5 V CATV MODEM RF TRANSCEIVER



FEATURES

• TWO WAY DOCSIS BASED DESIGN:

91-860 MHz Downstream 5-42 MHz Upstream

- INTERFACES DIRECTLY WITH QAM DEMOD/MOD ICs
- BUILT IN RF TRANSMITTER:

50 dB AGC Driver

• LOW PHASE NOISE:

-83 dBc/Hz @ 10 KHz

• BUSINESS CARD SIZE:

3.4" x 2.0" x 0.5"

• RUGGED DESIGN/NO MICROPHONICS:

All SMD Components "Coiless"

• LOW POWER CONSUMPTION:

Rx: 1.2 Watts, Tx: 0.6 Watts

• EUROPEAN AND JAPANESE VERSIONS AVAILABLE

DESCRIPTION AND APPLICATIONS

The ISG2000 is a complete RF transceiver designed for use in cable modem applications. The transceiver integrates a diplex filter, triple conversion receiver and transmit AGC amplifier (see Figure 1). The diplex filter provides over 40 dB of isolation between the Tx band and the Rx band. The receiver section provides channel selectivity and converts QAM carriers to low IF sampling frequency for digital signal processing. The transmitter section provides 50 dB of gain control while maintaining excellent linearity performance.

ELECTRICAL CHARACTERISTICS (Vcc = 5 V, Vcc2 = 24 V, TA = 25°C)

PART NUMBER			ISG2000			
SYMBOLS	PARAMETERS	UNITS	MIN	TYP	MAX	
RF Performance (Rx)						
fop	Operating Frequency Range	MHz	91		860	
	Input Signal Level	dBmV	-20		15	
	Gain Range	dB	25		75	
Vagc	Automatic Gain Control Voltage RF		0		3.3	
Vagc	Automatic Gain Control Voltage IF		0		3.3	
NFMAX	Noise Figure (Max Gain)	dB		8	10	
	Phase Noise at 10 kHz Offset	dBc/Hz		-83	-80	
	LO Radiation at RF Input	dBm		-40		
	Resolution	KHz		62.5		
	Lock Time (end to end channel)	msec.		18		
	Input Impedance (Nominal)	ohms		75		
RLIN	Input Return Loss	dB	5			
	Channel Bandwidth USA	MHz		6		
	Output Frequency ¹	MHz	5.70	5.75	5.80	
	Passband Ripple	dB		1	2	
	Image Rejection	dB	50			
	Inband Group Delay	ns			100	
	CSO ²	dBc			45	
	CTB ²	dBc			45	
	Frequency Offset	KHz	-50		+50	

Notes

- 1. Optional Output Frequency of 43.75 MHz Available
- 2. 110 Channels at +15 dBmV/tone



ELECTRICAL CHARACTERISTICS (Vcc = 5 V, Vcc2 = 24 V, TA = 25°C)

	PART NUMBER		ISG2000		
SYMBOLS	PARAMETERS	UNITS	MIN	TYP	MAX
RF Perfori	mance (Tx)			1	
fop	Operating Frequency Range	MHz	5		42
	Gain (VAGC = 0 V)	dB		35	
	Gain (VAGC = 3.3 V)	dB		-30	
Vagc	Automatic Gain Control Voltage	V	0		3.3
	2nd Harmonic Level (Single Tone, Pout = +58 dBmV)	dBc	-53	-56	
	3rd Harmonic Level (Single Tone, Pout = +58 dBmV)	dBc	-53	-56	
RLout	Output Return Loss	dB		10	6
Tx on	On/Off Setting Time (See Figure 1)	μS		12	
Tx off	On/Off Setting Time (See Figure 1)	μS		5	
Power Red	uirements				
	Supply Voltage V1 Rx (5 V)	V	4.75	5	5.25
	Supply Voltage V2 Rx (30 V) mA	V	22	24	31.5
	Supply Voltage V1 Tx (5 V)	V	4.75	5	9
Supply Cu	rrent				
Icc1 (Rx)	Supply Current 1 (Rx)	mA		240	320
Icc2 (Rx)	Supply Current 2 (Rx)	mA		1.5	2
Icc1 (Tx)	Supply Current 1 (Tx)	mA		120	160
Physical Ir	nterface				
	To the CATV Network			Female	
				F-Connector	
	To the Motherboard			16 Pin Header	
Physical D	imensions				
	LxWxH			3.4 x 2.0 x 0.5"	
Environme	ental Specs				
Тор	Operating Temperature ¹	°C	-5		60
Tstg	Storage Temperature	°C	-40		75

