#### **High Voltage EL Lamp Driver**

#### **Ordering Information**

		Package Options			
Device	Input Voltage	8-Lead SO	Die		
HV823	2.0V to 9.5V	HV823LG	HV823X		

#### **Features**

- □ Processed with HVCMOS<sup>®</sup> technology
- 2.0V to 9.5V operating supply voltage
- DC to AC conversion
- 180V peak-to-peak typical output voltage
- Large output load capability typically 50nF
- Permits the use of high-resistance elastomeric lamp components
- Adjustable output lamp frequency to control lamp color, lamp life, and power consumption
- Adjustable converter frequency to eliminate harmonics and optimize power consumption
- Enable/disable function
- Low current draw under no load condition

#### **Applications**

- Handheld personal computers
- Electronic personal organizers
- GPS units
- Pagers
- Cellular phones
- Portable instrumentation

#### **Absolute Maximum Ratings\***

Supply Voltage, $V_{DD}$	-0.5V to +10V
Output Voltage, V <sub>Cs</sub>	-0.5V to +120V
Operating Temperature Range	-25°C to +85°C
Storage Temperature Range	-65°C to +150°C
Power Dissipation	400mW

#### Note:

\*All voltages are referenced to GND

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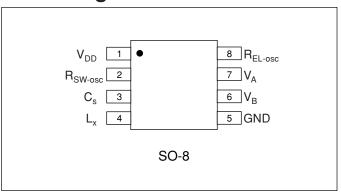
#### **General Description**

The Supertex HV823 is a high-voltage driver designed for driving EL lamps of up to 50nF. EL lamps greater than 50nF can be driven for applications not requiring high brightness. The input supply voltage range is from 2.0 to 9.5V. The device uses a single inductor and a minimum number of passive components. The nominal regulated output voltage that is applied to the EL lamp is  $\pm 90$ V. The chip can be enabled by connecting the resistors on  $R_{SW-osc}$  and  $R_{EL-osc}$  to  $V_{DD}$  and disabled when connected to GND.

The HV823 has two internal oscillators, a switching MOSFET, and a high-voltage EL lamp driver. The frequency for the switching converter MOSFET is set by an external resistor connected between the  $R_{SW\text{-}osc}$  pin and the supply pin  $V_{DD}$ . The EL lamp driver frequency is set by an external resistor connected between  $R_{EL\text{-}osc}$  pin and the  $V_{DD}$  pin. An external inductor is connected between the  $L_x$  and  $V_{DD}$  pins. A  $0.01\mu F$  to  $0.1\mu F$  capacitor is connected between  $V_A$  and  $V_B$ .

The switching MOSFET charges the external inductor and discharges it into the  $C_{\rm s}$  capacitor. The voltage at  $C_{\rm s}$  will start to increase. Once the voltage at  $C_{\rm s}$  reaches a nominal value of 90V, the switching MOSFET is turned OFF to conserve power. The outputs  $V_A$  and  $V_B$  are configured as an H-bridge and are switched in opposite states to achieve 180V peak-to-peak across the EL lamp.

#### **Pin Configuration**



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## **Electrical Characteristics**

**DC Characteristics** ( $V_{IN} = 3.0V$ ,  $R_{SW} = 750K\Omega$ ,  $R_{EL} = 2.0M\Omega$ ,  $T_A = 25^{\circ}C$  unless otherwise specified)

