

Data Sheet May 1999 File Number 2814.4

# **Binary Correlator**

The Intersil HSP45256 is a high-speed, 256 tap binary correlator. It can be configured to perform one-dimensional or two-dimensional correlations of selectable data precision and length. Multiple HSP45256's can be cascaded for increased correlation length. Unused taps can be masked out for reduced correlation length.

The correlation array consists of eight 32-tap stages. These may be cascaded internally to compare 1, 2, 4 or 8-bit input data with a 1-bit reference. Depending on the number of bits in the input data, the length of the correlation can be up to 256, 128, 64, or 32 taps. The HSP45256 can also be configured as two separate correlators with window sizes from 4 by 32 to 1 by 128 each. The mask register can be used to prevent any subset of the 256 bits from contributing to the correlation score.

The output of the correlation array (correlation score) feeds the weight and sum logic, which gives added flexibility to the data format. In addition, an offset register is provided so that a preprogrammed value can be added to the correlation score. This result is then passed through a user programmable delay stage to the cascade summer. The delay stage simplifies the cascading of multiple correlators by compensating for the latency of previous correlators.

The Binary Correlator is configured by writing a set of control registers via a standard microprocessor interface. To simplify operation, both the control and reference registers are double buffered. This allows the user to load new mask and reference data while the current correlation is in progress.

### Features

- Reconfigurable 256 Stage Binary Correlator
- 1-Bit Reference x 1, 2, 4, or 8-Bit Data
- Separate Control and Reference Interfaces
- 25.6, 33MHz Versions
- Configurable for 1-D and 2-D Operation
- Double Buffered Mask and Reference
- Programmable Output Delay
- Cascadable
- Standard Microprocessor Interface

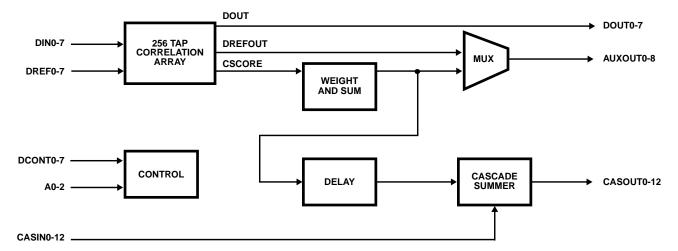
### **Applications**

- · Radar/Sonar
- Spread Spectrum Communications
- · Pattern/Character Recognition
  - Error Correction Coding

### Ordering Information

PART NUMBER	TEMP. RANGE (°C)	PACKAGE	PKG. NO.
HSP45256JC-25	0 to 70	84 Ld PLCC	N84.1.15
HSP45256JC-33	0 to 70	84 Ld PLCC	N84.1.15
HSP45256GC-25	0 to 70	85 Ld PGA	G85.A
HSP45256GC-33	0 to 70	85 Ld PGA	G85.A
HSP45256JI-25	-40 to 85	84 Ld PLCC	N84.1.15
HSP45256JI-33	-40 to 85	84 Ld PLCC	N84.1.15

### **Block Diagram**

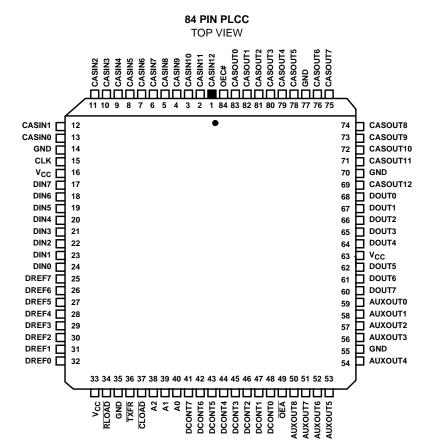


# **Pinouts**

# **85 PIN PGA**BOTTOM VIEW

L	O DREF0	GND	O TXFR	O A2	O DCONT7	O DCONT1 I	OCONT3 I	O DCONT0	O AUX	O AUX	O AUX OUTE
K	O DREF2	Vcc	O RLOAD	CLOAD	O A0	DCONT		OEA	OUT8 O AUX	OUT7 O AUX	OUT5 O AUX
J	O DREF3	O DREF1			O A1	O DCONT	O DCONT		OUT6	OUT4 O GND	OUT3 OUX OUT2
н	O DREF5	O DREF4				5	4		,	O AUXOUT A	$\circ$
G	O DIN0	O DREF7	O DIN1						0	0	O DOUT5
F	O DREF6	O DIN3	O DIN2						0	O DOUT7	O DOUT3
E	O DIN4	O DIN5	O DIN6						O DOUT0	O DOUT1	O DOUT2
D	O DIN7	O <sub>vcc</sub>								O GND C	CASOUT
С	O CLK	CASINO	O		CASIN 8	CASIN	OEC		(	CASOUT C	O CASOUT
В	O GND	O CASIN1	PIN O CASIN3	CASIN6	$\circ$	$\circ$	CASOUT O	CASOUT	$\circ$	$\circ$	$\circ$
Α	CASIN	CAŞIN	CASIN	CASIN	0	0	0	0	0	0	0
	CASIN 2	2	CASIN 3	CASIN 7	CASIN 10	6	7	8	9	10	ASOUT 11
	'	2	3	•		85 PIN P		•	3	10	"
						TOP VIE					
	1	2	3	4	5	TOP VIE	₹W 7	8	9	10	11
Α	1 CASIN 2	2 CASIN 4	3 CASIN 5	4 CASIN 7		TOP VIE	€W	8 CAS OUT3	9 CAS OUT5	10 GND	11 CAS OUT8
A B	CASIN	CASIN	CASIN	CASIN	5 CASIN	TOP VIE	7 CAS	CAS	CAS		CAS
	CASIN 2	CASIN 4 CASIN	CASIN 5 CASIN	CASIN 7 CASIN	5 CASIN 10 CASIN	6 CASIN 11 CAS	7 CAS OUT CAS	CAS OUT3	CAS OUT5	GND CAS	CAS OUT8 CAS
В	CASIN 2	CASIN 4 CASIN 1	CASIN 5 CASIN 3	CASIN 7 CASIN	5 CASIN 10 CASIN 9 CASIN	6 CASIN 11 CAS OUT2 CASIN	7 CAS OUT CAS OUT1	CAS OUT3	CAS OUT5	GND CAS OUT7 CAS	CAS OUT8 CAS OUT10 CAS
В	CASIN 2 GND CLK	CASIN 4 CASIN 1 CASIN 0	CASIN 5 CASIN 3	CASIN 7 CASIN	5 CASIN 10 CASIN 9 CASIN	6 CASIN 11 CAS OUT2 CASIN	7 CAS OUT CAS OUT1	CAS OUT3	CAS OUT5	GND  CAS OUT7  CAS OUT9	CAS OUT8 CAS OUT10 CAS OUT11
B C D	CASIN 2 GND CLK DIN7	CASIN 4 CASIN 1 CASIN 0 VCC	CASIN 5 CASIN 3 INDEX PIN	CASIN 7 CASIN	5 CASIN 10 CASIN 9 CASIN	6 CASIN 11 CAS OUT2 CASIN	7 CAS OUT CAS OUT1	CAS OUT3	CAS OUT5 CAS OUT6	GND  CAS OUT7  CAS OUT9  GND	CAS OUT8 CAS OUT10 CAS OUT11 CAS OUT12
B C D	CASIN 2 GND CLK DIN7 DIN4 DREF	CASIN 4 CASIN 1 CASIN 0 VCC DIN5	CASIN 5 CASIN 3 INDEX PIN	CASIN 7 CASIN	5 CASIN 10 CASIN 9 CASIN	6 CASIN 11 CAS OUT2 CASIN	7 CAS OUT CAS OUT1	CAS OUT3	CAS OUT5 CAS OUT6	GND  CAS OUT7  CAS OUT9  GND  DOUT	CAS OUT10 CAS OUT11 CAS OUT12 DOUT2
B C D E	CASIN 2 GND CLK DIN7 DIN4 DREF 6	CASIN 4  CASIN 1  CASIN 0  VCC  DIN5  DIN3	CASIN 5 CASIN 3 INDEX PIN DIN6 DIN2	CASIN 7 CASIN	5 CASIN 10 CASIN 9 CASIN	6 CASIN 11 CAS OUT2 CASIN	7 CAS OUT CAS OUT1	CAS OUT3	CAS OUT5 CAS OUT6	GND  CAS OUT7  CAS OUT9  GND  DOUT  DOUT  T  DOUT	CAS OUT10 CAS OUT11 CAS OUT11 CAS OUT12 DOUT2 DOUT 3 DOUT
B C D E F	CASIN 2 GND CLK DIN7 DIN4 DREF 6 DIN0 DREF	CASIN 4  CASIN 1  CASIN 0  VCC  DIN5  DIN3  DREF 7  DREF 4  DREF	CASIN 5 CASIN 3 INDEX PIN DIN6 DIN2	CASIN 7 CASIN	5 CASIN 10 CASIN 9 CASIN	CASIN 11 CAS OUT2 CASIN 12	7 CAS OUT CAS OUT1 OEC	CAS OUT3	CAS OUT5 CAS OUT6	GND  CAS OUT7  CAS OUT9  GND  DOUT 7  DOUT 6  AUX	CAS OUT8  CAS OUT10  CAS OUT11  CAS OUT12  DOUT2  DOUT 3  DOUT 5  AUX OUT0  AUX
B C D E F G	CASIN 2 GND CLK DIN7 DIN4 DREF 6 DIN0 DREF 5	CASIN 4  CASIN 1  CASIN 0  VCC  DIN5  DIN3  DREF 7  DREF 4	CASIN 5 CASIN 3 INDEX PIN DIN6 DIN2	CASIN 7 CASIN	CASIN 10 CASIN 9 CASIN 8	CASIN 11 CAS OUT2 CASIN 12	7 CAS OUT CAS OUT1 OEC	CAS OUT3	CAS OUT5 CAS OUT6	GND  CAS OUT7  CAS OUT9  GND  DOUT 7  DOUT 6  AUX OUT1	CAS OUT10 CAS OUT11 CAS OUT12 DOUT2 DOUT 3 DOUT 5 AUX OUT0

