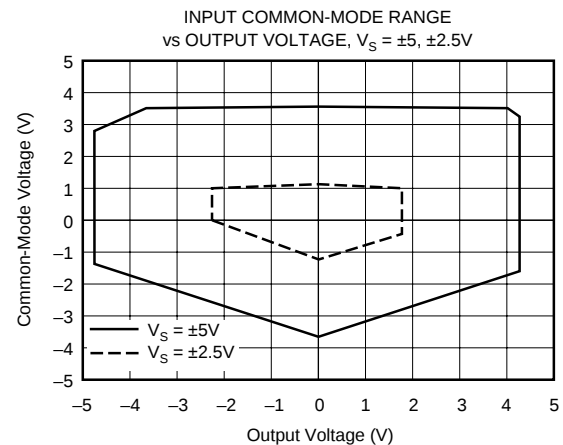
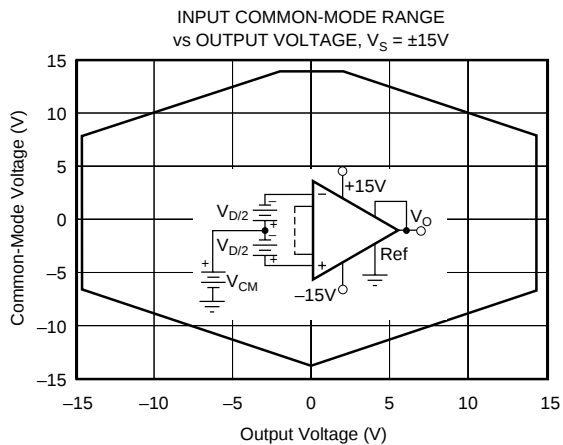
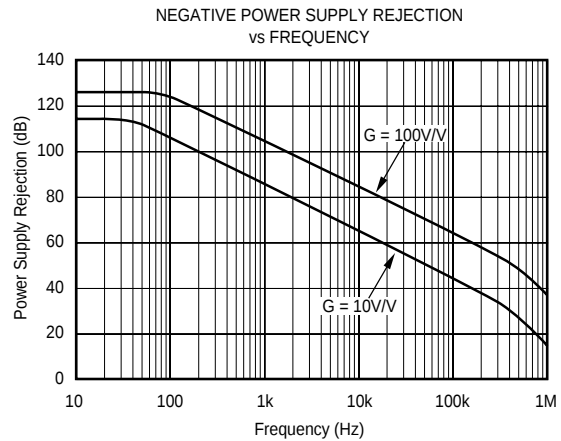
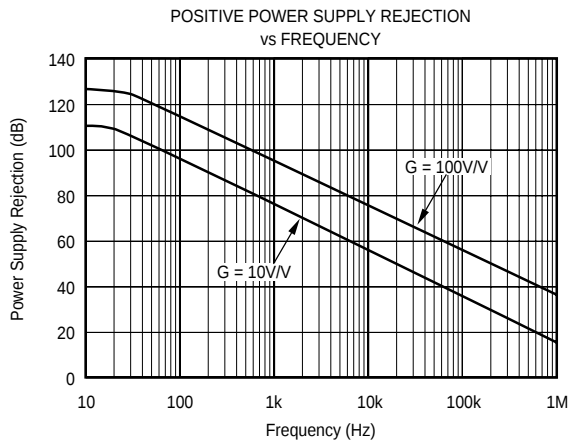
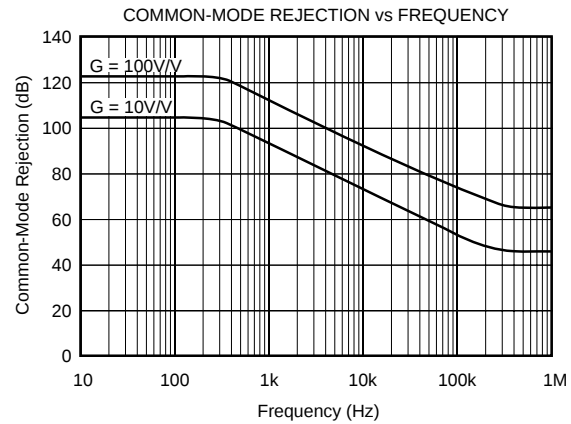
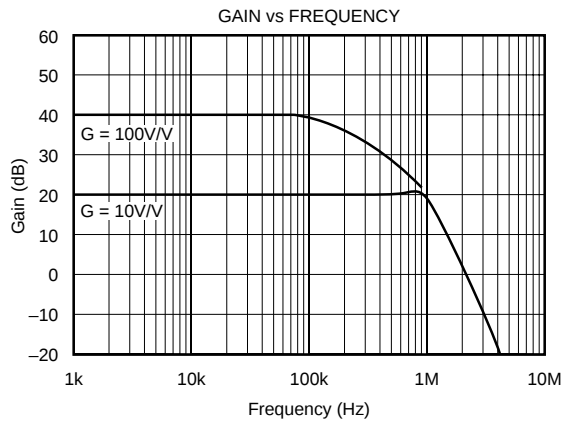
ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER ⁽¹⁾	TEMPERATURE RANGE
INA2141PA	16-Pin Plastic DIP	180	$-40^{\circ}C$ to $+85^{\circ}C$
INA2141P	16-Pin Plastic DIP	180	$-40^{\circ}C$ to $+85^{\circ}C$
INA2141UA	SOL-16 Surface-Mount	211	$-40^{\circ}C$ to $+85^{\circ}C$
INA2141U	SOL-16 Surface-Mount	211	$-40^{\circ}C$ to $+85^{\circ}C$