Symbol	Parameter	Min	Тур	Max	Units	Conditions
R <sub>DS(on)</sub>	On-resistance of switching transistor		2	6	Ω	I = 100mA
V <sub>CS</sub>	Output voltage V <sub>CS</sub> Regulation	80	90	100	V	V <sub>IN</sub> = 2.0 to 9.5V
V <sub>A</sub> - V <sub>B</sub>	Output peak to peak voltage	160	180	200	V	$V_{IN} = 2.0V \text{ to } 9.5V$
I <sub>DDQ</sub>	Quiescent V <sub>DD</sub> supply current, disabled		30	100	nA	R <sub>SW-osc</sub> = Low
I <sub>DD</sub>	Input current going into the V <sub>DD</sub> pin		150	200	μΑ	V <sub>IN</sub> = 3.0V. See Figure 1.
				300	μΑ	$V_{IN} = 5.0V$ . See Figure 2.
				500	μΑ	$V_{IN} = 9.0V$ . See Figure 3.
I <sub>IN</sub>	Input current including inductor current		25	33	mA	V <sub>IN</sub> = 3.0V. See Figure 1.
V <sub>CS</sub>	Output voltage on V <sub>CS</sub>	60	70	85	V	$V_{IN} = 3.0V$ . See Figure 1.
f <sub>EL</sub>	V <sub>A-B</sub> output drive frequency	330	380	450	Hz	$V_{IN} = 3.0V$ . See Figure 1.
f <sub>SW</sub>	Switching transistor frequency	50	60	70	KHz	V <sub>IN</sub> = 3.0V. See Figure 1.
D	Switching transistor duty cycle		88		%	

## **Recommended Operating Conditions**

Symbol	Parameter	Min	Тур	Max	Units	Conditions
$V_{DD}$	Supply voltage	2.0		9.5	V	
T <sub>A</sub>	Operating temperature	-25		+85	°C	

## Enable/Disable Table (See Figure 4)

R <sub>SW</sub> resistor	HV823
V <sub>DD</sub>	Enable
0V	Disable

## **Block Diagram**

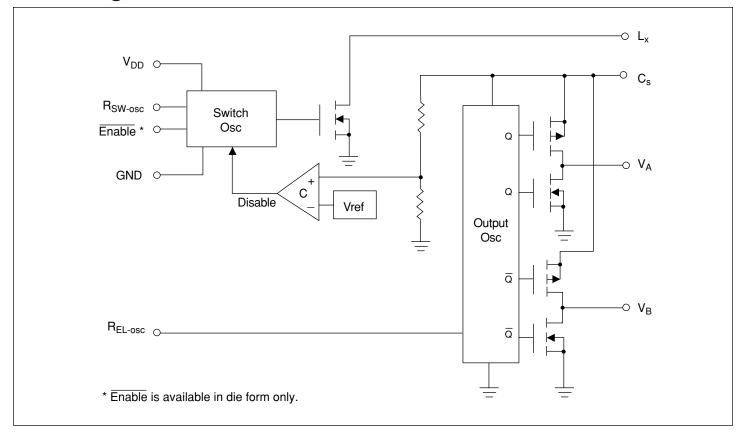
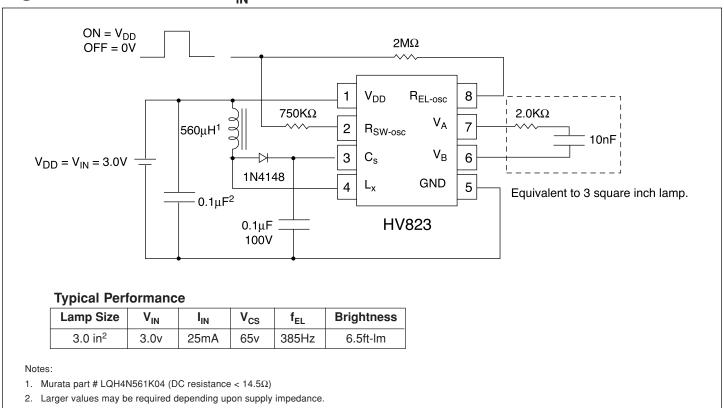
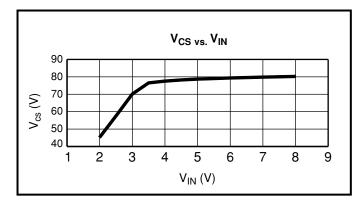


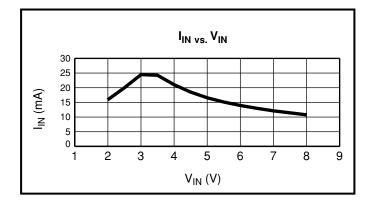
Figure 1: Test Circuit,  $V_{IN} = 3.0V$  (Low input current with moderate output brightness).