Note:

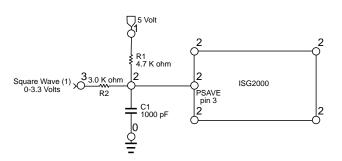
ABSOLUTE MAXIMUM RATINGS

(Tc = 25 °C unless otherwise noted)

SYMBOLS	PARAMETERS	UNITS	RATINGS
VIN	RF Input Voltage	dBmV	60
Vcc1 (Rx)	(Rx) Supply Voltage 1 (Rx)		6
Vcc2 (Rx)	Vcc2 (Rx) Supply Voltage 2 (Rx)		35
Vcc (Tx)	Vcc (Tx) Supply Voltage (Tx)		6
Top Operating Temperature		°C	-10 to 60
TSTG Storage Temperature		°C	-40 to 75
TsoL Soldering Temperature		°C	260
tsoL Soldering Time		sec.	4

Note:

TX ENABLE APPLICATION CIRCUIT (Figure 1)



^{1.} Temperature performance parameters will vary slightly.

^{1.} Operation in excess of any one of these parameters may result in permanent damage.

PIN FUNCTIONS

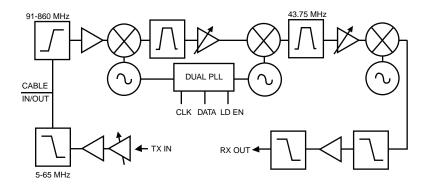
1 114 1 01	NCTIONS		
Pin No.	Pin Name	Description	Equivalent Circuit
1	RFAGC	The RFAGC pin is used to adjust gain in the dual conversion tuner. This pin has a positive gain vs. AGC slope. 20 dB of gain control is available by varying the voltage from 0.5 V to 3.3 V.	1 1 K 1 K 1 K 1 K 1 K 1 K 1 K 1 K 1 K 1
2	Vcc (Tx)	The Vcc (Tx) pin powers the Tx amplifier. A 5 V bias is required and nominal current is 125 mA.	
3	Txen	The TXEN pin is used to enable/disable the Tx amplifier. When TXEN is set LOW, the Tx amplifier is disabled. In this state, a standby current of 3 mA is required from Vcc (Tx). When TXEN is set HIGH, the TX amplifier is enabled. In this state, a nomimal current of 125 mA is required from Vcc (Tx).	3 10 K Y Y Y Y
4	Txin-	TXIN- is the inverting input to the TX amplifier. The input frequency range spans 5-42 MHz.	45
5	Txin+	TxIn+ is the non-inverting input to the Tx amplifier. The input frequency range spans 5-42 MHz.	4-1-5
6	Txagc	The TxAGC pin is used to adjust gain in the Tx amplifier. The pin has a positive gain vs. AGC slope. 50 dB of gain control is available by varying the voltage from 0.5 V to 3.0 V.	(6) 10 K 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1
7	Vcc2 (Rx)	The Vcc2 (Rx) pin powers the loop filter for the first LO. A bias of 24 V - 30 V is required and maximum current draw is 2 mA.	
8	Vcc1 (Rx)	The Vcc1 (Rx) pin powers the entire Rx section. A bias of 5 V is required and nominal current draw is 250 mA.	
9	IFAGC	The IFAGC pin is used to adjust gain in the final downconverter stage of the Rx section. The pin has a positive gain vs. AGC slope. 30 dB of gain control is available by varying the voltage from 0.5 to 2.0 V.	9 1 K 1 1000 p 1
10	GND	Ground.	
	<u> </u>		<u> </u>