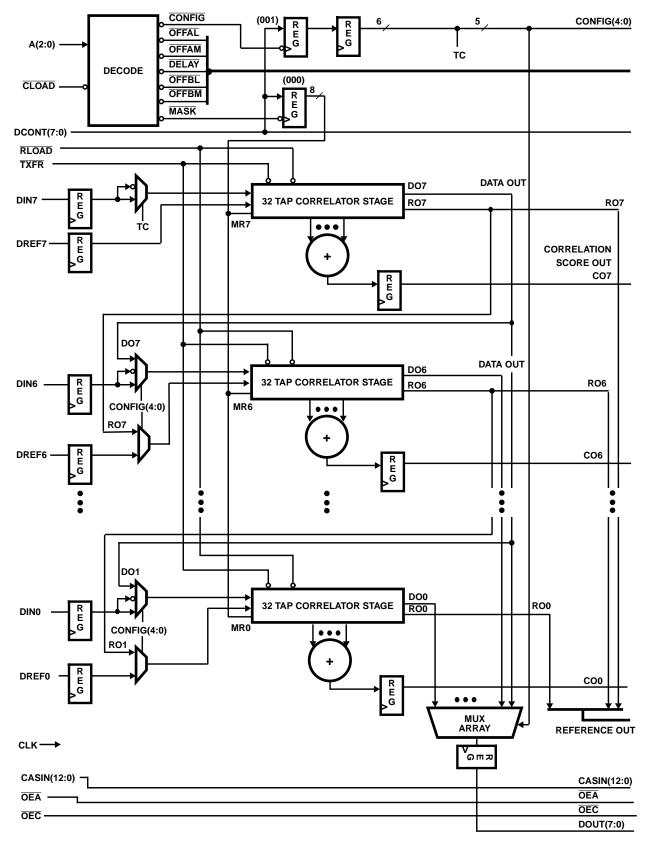
# Pinouts (Continued)



# Pin Descriptions

SYMBOL	PLCC PIN NUMBER	TYPE	DESCRIPTION
V <sub>CC</sub>	16, 33, 63		The +5V power supply pin.
GND	14, 35, 55, 70, 77		Ground.
DIN0-7	17-24	I	The DIN0-7 bus consists of eight single data input pins. The assignment of the active pins is determined by the configuration. Data is loaded synchronous to the rising edge of CLK. DIN0 is the LSB.
DOUT0-7	60-62, 64-68	0	The DOUT0-7 bus is the data output of the correlation array. The format of the output is dependent on the window configuration and bit weighting. DOUT0 is the LSB.
CLK	15	I	System Clock. Positive edge triggered.
CASIN0-12	1-13	I	CASIN0-12 allows multiple correlators to be cascaded by connecting CASOUT0-12 of one correlator to CASIN0-12 of another. The CASIN bus is added internally to the correlation score to form CASOUT. CASIN0 is the LSB.
CASOUT0-12	69, 71-76, 78-83	0	CASOUT0-12 is the output correlation score. This value is the delayed sum of all the 256 taps of one chip and CASIN0-12. When the part is configured to act as two independent correlators, CASOUT0-8 represents the correlation score for the first correlator while the second correlation score is available on the AUXOUT0-8 bus. In this configuration, the cascading feature is no longer an option. CASOUT0 is the LSB.
ŌEC	84	I	$\overline{\text{OEC}}$ is the output enable for CASOUT0-12. When $\overline{\text{OEC}}$ is high, the output is three-stated. Processing is not interrupted by this pin (active low).
TXFR	36	I	TXFR is a synchronous clock enable signal that allows the loading of the reference and mask inputs from the preload register to the correlation array. Data is transferred on the rising edge of CLK while TXFR is low (active low).
DREF0-7	25-32	I	DREF0-7 is an 8-bit wide data reference input. This is the input data bus used to load the reference data. RLOAD going active initiates the loading of the reference registers. This input bus is used to load the reference registers of the correlation array. The manner in which the reference data is loaded is determined by the window configuration. If the window configuration is 1 x 256, the reference bits are loaded one at a time over DREF7. When the HSP45256 is configured as an 8 x 32 array, the data is loaded into all stages in parallel. In this case, DREF7 is the reference data for the first stage and DREF0 is the reference data for the eighth stage. The contents of the reference data registers are not affected by changing the window configuration. DREF0 is the LSB.
RLOAD	34	I	RLOAD enables loading of the reference registers. Data on DREF0-7 is loaded into the pre- load registers on the rising edge of RLOAD. This data is transferred into the correlation array by TXFR (active low).
DCONT0-7	41-48	I	DCONT0-7 is the control data input which is used to load the mask bit for each tap, as well as the configuration registers. The mask data is sequentially loaded into the eight stages in the same manner as the reference data. DCONT0 is the LSB.
CLOAD	37	I	CLOAD enables the loading of the data on DCONT0-7. The destination of this data is controlled by A0-2 (active low).
A0-2	38-40	I	A0-2 is a 3-bit address that determines what function will be performed when $\overline{\text{CLOAD}}$ is active. This address bus is set up with respect to the rising edge of the load signal, $\overline{\text{CLOAD}}$ . A0 is the LSB.
AUXOUT0-8	50-54, 56-59	0	AUXOUT0-8 is a 9-bit bus that provides either the data reference output in the single correlation configuration or the 9-bit correlation score of the second correlator, in the dual correlator configuration. When the user programs the chip to be two separate correlators, the score of the second correlator is output on this bus. When the user has programmed the chip to be one correlator, AUXOUT0-7 represents the reference data out, with the state of AUXOUT8 undefined. AUXOUT0 is the LSB.
ŌĒĀ	49	I	The $\overline{\text{OEA}}$ signal is the output enable for the AUXOUT0-8 output. When $\overline{\text{OEA}}$ is high, the output is disabled. Processing is not interrupted by this pin (active low).