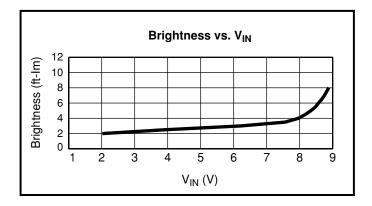


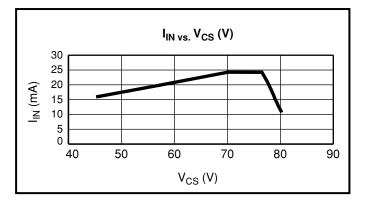
For additional information, see Application Notes AN-H33 and AN-H34.

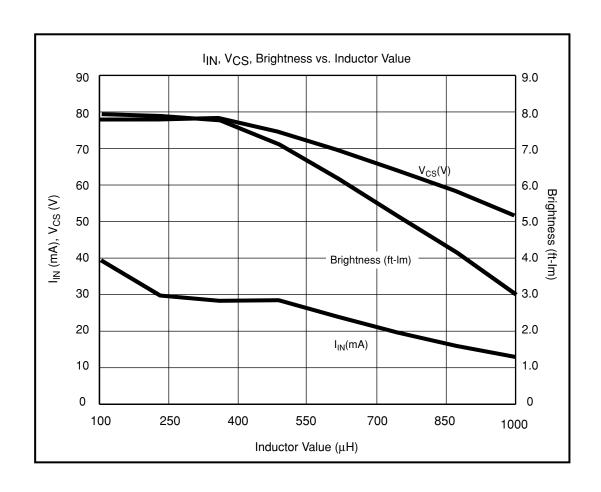
## Typical Performance Curves for Figure 1 using 3in<sup>2</sup> EL Lamp.



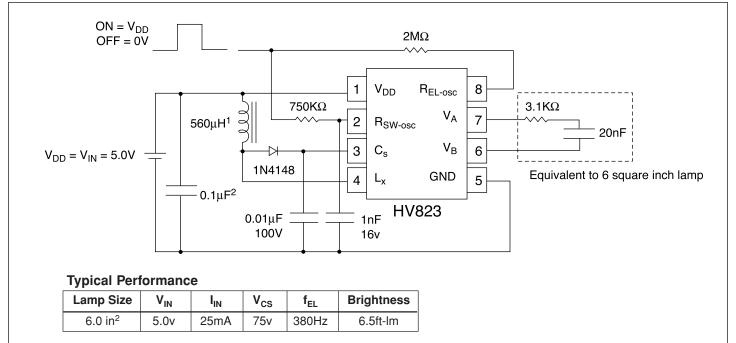








## Figure 2: Typical 5.0V Application

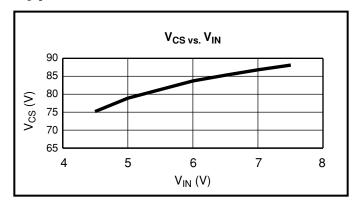


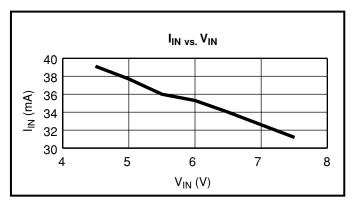
#### Notes:

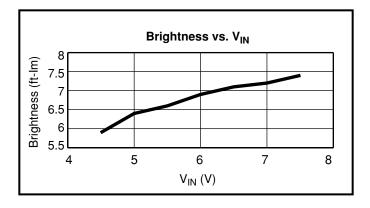
- 1. Murata part # LQH4N561K04 (DC resistance <  $14.5\Omega$ )
- 2. Larger values may be required depending upon supply impedance.