PIN FUNCTIONS

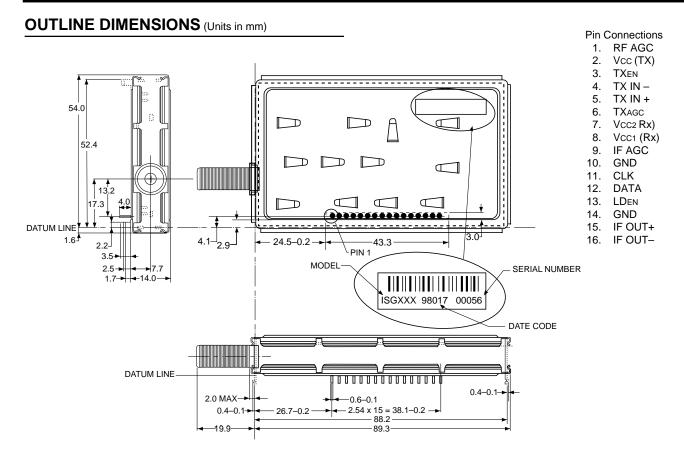
Pin No.	Pin Name	Description	Equivalent Circuit
11	CLK	Clock pin for the dual PLL. High impedance CMOS input. Data for the various latches is clocked in on the rising edge into a 20-bit shift register.	
12	DATA	Serial data pin for the dual PLL. High impedance CMOS input. MSB entered first. The last two bits are the control bits.	
13	LDEN	Latch enable pin for the dual PLL. High impedance CMOS input. When LE goes HIGH, data stored in the shift registers is loaded into one of the 4 latches determined by the 2 control bits.	
14	GND	Ground.	
15	IFout+	Non-Inverting final IF output.	240 5.6 µH 39 p
16	IFout-	Inverting final IF output.	39 p = 68 p = 1

Note:

FIGURE 1



^{1.} For programming information, refer to National LMX2336 data sheet (http://www.national.com)



Note: All tolerances ± 0.15 mm unless otherwise specified.

RECOMMENDED PAD LAYOUT (Units in mm [inches])

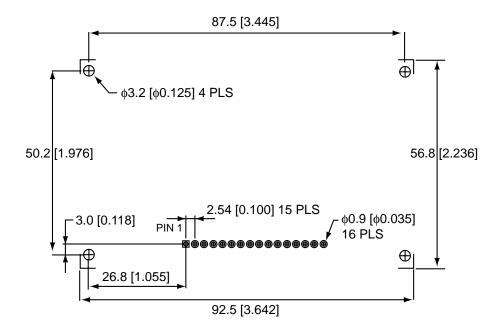


Table 1. Programmable Modes

C1	C2	R16	R17	R18	R19	R20
0	0	RF2 Phase	RF2 Icpo	RF2 Do	RF2 LD	RF2 Fo
		Detector	_	TRI-STATE		
		Polarity				
0	1	RF1 Phase	RF1 IcPo	RF1Do	RF1 LD	RF1 Fo
		Detector Polarity		TRI-STATE		

C1	C2	N19	N20
1	0	RF2 Prescaler	Pwdn RF2
1	1	RF1 Prescaler	Pwdn RF1

Table 2. Mode Select Truth Table

	Phase Detector Polarity ³		ICP _O ¹	RF1 Prescaler	RF2 Prescaler	Pwdn ²
0	Negative	Normal	LOW	64/65	64/65	pwrd up
		Operation				
1	Positive	TRI-STATE	HIGH	128/129	128/129	pwrd dn