NOTE: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book.

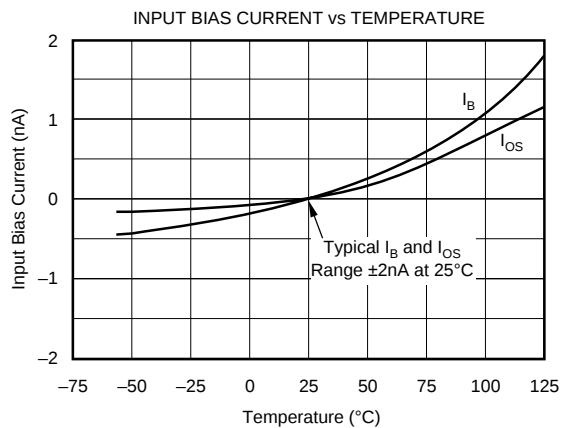
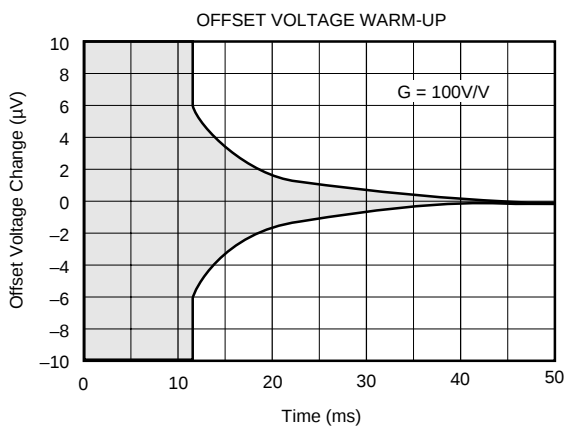
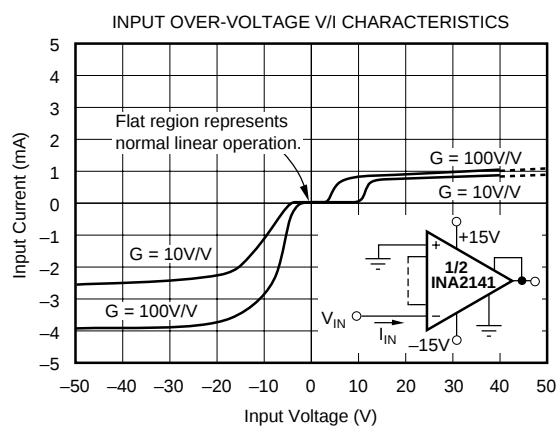
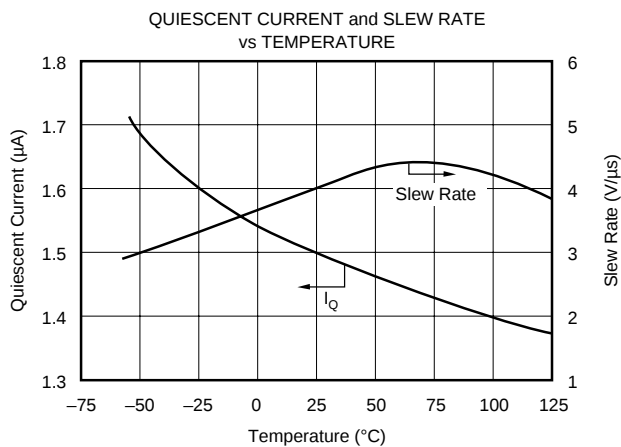
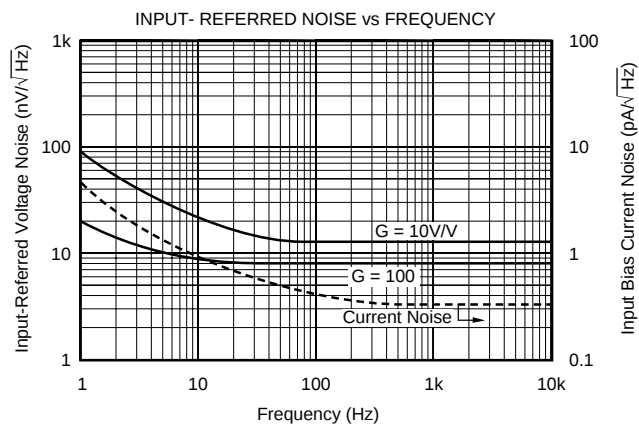
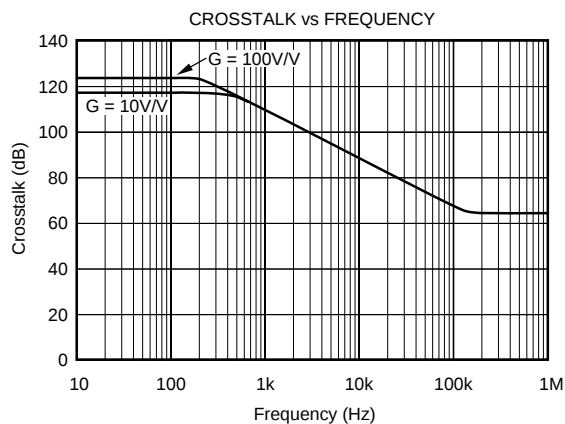
TYPICAL PERFORMANCE CURVES