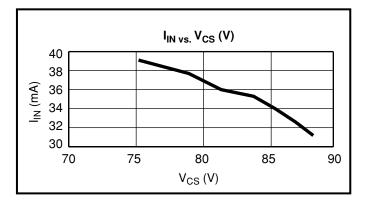
For additional information, see Application Notes AN-H33 and AN-H34.

## **Typical Performance Curves for Figure 2**

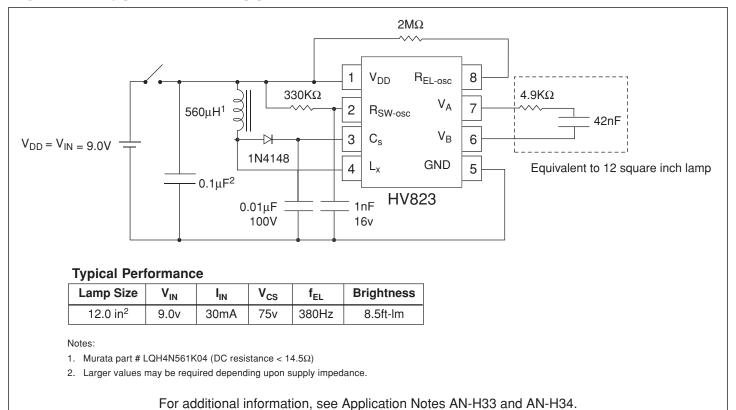




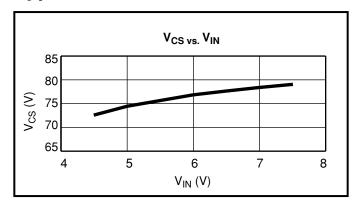


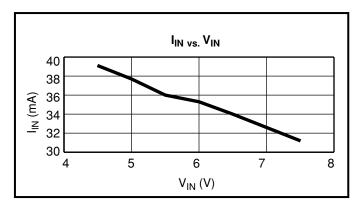


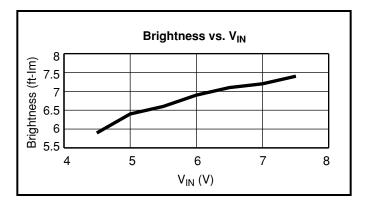
## Figure 3: Typical 9.0V Application\*

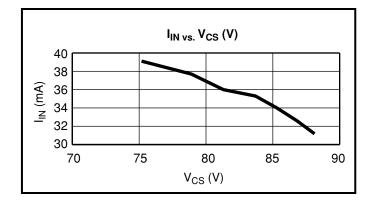


## **Typical Performance Curves for Figure 3**









**External Component Description** 

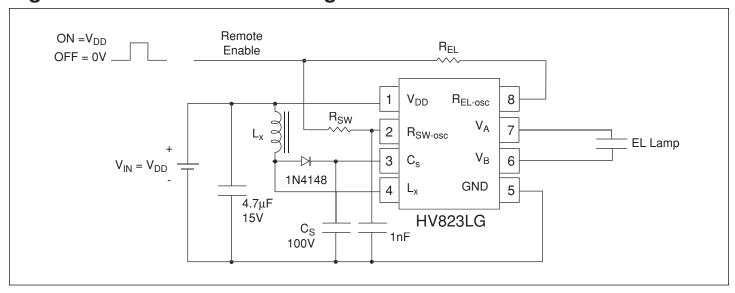
External Component	Selection Guide Line
Diode	Fast reverse recovery diode, 1N4148 or equivalent.
Cs Capacitor	$0.01\mu F$ to $0.1\mu F$ , $100V$ capacitor to GND is used to store the energy transferred from the inductor.
R <sub>EL-osc</sub>	The EL lamp frequency is controlled via an external $R_{EL}$ resistor connected between $R_{EL\text{-osc}}$ and $V_{DD}$ of the device. The lamp frequency increases as $R_{EL}$ decreases. As the EL lamp frequency increases, the amount of current drawn from the battery will increase and the output voltage $V_{CS}$ will decrease. The color of the EL lamp is dependent upon its frequency.
	A $2M\Omega$ resistor would provide lamp frequency of 330 to 450Hz. Decreasing the $R_{EL\text{-osc}}$ by a factor of 2 will increase the lamp frequency by a factor of 2.
R <sub>SW-osc</sub>	The switching frequency of the converter is controlled via an external resistor, $R_{SW}$ between $R_{SW-osc}$ and $V_{DD}$ of the device. The switching frequency increases as $R_{SW}$ decreases. With a given inductor, as the switching frequency increases, the amount of current drawn from the battery will decrease and the output voltage, $V_{CS}$ , will also decrease.
C <sub>SW</sub> Capacitor	A 1nF capacitor is recommended on $R_{SW-osc}$ to GND when a $0.01\mu F$ $C_S$ capacitor is used. This capacitor is used to shunt any switching noise that may couple into the $R_{SW-osc}$ pin. The $C_{SW}$ capacitor may also be needed when driving large EL lamp due to increase in switching noise.
Lx Inductor	The inductor $L_x$ is used to boost the low input voltage by inductive flyback. When the internal switch is on, the inductor is being charged. When the internal switch is off, the charge stored in the inductor will be transferred to the high voltage capacitor $C_S$ . The energy stored in the capacitor is connected to the internal H-bridge and therefore to the EL lamp. In general, smaller value inductors, which can handle more current, are more suitable to drive larger size lamps. As the inductor value decreases, the switching frequency of the inductor (controlled by $R_{SW}$ ) should be increased to avoid saturation.
