



# 5 V CATV MODEM RF TRANSCEIVER ISG3300EU

## FEATURES

- **TWO WAY DOCSIS COMPLIANT:**  
100-860 MHz Downstream  
5-65 MHz Upstream
- **INTERFACES DIRECTLY WITH QAM DEMOD/MOD ICs**
- **INCORPORATES ALL RF FILTERS INCLUDING Tx ANTI ALIASING**
- **DIGITAL LINEAR RF TRANSMITTER:**  
All Harmonics -51 dBc @ 58 dBmV :  $\geq 50$  dB Combined  
Gain Control w/ Digital Step Attenuator
- **SUPERIOR SNR:**  
37 dB for 64 QAM / 256 QAM (TYP)
- **RUGGED DESIGN/NO MICROPHONICS:**  
SMD Components, "Coiless"

## DESCRIPTION AND APPLICATIONS

The ISG3300EU is a complete RF transceiver designed for use in European cable modem applications. The transceiver integrates a diplex filter, triple conversion receiver and transmit AGC amplifier with antialiasing filters (see Figure 1). The diplex filter provides over 40 dB of isolation between the Tx band and the Rx band. The receiver channel selects and converts QAM signals in the Rx band down to the desired IF sampling frequency. It also provides the necessary gain control input power to the QAM demodulator. The RF transmitter section combined with the digital modulator provides a digitally controlled gain range of  $\geq 50$  dB, while maintaining excellent linearity performance.

## ELECTRICAL CHARACTERISTICS (V<sub>CC</sub> = 5 V, V<sub>CC2</sub> = 24 V, T<sub>A</sub> = 25°C)

PART NUMBER			ISG3300EU		
SYMBOLS	PARAMETERS	UNITS	MIN	TYP	MAX
<b>RF Performance (Rx)</b>					
fOP	Operating Frequency Range	MHz	100		860
	Input Signal Level	dBmV	-20		15
	Gain Range	dB	25		75
VAGC	Automatic Gain Control Voltage RF		0		3.3
VAGC	Automatic Gain Control Voltage IF		0		3.3
NFMAX	Noise Figure (Max Gain)	dB		8	10
	Phase Noise at 10 kHz Offset	dBc/Hz		-83	-80
	LO Radiation at RF Input	dBm		-40	
	Resolution	KHz		62.5	
	Lock Time (end to end channel)	msec.		18	
	Input Impedance (Nominal)	ohms		75	
RLIN	Input Return Loss	dB	6		
	Channel Bandwidth EU	MHz		8	
	Output Frequency <sup>1</sup>	MHz		8.125	
	Passband Ripple	dB		1	2
	Image Rejection	dB	50		
	Inband Group Delay	ns			100
	CSO <sup>2</sup>	dBc			45
	CTB <sup>2</sup>	dBc			45
	Frequency Offset	KHz	-35		+35

Notes:

1. Optional Output 36.125 MHz Available
2. 110 Channels at +15 dBmV/tone

**ELECTRICAL CHARACTERISTICS** ( $V_{CC} = 5\text{ V}$ ,  $V_{CC2} = 24\text{ V}$ ,  $T_A = 25^\circ\text{C}$ )

PART NUMBER			ISG3300EU		
SYMBOLS	PARAMETERS	UNITS	MIN	TYP	MAX
<b>RF Performance (Tx)</b>					
f <sub>OP</sub>	Operating Frequency Range	MHz	5		65
	G <sub>MAX</sub> (V = 0 V)	dB		25	
	G <sub>MID</sub> (V = 1.5 V)	dB		9	
	G <sub>MIN</sub> (V = 3.0 V)	dB		-7	
	2nd Harmonic Level (Single Tone, P <sub>OUT</sub> = +58 dBmV)	dBc		-51	-49
	3rd Harmonic Level (Single Tone, P <sub>OUT</sub> = +58 dBmV)	dBc		-51	-49
R <sub>L</sub> OUT	Output Return Loss	dB		10	6
T <sub>X</sub> ON	On/Off Setting Time	μs		12	
T <sub>X</sub> OFF	On/Off Setting Time	μs		5	
<b>Power Requirements</b>					
	Supply Voltage V1 Rx	V	4.75	5	5.25
	Supply Voltage V2 Rx	V	20	24	31.5
	Supply Voltage V1 Tx ( 5 V)	V	4.75	5	5.25
<b>Supply Current</b>					
I <sub>CC1</sub> (Rx)	Supply Current 1 (Rx)	mA		240	320
I <sub>CC2</sub> (Rx)	Supply Current 2 (Rx)	mA		1.5	2
I <sub>CC1</sub> (Tx)	Supply Current 1 (Tx)	mA		80	110
<b>Physical Interface</b>					
	To the CATV Network			Female F-Connector	
	To the Motherboard			16 Pin Header	
<b>Physical Dimensions</b>					
	L x W x H			3.4 x 2.0 x 0.5"	
<b>Environmental Specs</b>					
T <sub>OP</sub>	Operating Temperature <sup>1</sup>	°C	-5		60
T <sub>STG</sub>	Storage Temperature	°C	-40		75

Note:

1. Temperature performance parameters will vary slightly.

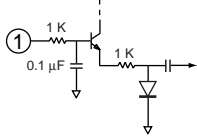
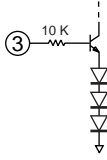
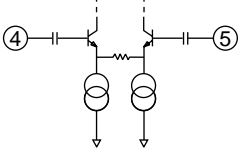
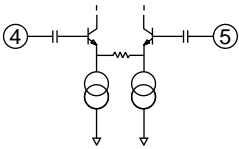
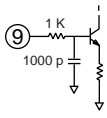
**ABSOLUTE MAXIMUM RATINGS**(T<sub>C</sub> = 25 °C unless otherwise noted)

SYMBOLS	PARAMETERS	UNITS	RATINGS
V <sub>IN</sub>	RF Input Voltage	dBmV	60
V <sub>CC1</sub> (Rx)	Supply Voltage 1 (Rx)	V	6
V <sub>CC2</sub> (Rx)	Supply Voltage 2 (Rx)	V	35
V <sub>CC</sub> (Tx)	Supply Voltage (Tx)	V	6
T <sub>OP</sub>	Operating Temperature	°C	-10 to 60
T <sub>STG</sub>	Storage Temperature	°C	-55 to 150
T <sub>SOL</sub>	Soldering Temperature	°C	260
t <sub>SOL</sub>	Soldering Time	sec.	4

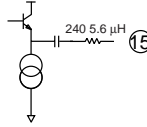
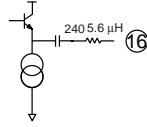
Note:

1. Operation in excess of any one of these parameters may result in permanent damage.

## PIN FUNCTIONS

Pin No.	Pin Name	Description	Equivalent Circuit
1	RFAGC	The RFAGC pin is used to adjust gain in the dual conversion tuner. This pin has a positive gain vs. AGC slope. 20 dB of gain control is available by varying the voltage from 0.5 V to 3.3 V.	
2	VCC (Tx)	The VCC (Tx) pin powers the Tx amplifier. A 5 V bias is required and nominal current is 125 mA.	
3	TXEN	The TXEN pin is used to enable/disable the Tx amplifier. When TXEN is set LOW, the Tx amplifier is disabled. In this state, a standby current of 3 mA is required from VCC (Tx). When TXEN is set HIGH, the Tx amplifier is enabled. In this state, a nominal current of 125 mA is required from VCC (Tx).	
4	TXIN-	TXIN- is the inverting input to the Tx amplifier. The input frequency range spans 5-55 MHz.	
5	TXIN+	TXIN+ is the non-inverting input to the Tx amplifier. The input frequency range spans 5-55 MHz.	
6	TXAGC	The TXAGC pin is used to step control the gain in the Tx amplifier. There are 3 bits of control.	
7	VCC2 (Rx)	The VCC2 (Rx) pin powers the loop filter for the first LO. A bias of 24 V - 30 V is required and maximum current draw is 2 mA.	
8	VCC1 (Rx)	The VCC1 (Rx) pin powers the entire Rx section. A bias of 5 V is required and nominal current draw is 250 mA.	
9	IFAGC	The IFAGC pin is used to adjust gain in the final downconverter stage of the Rx section. The pin has a positive gain vs. AGC slope. 30 dB of gain control is available by varying the voltage from 0.5 to 2.0 V.	
10	GND	Ground.	