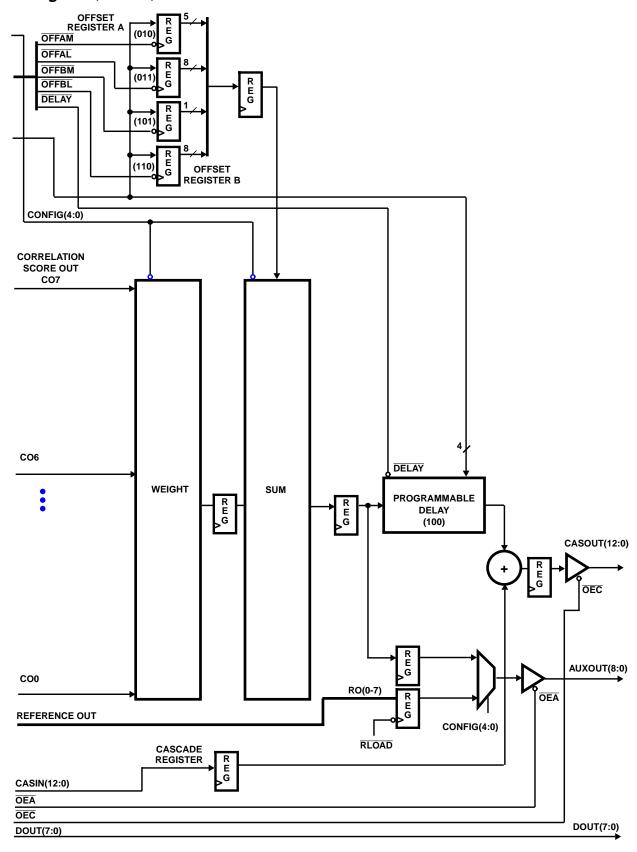
# **Block Diagram**



NOTE: All registers clocked with CLK unless otherwise specified.

**CORRELATOR BLOCK DIAGRAM** 

# **Block Diagram** (Continued)



NOTE: All registers clocked with CLK unless otherwise specified.

### Functional Description

The correlation array consists of eight 32-bit stages. The first stage receives data directly from input pin DIN7. The other seven stages receive input data from either an external data pin, DIN0-6, or from the Shift Register output of the previous stage, as determined by the Configuration Register. When the part is configured as a single correlator the sum of correlation score, Offset Register and cascade input appears on CASOUT0-12. Delayed versions of the data and reference inputs appear on DOUT0-7 and AUXOUT0-7, respectively. The input and output multiplexers of the correlation array are controlled together; for example, in a 1 x 256 correlation, the input data is loaded into DIN7 and the output appears on DOUT7. The configuration of the data bits, the length of the correlation (and in the two-dimensional data, the number of rows), is commonly called the correlation window. A top level Block Diagram of the single correlator configuration is shown in Figure 1. Compare the single correlator configuration data output and correlation output to the top level Block Diagram of the dual correlator configuration shown in Figure 2.

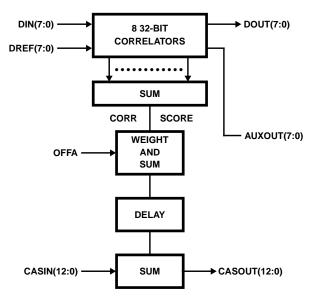


FIGURE 1. SINGLE CORRELATOR CONFIGURATION

### Correlator Array

The core of the HSP45256 is the correlation array, which consists of eight 32-tap stages. A single correlator cell consists of an XNOR gate for the individual bit comparison; i.e., if the data and reference bits are either both high or both low, the output of the correlator cell is high. Figure 3 details the circuitry of a single correlation cell and Figure 4 shows the timing for that single correlation cell. In addition, two latches, one for the reference and one for the control data path are contained in this cell. These latches are loaded from the Preload Registers on the rising edge of CLK when TXFR is low so that the reference and mask values are updated without interrupting data processing.

The mask function is implemented with an AND gate. When a mask bit is a logic low, the corresponding correlator cell output is low.

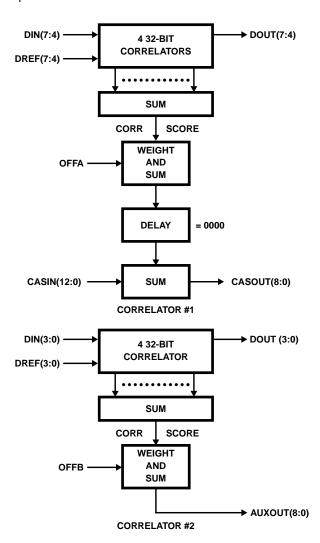


FIGURE 2. DUAL CORRELATOR CONFIGURATION

The function performed by one correlation cell is:

 $(D_{i,n} XNOR R_{i,n}) AND M_{i,n}$  where:

D<sub>i.n</sub> = Bit i of data register n

 $R_{i,n}$  = Bit i of reference register n

Min = Bit i of mask register n

The reference and mask bits are loaded sequentially, N bits at a time, where N depends on the current configuration (see Tables 2 and 9). New reference data is loaded on the rising edge of  $\overline{\text{RLOAD}}$  and new mask data is loaded on the rising

edge of CLOAD. The mask and reference bits are stored internally in Shift Registers, so that the mask and reference information that was loaded most recently will be used to process the newest data. When new information is loaded in, the previous contents of the mask and reference bits are shifted over by one sample, and the oldest information is lost. There are no registers in the multiplexer array (see Block Diagram), so the data on DOUT0-7 corresponds to the data in the last element of the correlation array. When monitoring DOUT0-7, AUXOUT0-8, and REFOUT0-7, only those bits listed in Table 9 are valid.