At $T_A = +25^\circ\text{C}$, and $V_S = \pm 15\text{V}$, unless otherwise noted.



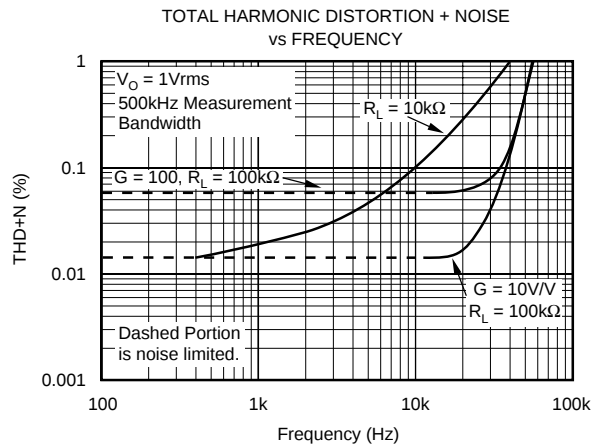
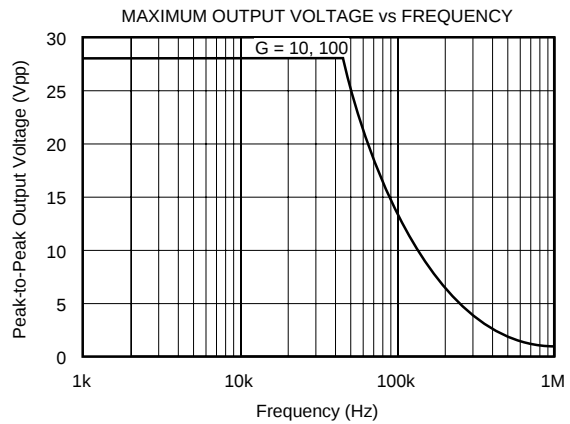
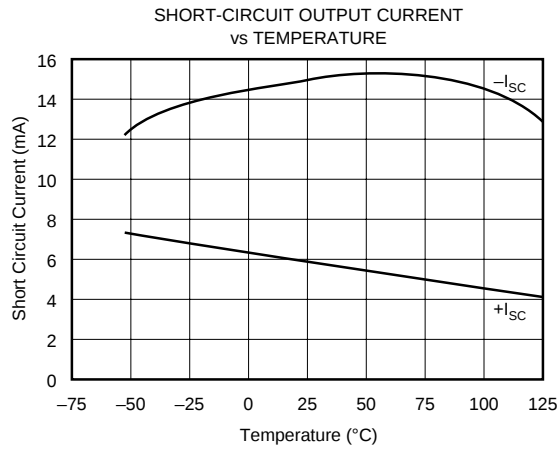
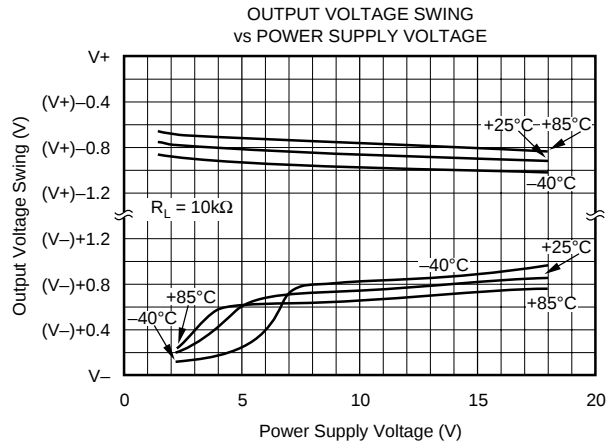
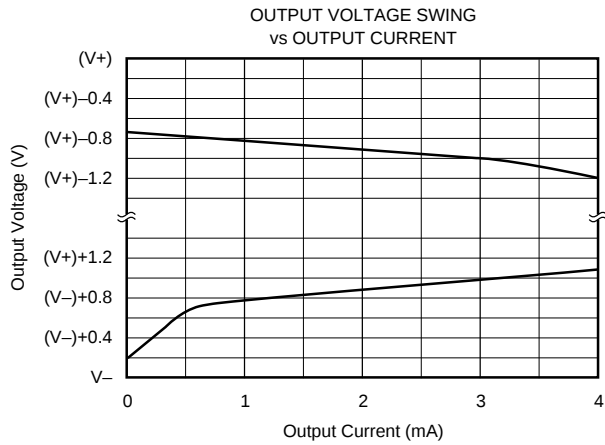
TYPICAL PERFORMANCE CURVES (CONT)

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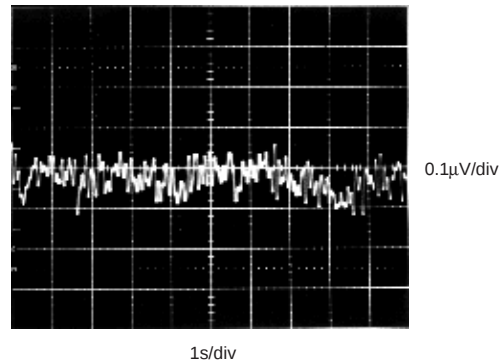