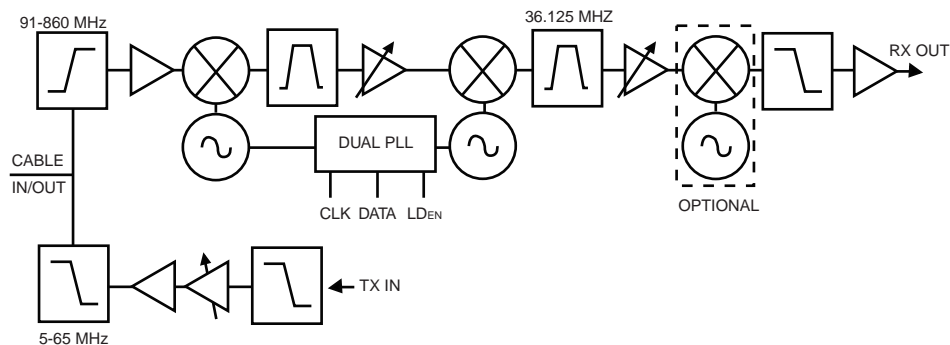
## PIN FUNCTIONS

Pin No.	Pin Name	Description	Equivalent Circuit
11	CLK	Clock pin for the dual PLL. High impedance CMOS input. Data for the various latches is clocked in on the rising edge into a 20-bit shift register.	
12	DATA	Serial data pin for the dual PLL. High impedance CMOS input. MSB entered first. The last two bits are the control bits.	
13	LDEN	Latch enable pin for the dual PLL. High impedance CMOS input. When LE goes HIGH, data stored in the shift registers is loaded into one of the 4 latches determined by the 2 control bits.	
14	GND	Ground.	
15	IFOUT+	Non-inverting final IF output.	
16	IFOUT-	Inverting final IF output.	

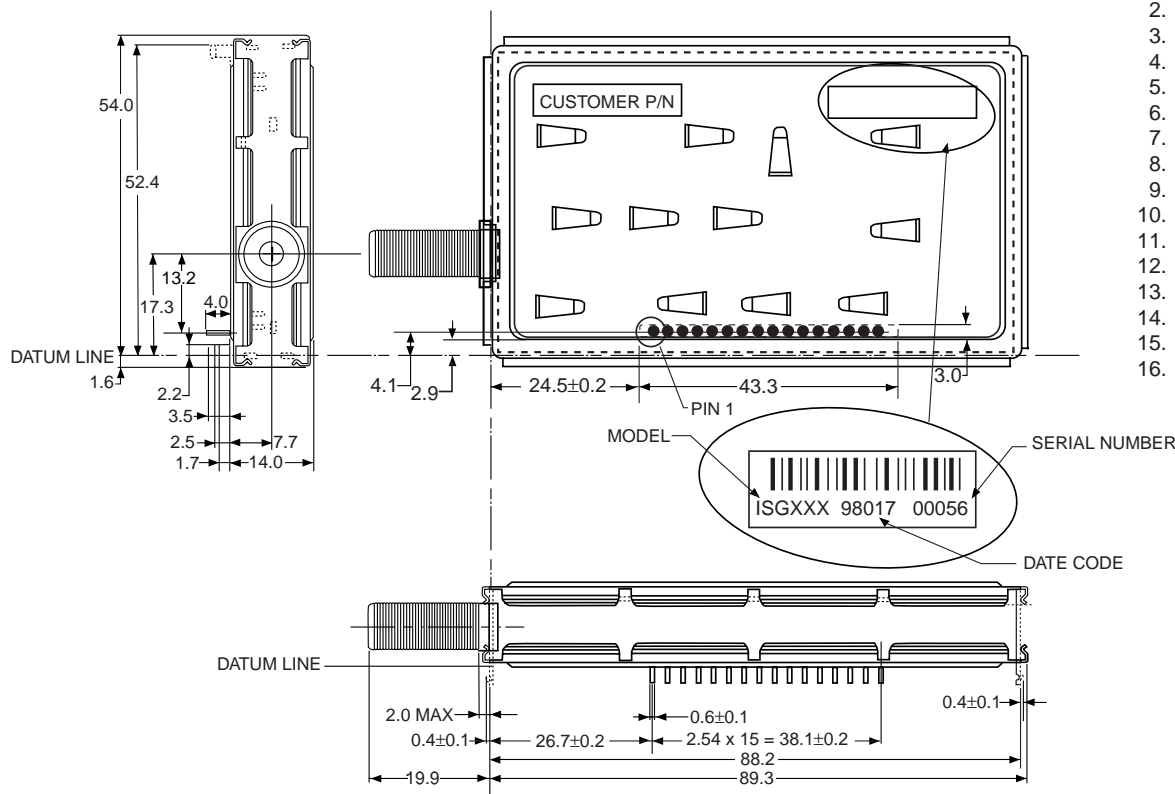
Note:

1. For programming information, refer to National LMX2336 data sheet (<http://www.national.com>)

FIGURE 1



# OUTLINE DIMENSIONS (Units in mm)



**Table 1. Programmable Modes**

C1	C2	R16	R17	R18	R19	R20
0	0	RF2 Phase Detector Polarity	RF2 ICP <sub>O</sub>	RF2 Do TRI-STATE	RF2 LD	RF2 Fo
0	1	RF1 Phase Detector Polarity	RF1 ICP <sub>O</sub>	RF1 Do TRI-STATE	RF1 LD	RF1 Fo

C1	C2	N19	N20
1	0	RF2 Prescaler	Pwdn RF2
1	1	RF1 Prescaler	Pwdn RF1

**Table 2. Mode Select Truth Table**

	Phase Detector Polarity <sup>3</sup>	Do TRI-STATE	ICP <sub>O</sub> <sup>1</sup>	RF1 Prescaler	RF2 Prescaler	Pwdn <sup>2</sup>
0	Negative	Normal Operation	LOW	64/65	64/65	pwdn up
1	Positive	TRI-STATE	HIGH	128/129	128/129	pwdn dn

## Notes:

- The ICP<sub>O</sub> LOW Current State = 1/4 x ICP<sub>O</sub> HIGH Current.
- Activation of the RF2 PLL or RF1 PLL powerdown modes result in the disabling of the respective N counter divider and debiasing of its respective fin inputs (to a high impedance state). The powerdown function is gated by the charge pump to prevent unwanted frequency jumps. Once the powerdown program mode is loaded, the part will go into powerdown mode when the charge pump reaches a TRI-STATE condition. The R counter and Oscillator functionality does not become disabled until both RF2 and RF1 powerdown bits are activated. The OSC<sub>IN</sub> is connected to Vcc through a 100 kΩ resistor and the OSC<sub>OUT</sub> goes HIGH when this condition exists. The MICROWAVE™ control register remains active and capable of loading and latching data during all the powerdown modes.
- Phase Detector Polarity  
Depending upon VCO characteristics, the R16 bits should be set accordingly:  
When VCO characteristics are positive like (1), R16 should be set HIGH, when VCO characteristics are negative like (2), R16 should be set LOW.