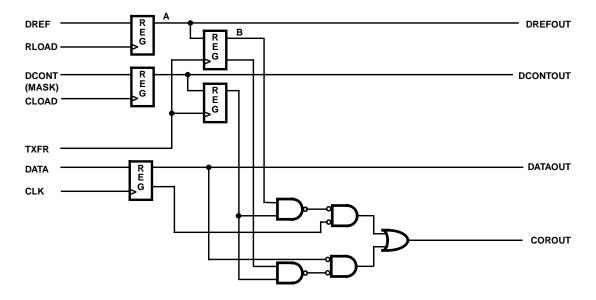


FIGURE 3. CORRELATION CELL BLOCK DIAGRAM

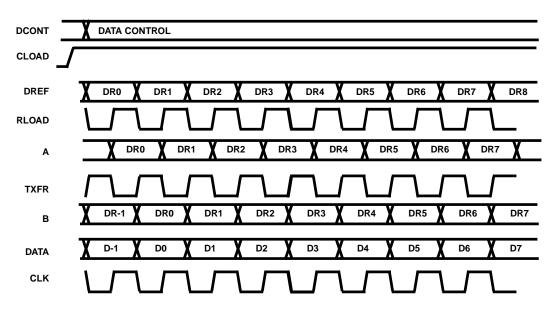


FIGURE 4. CORRELATION CELL TIMING DIAGRAM

### Weight and Sum Logic

The Weight and Sum Logic provides the bit weighting and the final correlator score from the eight stages of the correlation array. For a 1 x 256 1-D configuration, the outputs of each of the stages are given a weight of 1 and then added together. In a 8 x 32 (8-bit data) configuration, the output of each stage will be shifted so that the output data represents an 8-bit word, with stage seven being the MSB.

The 13-bit Offset Register is loaded from the control data bus. Its output is added to the correlation score obtained from the correlator array. This sum then goes to the programmable delay register data input.

When the chip is configured as dual correlators, the user has the capability of loading two different offset values, one for each of the two correlators.

The Programmable Delay Register sets the number of pipeline stages between the output of the weight and sum logic and the input of the Cascade Summer. This delay register is used to align the correlation scores of multiple correlators in HSP45256 cascaded configurations (see Applications Section). The number of delays is programmable from 1 to 16, allowing for up to 16 correlators to be cascaded. When the HSP45256 is configured as dual correlators, the delay must be set to 0000, which specifies a delay of 1.

### Cascade Summer

The Cascade Summer is used for cascading several correlator chips together. The value present on this bus represents the correlation score from the previous HSP45256 that will be summed with the current score to provide the final correlation score. When several correlator chips are cascaded, the CASOUT0-12 of each correlator is connected to the CASIN0-12 of the next correlator in the chain. The CASIN0-12 of the first chip is tied low. The following function represents the correlation score present on CASOUT0-12 of each correlator:

```
CASOUT(n) = (W7 x CO7)(n-Delay) + (W6 x CO6)(n-Delay) + (W5 x CO5)(n-Delay) + (W4 x CO4)(n-Delay) + (W3 x CO3)(n-Delay) + (W2 x CO2)(n-Delay) + (W1 x CO1)(n-Delay) + (W0 x CO0)(n-Delay) + Offset (n-Delay) + CASIN.
```

### where:

CO0-CO7 are the correlation score outputs out of the correlation stages; W0-W7 is the weight given to each stage; n-Delay represents the delay on the weighted and summed correlation score through the Programmable Delay Register; Offset is the value programmed into the Offset register; CASIN is the cascade input.

### **Control Registers**

The 3-bit address value, A0-2, is used to determine which internal register will be loaded with the data on DCONT0-7. The function is initiated when  $\overline{\text{CLOAD}}$  is brought low, and the register is loaded on the rising edge of  $\overline{\text{CLOAD}}$ . Table 1 indicates the function associated with each address. Tables 2 - 8 define the function of the bits in each of the control registers.

**TABLE 1. ADDRESS MAPPING** 

A2	<b>A</b> 1	A0	DESTINATION
0	0	0	Mask Register
0	0	1	Configuration Register
0	1	0	Offset Register A-Most Significant Bits
0	1	1	Offset Register A-Least Significant Bits
1	0	0	Programmable Delay Register
1	0	1	Offset Register B-Most Significant Bits
1	1	0	Offset Register B-Least Significant Bits
1	1	1	Reserved

### TABLE 2. MASK REGISTER

	DESTINATION ADDRESS = 0 (000)						
BIT POSITIONS	FUNCTION	DESCRIPTION					
7-0	Mask Register Bit Enable	MR(7:0): Mask Register. When mask register bit $N = 1$ , the corresponding reference register bit is enabled. Mask register data is loaded from the DCONT(7:0) bus into a holding register on the rising edge of $\overline{\text{CLOAD}}$ and is written to the mask register on the rising edge of $\overline{\text{TXFR}}$ .					

### TABLE 3. CONFIGURATION REGISTER

	DESTINATION ADDRESS = 1 (001)							
BIT POSITION	FUNCTION	DESCRIPTION						
7-6	Reserved	Reserved; Program to zero.						
5	тс	Configures correlator for twos complement input format, where the position of the MSB is depends on the current configuration. TC = 1 is twos complement; TC = 0 is offset binary.						
4	CONFIG(4)	CONFIG4: The state of CONFIG4 configures the HSP45256 as either one or two correlators. When CONFIG4 = 0, the HSP45256 is configured as one correlator with the correlation score available on CASOUT0-12.						
		When CONFIG4 = 1, the HSP45256 is configured as dual correlators with the first correlators score available on CASOUT0-8 and the second score available on AUXOUT0-8. When the chip is configured as dual correlators, the Programmable Delay must be set to 0000 for a delay of 1.						
3-2	CONFIG(3:2):	CONFIG(3:2): Control the number of data bits to be correlated. See Table 9.						
1-0	CONFIG(1:0)	CONFIG(1:0): CONFIG1 and CONFIG0 represent the length of the correlation window as indicated in Table 9.						

### TABLE 4. MS OFFSET REGISTER A

	DESTINATION ADDRESS = 2 (010)						
BIT POSITION							
7-5	Reserved	Reserved. Program to zero.					
4-0	Offset Register A MSB	OFFA(12:8): Most significant bits of Offset Register A. This is the register used in single correlator mode.					

### TABLE 5. LS OFFSET REGISTER A

	DESTINATION ADDRESS = 3 (011)						
BIT POSITION	FUNCTION	DESCRIPTION					
7-0	Offset Register A LSB	OFFA(7:0): Least significant bits of Offset Register A.					

### TABLE 6. PROGRAMMABLE DELAY REGISTER

	DESTINATION ADDRESS = 4 (100)							
BIT POSITION								
7-4	Reserved	Reserved. Program to zero.						
3-0	Programmable Delay	PDELAY(3:0): Controls amount of delay from the weight and sum logic to the cascade summer. The number of delays is 1-16, with PDELAY = 0000 corresponding to a delay of 1 and PDELAY = 1111 corresponding to a delay of 16.						

### TABLE 7. MS OFFSET REGISTER B

	DESTINATION ADDRESS = 5 (101)						
BIT POSITION							
7-1	Reserved	Reserved. Program to zero.					
0	Offset Register B MSB	OFFB8: Most significant bit of Offset Register B. In dual correlator mode, this register is used for the correlator whose output appears on the AUXOUT pins.					