TYPICAL PERFORMANCE CURVES (CONT)

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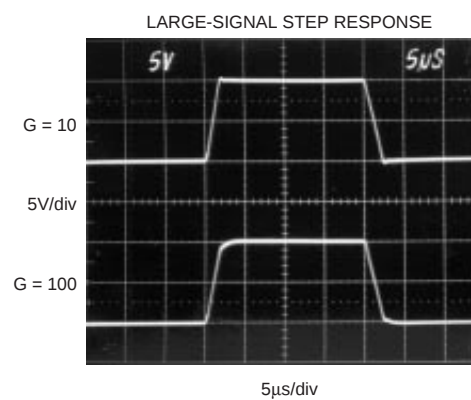
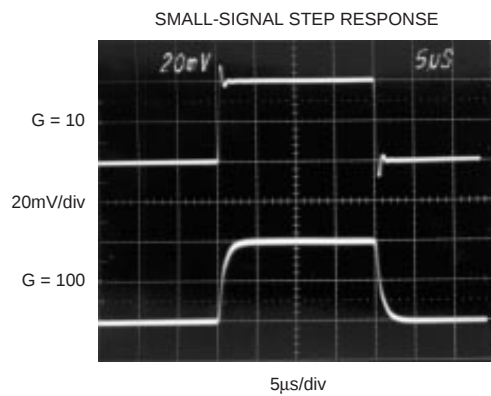


VOLTAGE NOISE 0.1 to 10Hz
INPUT-REFERRED, $G = 100$



TYPICAL PERFORMANCE CURVES (CONT)

At $T_A = +25^\circ\text{C}$, and $V_S = \pm 15\text{V}$ unless otherwise noted.