## VCO CHARACTERISTICS

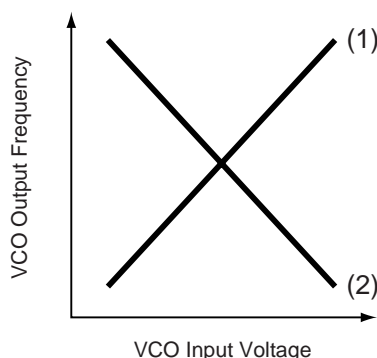


Table 3. The FoLD Output Truth Table

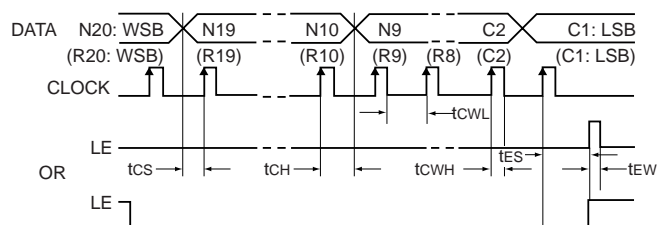
RF1 R (19) (RF1 LD)	RF2 R (19) (RF2 LD)	RF1 R (20) (RF1 Fo)	RF2 R (20) (RF2 Fo)	FoLD Output State
0	0	0	0	Disabled <sup>1</sup>
0	1	0	0	RF2 Lock Detect <sup>2</sup>
1	0	0	0	RF1 Lock Detect <sup>2</sup>
1	1	0	0	RF1/RF2 Lock Detect <sup>2</sup>
X	0	0	1	RF2 Reference Divider Output
X	0	1	0	RF1 Reference Divider Output
X	1	0	1	RF2 Programmable Divider Output
X	1	1	0	RF1 Programmable Divider Output
0	0	1	1	Fastlock <sup>3</sup>
0	1	1	1	For internal use only
1	0	1	1	For internal use only
1	1	1	1	Counter Reset <sup>4</sup>

X - Don't care condition

## Notes:

- When the FoLD output is disabled, it is actively pulled to a low logic state.
- Lock detect output provided to indicate when the VCO frequency is in "lock". When the loop is locked and a lock detect mode is selected, the pin's output is HIGH, with narrow pulses LOW. In the RF1/RF2 lock detect mode a locked condition is indicated when RF2 and RF1 are both locked.
- The Fastlock mode utilized the FoLD output pin to switch a second loop filter damping resistor to ground during fastlock operation. Activation of Fastlock occurs whenever the RF loop's  $I_{cpo}$  magnitude bit #17 is selected HIGH (while the #19 and #20 mode bits are set for Fastlock).
- The Counter Reset mode bits R19 and R20 when activated reset all counters. Upon removal of the Reset bits the N counter resumes counting in "close" alignment with R counter. (The maximum error is one prescaler cycle). If the Reset bits are activated the R counter is also forced to Reset, allowing smooth acquisition upon powering up.

## Serial Data Input Timing



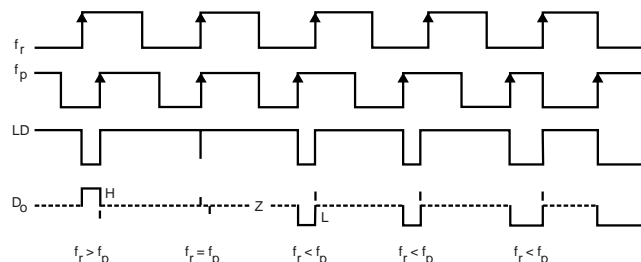
## Notes:

- Parenthesis data indicates programmable reference divider data.
- Data shifted into register on clock rising edge.
- Data is shifted in MSB first.

## Test Conditions:

The Serial Data Input Timing is tested using a symmetrical waveform around  $V_{cc}/2$ . The test waveform has an edge rate of 0.6V/ns with amplitudes of 2.2 V @  $V_{cc} = 2.7$  V and 2.6 V @  $V_{cc} = 5.5$  V.

## Phase Comparator and Internal Charge Pump Characteristics.



## Notes:

- Phase difference detection range:  $-2\pi$  to  $+2\pi$
- The minimum width pump up and pump down current pulses occur at the Do pin when the loop is locked.