### TABLE 8. LS OFFSET REGISTER B

DESTINATION ADDRESS = 6 (110)						
BIT POSITION	FUNCTION	DESCRIPTION				
7-0	Offset Register B LSB OFFB0-7: Least significant bits of Offset Register B.					

### TABLE 9. CONFIGURATION SETUP

	CONFIGURATION					NO. OF				ACTIVE INPUTS ACTIVE OUTPUTS			OUTPUT WEIGHTING									
4	3	2	1	0	CORRE- LATORS	DATA BITS	ROWS	LENGTH	CORRE- LATOR	DIN	DREF	DOUT	AUXOUT	CASOUT	CO7	CO6	CO5	CO4	СОЗ	CO2	CO1	C00
0	0	0	0	0	1	1	1	256	-	7	7	7	7	12-0	1	1	1	1	1	1	1	1
0	0	0	0	1	1	1	2	128	-	7, 3	7, 3	7, 3	7, 3	12-0	1	1	1	1	1	1	1	1
0	0	0	1	0	1	1	4	64	-	7, 5, 3, 1	7, 5, 3, 1	7, 5, 3, 1	7, 5, 3, 1	12-0	1	1	1	1	1	1	1	1
0	0	0	1	1	1	1	8	32	-	7-0	7-0	7-0	7-0	12-0	1	1	1	1	1	1	1	1
0	0	1	0	1	1	2	1	128	-	7, 3	7	7, 3	7, 3	12-0	2	2	2	2	1	1	1	1
0	0	1	1	0	1	2	2	64	-	7, 5, 3, 1	7, 5	7, 5, 3, 1	7, 5, 3, 1	12-0	2	2	2	2	1	1	1	1
0	0	1	1	1	1	2	4	32	-	7-0	7, 6, 5, 4	7-0	7-0	12-0	2	2	2	2	1	1	1	1
0	1	0	1	0	1	4	1	64	-	7, 5, 3, 1	7	7, 5, 3, 1	7, 5, 3, 1	12-0	8	8	2	2	4	4	1	1
0	1	0	1	1	1	4	2	32	-	7-0	7, 6	7-0	7-0	12-0	8	8	2	2	4	4	1	1
0	1	1	1	1	1	8	1	32	-	7-0	7	7-0	7-0	12-0	128	8	32	4	64	2	16	1
1	0	0	0	1	2	1	1	128	A B	7 3	7 3	7 3	- 8-0	12-0 -	1 -	1 -	1 -	1 -	- 1	- 1	- 1	- 1
1	0	0	1	0	2	1	2	64	A B	7, 5 3, 1	7, 5 3, 1	7, 5 3, 1	- 8-0	12-0	1 -	1 -	1 -	1 -	- 1	- 1	- 1	- 1
1	0	0	1	1	2	1	4	32	A B	7-4 3-0	7-4 3-0	7-4 3-0	- 8-0	12-0	1 -	1 -	1 -	1 -	- 1	- 1	- 1	- 1
1	0	1	1	0	2	2	1	64	A B	7, 5 3, 1	7 3	7, 5 3, 1	- 8-0	12-0	2	2 -	1 -	1 -	- 2	- 2	- 1	- 1
1	0	1	1	1	2	2	2	32	A B	7-4 3-0	7, 6 3, 2	7-4 3-0	- 8-0	12-0	2	2 -	1 -	1 -	- 2	- 2	- 1	- 1
1	1	0	1	1	2	4	1	32	A B	7-4 3-0	7 3	7-4 3-0	- 8-0	12-0	8 -	2 -	4 -	1 -	- 8	- 2	- 4	- 1

During reference register loading, the 8-bits, DREF0-7 are used as reference data inputs. The falling edge of  $\overline{\text{RLOAD}}$  initiates reference data loading; when  $\overline{\text{RLOAD}}$  returns high, the data on DREF0-7 is latched into the selected correlation stages. The active bits on DREF0-7 are determined by the current configuration.

The window configuration is determined by the state of control signals upon programming the Control Register. Table 9 represents the programming information required for each window configuration. In Table 9, note that the data listed for Output Weighting refers to the weights given to each of the Correlation Sum Outputs (CO0-7 in the Block Diagram).

During initialization, the loading configuration for the reference data is set by the user. Table 9 shows the loading options. These load controls specify whether the reference data for a given stage comes from the shift register output of the previous stage or from an external data pin.

# **Applications**

There are 10 single correlator configurations possible with the HSP45256. There are six dual correlator configurations possible with the HSP45256. Table 10 details the configuration (bits x rows x length) and the maximum correlation sums of all combinations.

TABLE 10. CORRELATION SCORE FORMULAS FOR SINGLE CORRELATOR CONFIGURATIONS

FIGURE NUMBER	CONFIGURATION BITS x ROWS x LENGTH	HIGHEST POSSIBLE TOTAL CORRELA- TION SCORE	CORRELATION SCORE
Figure 5	1 x 1 x 256	256	CS=CO7+CO6+CO5+CO4+CO3+CO2+CO1+CO0
Figure 6	1 x 2 x 128	256	CS=CO7+CO6+CO5+CO4+CO3+CO2+CO1+CO0
Figure 7	1 x 4 x 64	256	CS=CO7+CO6+CO5+CO4+CO3+CO2+CO1+CO0
Figure 8	1 x 8 x 32	256	CS=CO7+CO6+CO5+CO4+CO3+CO2+CO1+CO0
Figure 9	2 x 1 x 128	384	CS=2(CO7+CO6+CO5+CO4)+CO3+ CO2+CO1+CO0
Figure 10	2 x 2 x 64	384	CS=2(CO7+CO6+CO5+CO4)+CO3+CO2+CO1+CO0
Figure 11	2 x 4 x 32	384	CS=2(CO7+CO6+CO5+CO4)+CO3+CO2+CO1+CO0
Figure 12	4 x 1 x 64	960	CS=8(CO7+CO6)+4(CO5+CO4)+2(CO3+CO2)+CO1+CO0
Figure 13	4 x 2 x 32	960	CS=8(CO7+CO6)+4(CO5+CO4)+2(CO3+CO2)+CO1+CO0
Figure 14	8 x 1 x 32	8160	CS=128C07+64CO6+32C05+16CO4+8CO3+4CO2+2CO1+CO0
Figure 15	1 x 1 x 128 1 x 1 x 128	128	CS=C07+C06+C05+C04CS=C031C02+C01+C00
Figure 16	1 x 2 x 64 1 x 2 x 64	128	CS=C07+C06+C05+C04CS=C031C02+C01+C00
Figure 17	1 x 4 x 32 1 x 4 x 32	128	CS=C07+C06+C05+CORCS=C031C02+C01+C00
Figure 18	2 x 1 x 64 2 x 1 x 64	192	CS=2(CO7+CO6)+CO5+CO4CS=(CO3+CO2)+CO1+CO0
Figure 19	2 x 2 x 32 2 x 2 x 32	192	CS=2(CO7+CO6)+CO5+CO4CS=(CO3+CO2)+CO1+CO0
Figure 20	4 x 1 x 32 4 x 1 x 32	480	CS=8CO7+4CO6+2CO5+CO4CS= 8CO3+4CO2+2CO1+CO0

# Single Correlator Configurations

### 1-Bit Data, Single Row, 256 Samples Configuration

A 1 x 256 (1-D configuration) correlation requires only 1 HSP45256. To initialize the correlator, all the reference bits, control bits, the delay value of the variable delay, and the window configuration must be specified. Table 11 details these settings for the 1-bit data, 256 Samples Configuration. Figure 5 illustrates the data flow through the correlator.