APPLICATION INFORMATION

Figure 1 shows the basic connections required for operation of the INA2141. Applications with noisy or high impedance power supplies may require decoupling capacitors close to the device pins as shown.

The output is referred to the output reference (Ref) terminals (Ref_A and Ref_B) which are normally grounded. These must be low-impedance connections to assure good common-mode rejection. A resistance of 8Ω in series with a Ref pin will cause a typical device to degrade to approximately 80dB CMR (G = 1).

The INA2141 has a separate output sense feedback connections Sense_A and Sense_B. These must be connected to their respective output terminals for proper operation. The output sense connection can be used to sense the output voltage directly at the load for best accuracy.

SETTING THE GAIN

Gain of each IA can be independently selected with a jumper connection as shown in Figure 1. $G = 10\text{V/V}$ with no jumper installed. With a jumper installed $G = 100\text{V/V}$. To preserve good gain accuracy, this jumper must have low series resistance. A resistance of 0.5Ω in series with the jumper will decrease the gain by 0.1%.

Internal resistor ratios are laser trimmed to assure excellent gain accuracy. Actual resistor values can vary by approximately $\pm 25\%$ from the nominal values shown.

Gains between 10 and 100 can be achieved by connecting an external resistor to the jumper pins. This is not recommended, however, because the $\pm 25\%$ variation of internal resistor values makes the required external resistor value uncertain. A companion model, INA2128, features accurately trimmed internal resistors so that gains from 1 to 10,000 can be set with an external resistor.

DYNAMIC PERFORMANCE

The typical performance curve "Gain vs Frequency" shows that despite its low quiescent current, the INA2141 achieves wide bandwidth, even at high gain. This is due to its current-feedback topology. Settling time also remains excellent at high gain.

NOISE PERFORMANCE

The INA2141 provides very low noise in most applications. Low frequency noise is approximately $0.2\mu\text{Vp-p}$ measured from 0.1 to 10Hz ($G = 100$). This provides dramatically improved noise when compared to state-of-the-art chopper-stabilized amplifiers.