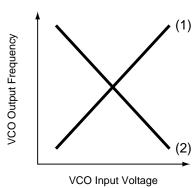


Table 3. The FoLD Output Truth Table

RF1 R (19)	RF2 R (19)	RF1 R (20)	RF2 R (20)	FoLD
(RF1 LD)	(RF2 LD)	(RF1 Fo)	(RF2 Fo)	Output State
0	0	0	0	Disabled ¹
0	1	0	0	RF2 Lock
				Detect ²
1	0	0	0	RF1 Lock
				Detect ²
1	1	0	0	RF1/RF2
				Lock Detect ²
X	0	0	1	RF2 Reference
				Divider Output
X	0	1	0	RF1 Reference
				Divider Output
X	1	0	1	RF2
				Programmable
				Divider Output
X	1	1	0	RF1
				Programmable
				Divider Output
0	0	1	1	Fastlock ³
0	1	1	1 For inte	
				use only
1	0	1	1	For internal
				use only
1	1	1	1	Counter Reset ⁴

X - Don't care condition

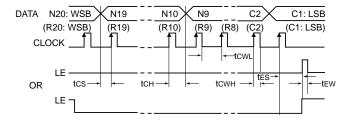
Notes:

- The ICP_O LOW Current State = 1/4 x ICP_O HIGH Current
- 2. Activitation of the RF2 PLL or RF1 PLL powerdown modes result in the disabling of the respective N counter divider and debiasing of its respective fin inputs (to a high impedance state). The powerdown function is gated by the charge pump to prevent unwanted frequency jumps. Once the powerdown program mode is loaded, the part will go into powerdown mode when the charge pump reaches a TRI-STATE condition. The R counter and Oscillator functionality does not become disabled until both RF2 and RF1 powerdown bits are activated. The OSCIN is connected to Vcc through a 100 kΩ resistor and the OSCOUT goes HIGH when this condition exists. The MICROWAVETM control register remains active and capable of loading and latching data during all the powerdown modes.
- Phase Detector Polarity
 Depending upon VCO characteristics, the R16 bits should be set accordingly:
 When VCO characteristics are positive like (1), R16 should be set HIGH, when VCO characteristics are negative like (2), R16 should be set LOW.

Notes:

- When the FoLD output is disabled, it is actively pulled to a low logic state.
- Lock detect output provided to indicate when the VCO frequency is in "lock". When the loop is locked and a lock detect mode is selected, the pin's output is HIGH, with narrow pulses LOW. In the RF1/RF2 lock detect mode a locked condition is indicated when RF2 and RF1 are both locked.
- 3. The Fastlock mode utilized the FoLD output pin to switch a second loop filter damping resistor to ground during fastlock operation. Activation of Fastlock occurs whenever the RF loop's Icpo magnitude bit #17 is selected HIGH (while the #19 and #20 mode bits are set for Fastlock).
- 4. The Counter Reset mode bits R19 and R20 when activated reset all counters. Upon removal of the Reset bits the N counter resumes counting in "close" alignment with R counter. (The maximum error is one prescaler cycle). If the Reset bits are activated the R counter is also forced to Reset, allowing smooth acquisition upon powering up.

Serial Data Input Timing



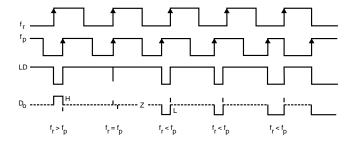
Notes

- Parenthesis data indicates programmable reference divider data.
- 2. Data shifted into register on clock rising edge.
- 3. Data is shifted in MSB first.

Test Conditions:

The Serial Data Input Timing is tested using a symmetrical waveform around Vcc/2. The test waveform has an edge rate of 0.6V/ns with amplitudes of 2.2 V @ Vcc = 2.7 V and 2.6 V @ Vcc = 5.5 V.

Phase Comparator and Internal Charge Pump Characteristics.



Notes:

- 1. Phase difference detection range: -2π to $+2\pi$
- 2. The minimum width pump up and pump down current pulses occur at the Do pin when the loop is locked.