TABLE 11. REGISTER CONTENTS FOR 1 X 256 CORRELATOR WITH EQUAL WEIGHTING

A0-2	DCONT0-7	NOTES
001	00000000	1 256-tap correlator: 1 x 256 window configuration, reference loaded from DREF7, eight stages weighted equally, DIN 7 and DOUT7 are the data input and output, respectively.
010	000000f00	Offset Register A = 0.
011	00000000	
100	00000000	Programmable Delay = 0.
101	00000000	Offset Register B = 0 (Loading of this reg-
110	00000000	ister optional in this mode).

The loading of the Reference and Mask Registers may be done simultaneously by setting A0-2 = 000, setting the DREF and DCONT inputs to their proper values and pulsing RLOAD and CLOAD low. In this configuration, DREF7 loads the reference data and DCONT7 loads the mask information; both sets of data are loaded serially. It will take 256 load pulses (RLOAD) to load the reference array, and 256 CLOAD pulses to load the mask array. Upon completion of the mask and register loading, TXFR is pulsed low, which transfers the reference and control data from the preload registers to the Reference and Mask Registers, updating the data that will be used in the correlation. Reference and mask data can be loaded more quickly by configuring the correlator as an 8 row by 32 sample array, loading the bits eight at a time, then changing the configuration back to 1 x 256 to perform the correlation.

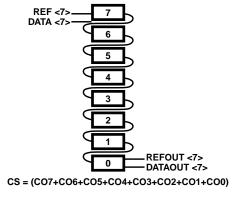


FIGURE 5. 1-BIT, 1 ROW OF 256 TAPS

# Other 1-Bit Configurations

### 1-Bit, Dual Row, 128 Sample Configuration

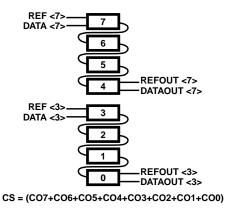
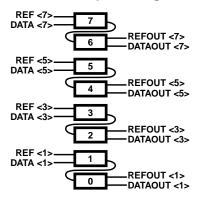


FIGURE 6. 1-BIT, 2 ROWS OF 128 TAPS

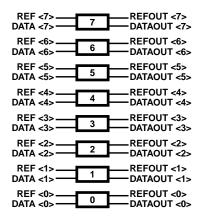
### 1-Bit, Quad Row, 64 Sample Configuration



CS = (CO7+CO6+CO5+CO4+CO3+CO2+CO1+CO0)

FIGURE 7. 1-BIT, 4 ROWS OF 64 TAPS

### 1-Bit, Octal Row, 32 Sample Configuration

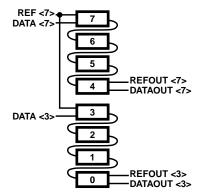


CS = (CO7+CO6+CO5+CO4+CO3+CO2+CO1+CO0)

FIGURE 8. 1-BIT, 8 ROWS OF 32 TAPS

### 2-Bit Configurations

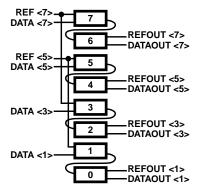
### 2-Bit, Single Row, 128 Sample Configuration



CS = 2(CO7+CO6+CO5+CO4)+(CO3+CO2+CO1+CO0)

FIGURE 9. 2 BITS, 1 ROW OF 128 TAPS

### 2-Bit Data, Dual Row, 64 Samples

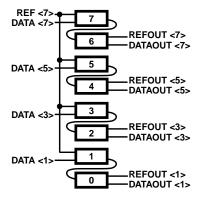


CS = 2(CO7+CO6+CO5+CO4)+(CO3+CO2+CO1+CO0)

FIGURE 10. 2-BITS, 2 ROWS OF 64 TAPS

# 4-Bit Configurations

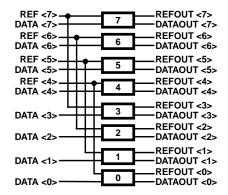
### 4-Bit, Single Row, 64 Sample Configuration



CS = 8(CO7+CO6)+4(CO5+CO4)+2(CO3+CO2)+(CO1+CO0)

FIGURE 12. 4-BITS, 1 ROW OF 64 TAPS

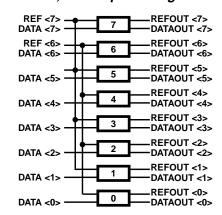
### 2-Bit, Quad Row, 32 Sample Configuration



CS = 2(CO7+CO6+CO5+CO4)+(CO3+CO2+CO1+CO0)

FIGURE 11. 2-BITS, 4 ROWS OF 32 TAPS

### 4-Bit Dual Row, 32 Sample Configurations



CS = 8(CO7+CO6)+4(CO5+CO4)+2(CO3+CO2)+(CO1+CO0)

FIGURE 13. 4 BITS, 2 ROWS OF 32 TAPS

### 8-Bit Configurations

### 8-Bit Data, Single Row, 32 Sample Configurations

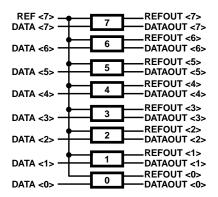
An 8 x 32 correlation also requires only 1 HSP45256. To initialize the correlator, all the reference bits, control bits, the value of the programmable delay, and the window configuration must be specified. Table 12 details these settings.

Again, the loading of the reference and mask registers can be done simultaneously. Due to the programming initialization, DREF0-7 are used to load the reference data 8-bits at a time. It will take 32 load pulses each of  $\overline{\text{RLOAD}}$  and  $\overline{\text{CLOAD}}$  to load both arrays. Upon completion of the mask and register loading,  $\overline{\text{TXFR}}$  is pulsed low, which transfers the reference and control data from the preload registers to the registers that store the active data.

This configuration performs correlation of an 8-bit number with a 1-bit reference. Each byte out of the correlation array gives an 8-bit level of confidence that the data corresponds to the reference. The correlation score is the sum of these confidence levels.

TABLE 12. REGISTER LOADING FOR 8 X 32 CORRELATOR WITH BINARY WEIGHTING

A0-2	DCONT0-7	NOTES
001	00001111	1 256-tap correlator; 8 x 32 window configuration, 8-bit data stream; reference register is loaded from DREF7 for all stages. Correlator score = (128 x CO7) + (64 x CO3) + (32 x CO5) + (16 x CO1) + (8 x CO6) + (4 x CO4) + (2 x CO2) + CO0.
010	00000000	Offset Register A = 000000010000.
011	00010000	
100	00000000	Programmable Delay = 0.
101	00000000	Offset Register B = 0 (Loading optional in
110	00000000	this mode).