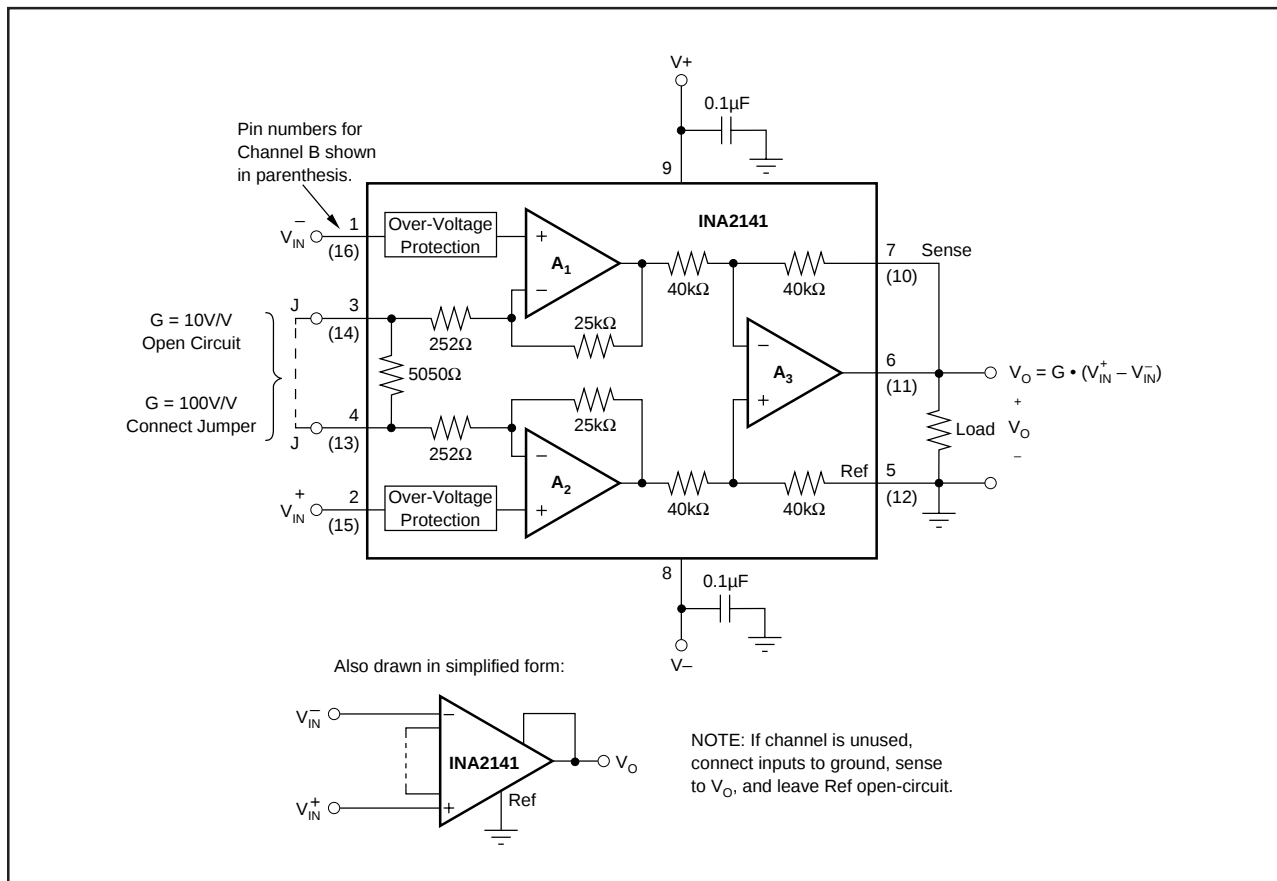


FIGURE 1. Basic Connections.

OFFSET TRIMMING

The INA2141 is laser trimmed for low offset voltage and offset voltage drift. Most applications require no external offset adjustment. Figure 2 shows an optional circuit for trimming the output offset voltage. The voltage applied to Ref terminal is summed with the output. The op amp buffer provides low impedance at the Ref terminal to preserve good common-mode rejection.

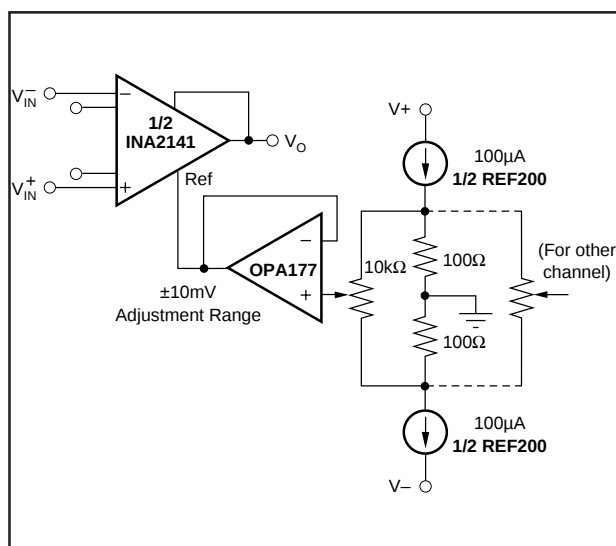


FIGURE 2. Optional Trimming of Output Offset Voltage.

INPUT BIAS CURRENT RETURN PATH

The input impedance of the INA2141 is extremely high—approximately $10^{10}\Omega$. However, a path must be provided for the input bias current of both inputs. This input bias current is approximately $\pm 2\text{nA}$. High input impedance means that this input bias current changes very little with varying input voltage.

Input circuitry must provide a path for this input bias current for proper operation. Figure 3 shows various provisions for an input bias current path. Without a bias current path, the inputs will float to a potential which exceeds the common-mode range of the INA2141 and the input amplifiers will saturate.

If the differential source resistance is low, the bias current return path can be connected to one input (see the thermocouple example in Figure 3). With higher source impedance, using two equal resistors provides a balanced input with possible advantages of lower input offset voltage due to bias current and better high-frequency common-mode rejection.