	$560\mu H$ Murata inductors with $14.5\Omega$ series DC resistance is typically recommended. For inductors with the same inductance value but with lower series DC resistance, lower $R_{SW}$ value is needed to prevent high current draw and inductor saturation.
Lamp	As the EL lamp size increases, more current will be drawn from the battery to maintain high voltage across the EL lamp. The input power, $(V_{IN} \times I_{IN})$ , will also increase. If the input power is greater than the power dissipation of the package (400mW), an external resistor in series with one side of the lamp is recommended to help reduce the package power dissipation.

#### **Enable/Disable Configuration**

The HV823 can be easily enabled and disabled via a logic control signal on the  $R_{\text{SW}}$  and  $R_{\text{EL}}$  resistors as shown in Figure 4 below. The control signal can be from a microprocessor.  $R_{\text{SW}}$  and  $R_{\text{EL}}$  are typically very high values. Therefore, only 10's of microam-

peres will be drawn from the logic signal when it is at a logic high (enable) state. When the microprocessor signal is high the device is enabled and when the signal is low, it is disabled.

## Figure 4: Enable/Disable Configuration



# Split Supply Configuration Using a Single Cell (1.5V) Battery

The HV823 can also be used for handheld devices operating from a single cell 1.5V battery where a regulated voltage is available. This is shown in Figure 5. The regulated voltage can be used to run the internal logic of the HV823. The amount of current necessary to run the internal logic is typically  $100\mu A$  at a  $V_{DD}$  of 3.0V. Therefore, the regulated voltage could easily provide the current without being loaded down. The HV823 used in this configuration can also be enabled/disabled via logic control signal on the  $R_{SW}$  and  $R_{EL}$  resistors as shown in Figure 4.

# Split Supply Configuration for Battery Voltages of Higher than 9.5V

Figure 5 can also be used with high battery voltages such as 12V as long as the input voltage,  $V_{DD}$ , to the HV823 device is within its specifications of 2.0V to 9.5V.

#### Figure 5: Split Supply Configuration