CS = 128CO7+64CO6+32CO5+16CO4+8CO3+4CO2+2CO1+CO0

FIGURE 14. 8 BITS, 1 ROW OF 32 TAPS

# **Dual Correlator Configurations**

### 1-Bit, Single Row, 128 Sample Configuration



FIGURE 15. DUAL 1-BIT, 1 ROW OF 128 TAPS

### 1-Bit, Dual Row, 64 Sample Configuration



FIGURE 16. 1-BIT, 2 ROWS OF 64 TAPS

### 1-Bit, Quad Row, 32 Sample Configuration

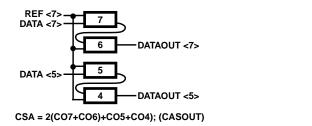


FIGURE 17. 1-BIT, 4 ROWS OF 32 TAPS

### 2-Bit, Dual Row, 64 Sample Configuration

Dual 2 x 64 correlators require only one HSP45256. To initialize the correlator, all the reference bits, control bits, the delay value of the variable delay, and the window configuration must be specified. Table 13 details the settings for the 2-bit Dual Row, 64 Sample Configuration.

In this example, each of the dual correlators compares 2-bit data to a 1-bit reference. It will take 64 load pulses  $(\overline{RLOAD}/\overline{CLOAD})$  to completely load the reference and mask registers in the array. The programmable delay must be set to 0 for the output of the two correlators to be aligned.



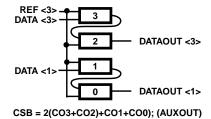


FIGURE 18. 2-BITS, 1 ROW OF 64 TAPS

TABLE 13. REGISTER LOADING FOR DUAL 2 X 64 CORRELATORS WITH EQUAL WEIGHTING

AO-2	DCONT0-7	NOTES
001	00010110	Dual correlators: Each 2 bit data, 64 taps; reference register for correlation A is loaded from DREF3 and DREF5, the reference register for correlator B is loaded from DREF3 and DREF1. Correlator #1 = 2x C07 + 2 x C06 + C05 + C04, correlator #2 = 2 x C03 + 2x C02 + C01 + C00.
010	00000000	Offset Register A = 000000010000.
011	00010000	
100	00000000	Programmable Delay = 0.
101	00000000	Offset Register B = 0.
110	00000000	

### 2-Bit, Dual Row, 32 Sample Configuration



FIGURE 19. 2-BITS, 2 ROWS OF 32 TAPS

### 4-Bit, Single Row, 32 Sample Configuration



FIGURE 20. 4-BITS, 1 ROW OF 32 TAPS

### Cascading Multiple Correlator Devices

Correlators can be cascaded in either a serial or parallel fashion. Longer correlations can be achieved by connecting several correlators together as shown in Figures 21- 23. In Figure 21, each correlator is in a one data bit, one row, 256 tap configuration. The number of bits of significance at the CASOUT output of each correlator builds up from one correlation to the next, that is, the maximum score out of the first correlator is 256, the maximum output of the second correlator is 512, etc. In this configuration, the maximum length of the correlation is 4096. This would be implemented with 16 HSP45256's. The Programmable Delay Register in the first correlator would be set for one delay, the second would be set for two, and so on, with the final HSP45256 being set for a delay of 16.

Correlations of more bits can be calculated by connecting CASOUT of each chip to the CASIN of the following chip

(Figure 21). The data on the CASOUT lines accumulates in a similar manner as in the 1 x 256 mode, except that the maximum output of the first correlator is decimal 960, (hexadecimal 3C0); in the general case, the maximum number of correlators that can be cascaded in this manner is eight, since the maximum output of the last one would be 1E00, which nearly uses up the 13-bit range of the cascade summer. More parts could be cascaded together if some bits are to be masked out or if the user has a prior knowledge of the maximum value of the correlation score. As before, the delay in the first correlator would be set to one, the second correlator would be set for a delay of two, and so on.

Multiple HSP45256's can be cascaded for two dimensional one bit data (Figure 22). The maximum output for each chip is the same as in the 1 x 256 case; the only difference is in the manner in which the correlators are connected. The programmable delay registers would be set as before.

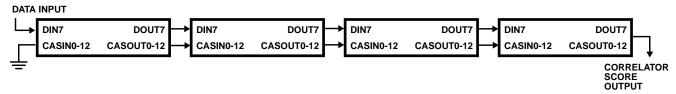


FIGURE 21. 1-BIT, 1024 SAMPLE CONFIGURATION

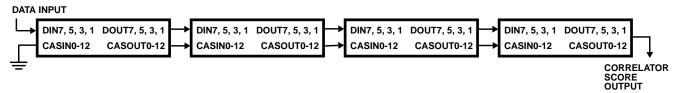


FIGURE 22. 4-BIT, 256 SAMPLE CONFIGURATION

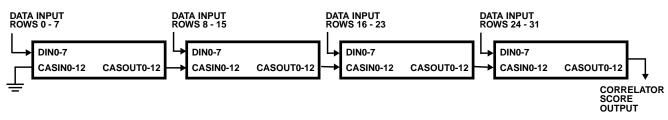


FIGURE 23. 1-BIT, 32 x 32 WINDOW CONFIGURATION

### Reloading Data During Operation

RLOAD and CLOAD are asynchronous signals that are designed to be driven by the memory interface signals of a microprocessor. TXFR is synchronized to CLK so that the mask or reference data is updated on a specific clock cycle. In the normal mode of operation, the user loads the reference and mask memories, then pulses TXFR to use that data. The correlator uses the new mask or reference information immediately. Loading of the reference and mask data remains asynchronous as long as there is at least one cycle of CLK between the rising edge of RLOAD or CLOAD and the TXFR pulse.

If the system timing makes it necessary for TXFR and RLOAD and/or CLOAD to be active during the same clock cycle, then they must be treated as synchronous signals; the timing for this case is shown in Figure 24 and given in the AC Timing Specifications (t<sub>THCL</sub> and t<sub>CLLH</sub>). In this example, data is loaded during clock cycle 1 and transferred on the rising edge of CLK that occurs in clock cycle two. Another set of data is loaded during clock cycle 2, which will be

transferred by a later  $\overline{\mathsf{TXFR}}$  pulse. The sequence of events is as follows:

- 1. In clock cycle 1,  $\overline{\text{TXFR}}$  becomes active at least  $t_{\text{TH}}$  nanoseconds after the rising edge of CLK.
- RLOAD and/or CLOAD pulses low; the timing is not critical as long as its rising edge occurs before the end of clock cycle 1. If this condition is not met, it is undetermined whether the data loaded by this pulse will be transferred by the current TXFR pulse.
- 3. The rising edge of TXFR occurs while CLK is high during clock cycle 2. The margin between the rising edge of TXFR and the falling edge of CLK is defined by t<sub>THCL</sub>.
- RLOAD and/or CLOAD pulses low. The rising edge of RLOAD and CLOAD must occur after the falling edge of CLK. The margin between the two is defined by t<sub>CLLH</sub>.

The time from the rising edge of  $\overline{TXFR}$  to the falling edge of CLK must be greater than  $t_{THCL}$ , and the time from the falling edge of CLK to the rising edge of  $\overline{RLOAD}$  or  $\overline{CLOAD}$  must be greater than  $t_s$ . If this timing is violated, the data being transferred by the  $\overline{TXFR}$  pulse shown may or may not include the data loaded in clock cycle 2.