INPUT COMMON-MODE RANGE

The linear input voltage range of the input circuitry of the INA2141 is from approximately 1.4V below the positive supply voltage to 1.7V above the negative supply. As a differential input voltage causes the output voltage increase, however, the linear input range will be limited by the output voltage swing of amplifiers A_1 and A_2 . So the linear com-

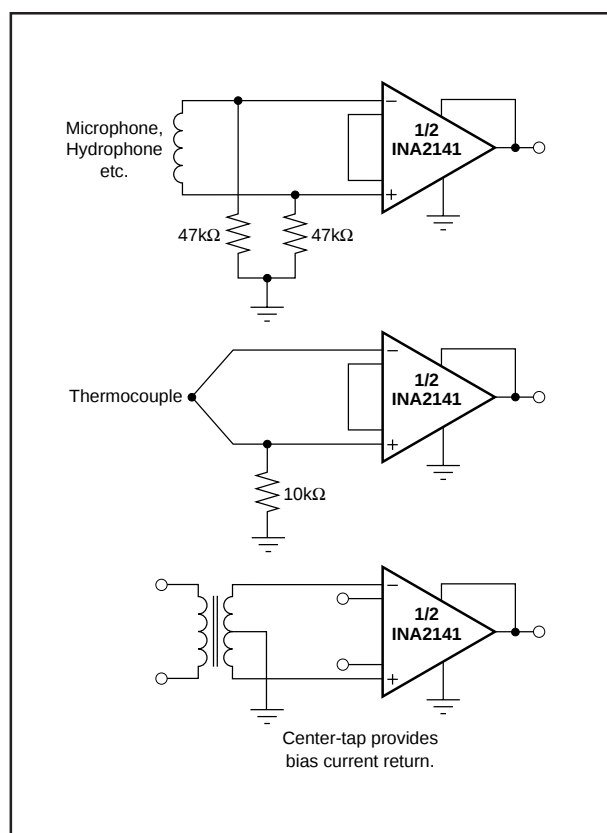


FIGURE 3. Providing an Input Common-Mode Current Path.

mon-mode input range is related to the output voltage of the complete amplifier. This behavior also depends on supply voltage—see performance curves “Input Common-Mode Range vs Output Voltage”.

Input-overload can produce an output voltage that appears normal. For example, if an input overload condition drives both input amplifiers to their positive output swing limit, the difference voltage measured by the output amplifier will be near zero. The output of the INA2141 will be near 0V even though both inputs are overloaded.

LOW VOLTAGE OPERATION

The INA2141 can be operated on power supplies as low as $\pm 2.25\text{V}$. Performance remains excellent with power supplies ranging from $\pm 2.25\text{V}$ to $\pm 18\text{V}$. Most parameters vary only slightly throughout this supply voltage range—see typical performance curves. Operation at very low supply voltage requires careful attention to assure that the input voltages remain within their linear range. Voltage swing requirements of internal nodes limit the input common-mode range with low power supply voltage. Typical performance curves, “Input Common-Mode Range vs Output Voltage” show the range of linear operation for $\pm 15\text{V}$, $\pm 5\text{V}$, and $\pm 2.5\text{V}$ supplies.

INPUT PROTECTION

The inputs of the INA2141 are individually protected for voltages up to $\pm 40\text{V}$. For example, a condition of -40V on one input and $+40\text{V}$ on the other input will not cause damage. Internal circuitry on each input provides low series impedance under normal signal conditions. To provide equivalent protection, series input resistors would contribute excessive noise. If the input is overloaded, the protection circuitry limits the input current to a safe value of approximately 1.5 to 5mA. The typical performance curve “Input Bias Current vs Common-Mode Input Voltage” shows this input current limit behavior. The inputs are protected even if the power supplies are disconnected or turned off.

CHANNEL CROSSTALK

The two channels of the INA2141 are completely independent, including all bias circuitry. At DC and low frequency

there is virtually no signal coupling between channels. Crosstalk increases with frequency and is dependent on circuit gain, source impedance and signal characteristics.

As source impedance increases, careful circuit layout will help achieve lowest channel crosstalk. Most crosstalk is produced by capacitive coupling of signals from one channel to the input section of the other channel. To minimize coupling, separate the input traces as far as practical from any signals associated with the opposite channel. A grounded guard trace surrounding the inputs helps reduce stray coupling between channels. Run the differential inputs of each channel parallel to each other or directly adjacent on top and bottom side of a circuit board. Stray coupling then tends to produce a common-mode signal which is rejected by the IA's input.

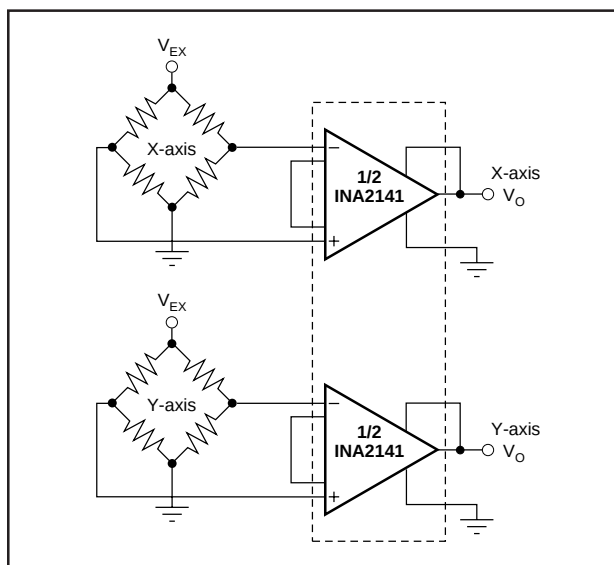


FIGURE 4. Two-Axis Bridge Amplifier.

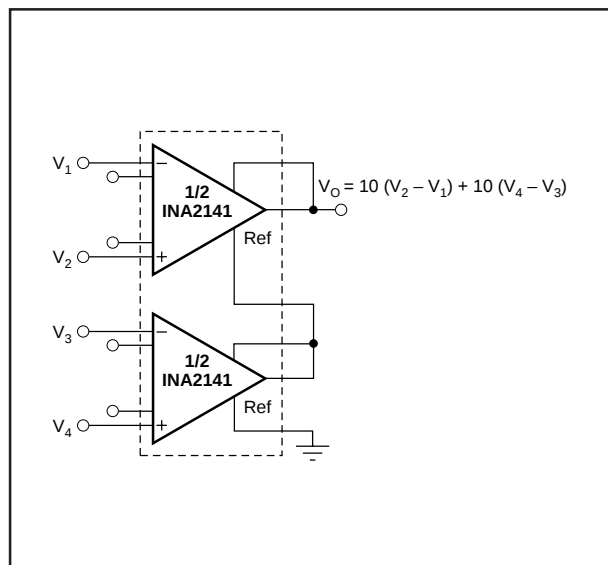


FIGURE 5. Sum of Differences Amplifier.

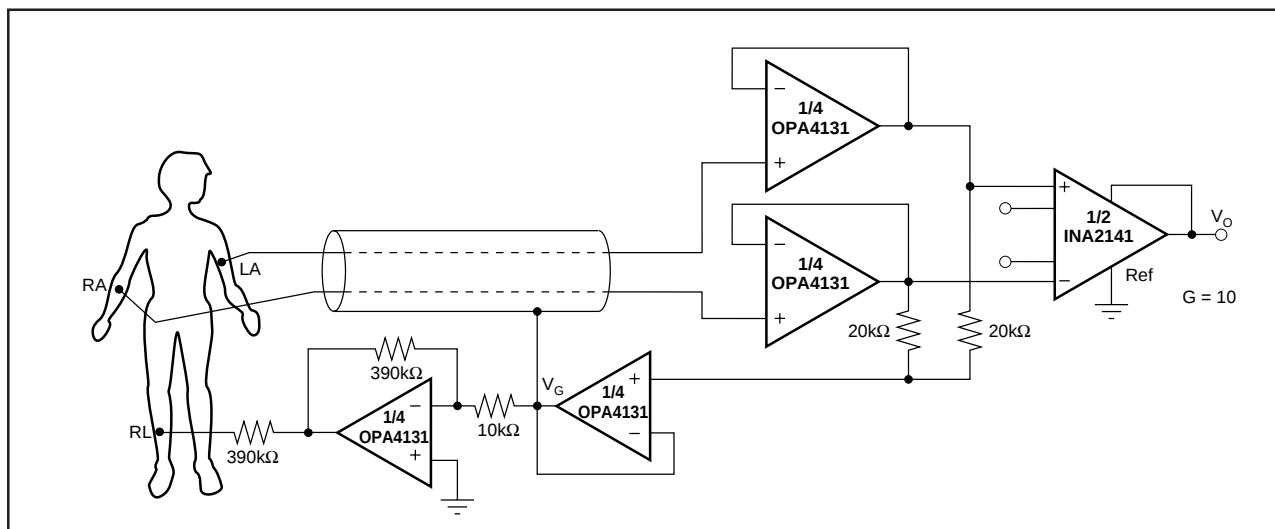


FIGURE 6. ECG Amplifier With Right-Leg Drive.

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