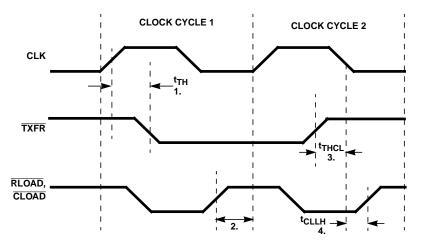


FIGURE 24. LOADING AND TRANSFERRING DATA DURING THE SAME CLOCK CYCLE

### **Absolute Maximum Ratings**

Supply Voltage	+8.0V
Input, Output or I/O Voltage	GND -0.5V to V <sub>CC</sub> +0.5V
FSD Classification	Class 1

### **Operating Conditions**

Voltage Range	+4.75V to +5.25V
Temperature Range	
Commercial	0°C to 70°C
Industrial	40°C to 85°C

### **Thermal Information**

Thermal Resistance (Typical, Note 1)	$\theta_{JA}$ (oC/W)	$\theta_{JC}$ (oC/W)
PLCC Package	34	-
PGA Package	36	10
Maximum Package Power Dissipation		
Commercial PGA		2.9W
Commercial PLCC		2.3W
Industrial PLCC		1.9W
Maximum Storage Temperature Range	65	OC to 150°C
Maximum Junction Temperature		
PLCC		150 <sup>o</sup> C
PGA		175°C
Maximum Lead Temperature (Soldering 1	0s)	300°C
Gate Count	1	13,000 Gates

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

### NOTE:

1.  $\theta_{\mbox{\scriptsize JA}}$  is measured with the component mounted on an evaluation PC board in free air.

### **DC Electrical Specifications**

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	MAX	UNITS
Logical One Input Voltage	V <sub>IH</sub>	V <sub>CC</sub> = 5.25V	2.0	-	V
Logical Zero Input Voltage	V <sub>IL</sub>	V <sub>CC</sub> = 4.75V	-	0.8	V
High Level Clock Input	V <sub>IHC</sub>	V <sub>CC</sub> = 5.25V	3.0	-	V
Low Level Clock Input	V <sub>ILC</sub>	V <sub>CC</sub> = 4.75V	-	0.8	V
Output High Voltage	Voн	I <sub>OH</sub> = 400μA, V <sub>CC</sub> = 4.75V	2.6	-	V
Output Low Voltage	V <sub>OL</sub>	I <sub>OL</sub> = +2.0mA, V <sub>CC</sub> = 4.75V	-	0.4	V
Input Leakage Current	lı	$V_{IN} = V_{CC}$ or GND, $V_{CC} = 5.25V$	-10	10	μΑ
Output Leakage Current	Io	V <sub>OUT</sub> = V <sub>CC</sub> or GND, V <sub>CC</sub> = 5.25V	-10	10	μΑ
Standby Power Supply Current	I <sub>CCSB</sub>	$V_{IN} = V_{CC}$ or GND, $V_{CC} = 5.25V$	-	500	μΑ
Operating Power Supply Current	ICCOP	$f = 25.6MHz$ , $V_{IN} = V_{CC}$ or GND, $V_{CC} = 5.25V$ , Note 2, 4	-	179	mA

# **Capacitance** $T_A = 25^{\circ}C$ , Note 3

PARAMETER	SYMBOL	MIN	MAX	UNITS	TEST CONDITIONS
Input Capacitance	C <sub>IN</sub>	-	10	pF	Frequency = 1MHz, V <sub>CC</sub> = Open All measurements are referenced to device ground.
Output Capacitance	CO	-	10	pF	

### NOTES:

- 2. Power supply current is proportional to operating frequency. Typical rating for  $I_{CCOP}$  is 7mA/MHz.
- 3. Not tested, but characterized at initial design and at major process/design changes.
- 4. Output load per test load circuit and  $C_L = 40 pF$ .

# AC Electrical Specifications $V_{CC} = 5.0V \pm 5\%$ , $T_A = 0^{\circ}C$ to $70^{\circ}C$ , $T_A = -40^{\circ}C$ to $85^{\circ}C$ , Note 5

			33MHz		25.6MHz		
PARAMETER	SYMBOL	NOTES	MIN	MAX	MIN	MAX	UNITS
CLK Period	t <sub>CP</sub>		30	-	39	-	ns
CLK High	tCH		12	-	15	-	ns
CLK Low	t <sub>CL</sub>		12	-	15	-	ns

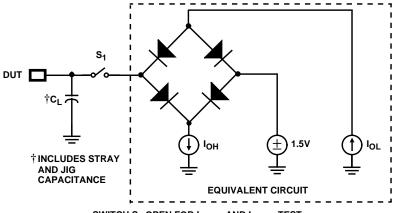
AC Electrical Specifications  $V_{CC} = 5.0V \pm 5\%$ ,  $T_A = 0^{o}C$  to  $70^{o}C$ ,  $T_A = -40^{o}C$  to  $85^{o}C$ , Note 5 (Continued)

			331	ИНz	25.6MHz		
PARAMETER	SYMBOL	NOTES	MIN	MAX	MIN	MAX	UNITS
Set-Up Time DIN to CLK High	t <sub>DS</sub>		12	-	13	-	ns
Hold Time CLK High to DIN	t <sub>DH</sub>		0	-	0	-	ns
TXFR Set-Up Time	t <sub>TS</sub>		12	-	13	-	ns
TXFR Hold Time	t <sub>TH</sub>		0	-	0	-	ns
Output Delay DOUT, AUXOUT, CASOUT	t <sub>DO</sub>		-	15	-	20	ns
CLOAD Cycle Time	t <sub>CLC</sub>		30	-	39	-	ns
CLOAD High	tCLH		12	-	15	-	ns
CLOAD Low	t <sub>CLL</sub>		12	-	15	-	ns
Set-Up Time, A to RLOAD, CLOAD	t <sub>AS</sub>		12	-	13	-	ns
Hold Time, RLOAD, CLOAD to A	t <sub>AH</sub>		0	-	0	-	ns
RLOAD Cycle Time	t <sub>RLC</sub>		30	-	39	-	ns
RLOAD High	t <sub>RLH</sub>		12	-	15	-	ns
RLOAD Low	t <sub>RLL</sub>		12	-	15	-	ns
Set-Up Time, DCONT to CLOAD	t <sub>DCS</sub>		12	-	13	-	ns
Hold Time, CLOAD to DCONT	t <sub>DCH</sub>		0	-	0	-	ns
Set-Up Time, DREF to RLOAD	t <sub>RS</sub>		12	-	13	-	ns
Hold Time, RLOAD to DREF	t <sub>RH</sub>		0	-	0	-	ns
Output Enable Time	toE		-	15	-	15	ns
Output Disable Time	t <sub>OD</sub>	Note 6	-	15	-	15	ns
Output Rise, Fall Time	t <sub>RF</sub>	Note 6	-	6		6	ns
TXFR High to CLK Low	tTHCL	Note 6	3	-	3	-	ns
CLK Low to RLOAD, CLOAD High	tCLLH	Note 6	1	-	1	-	ns

### NOTES:

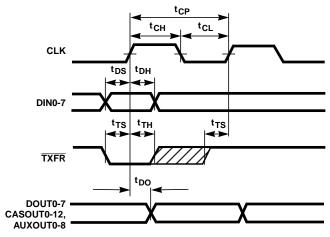
- AC testing is performed as follows: Input levels (CLK Input) 4.0V and 0V; Input levels (all other inputs) 0V and 3.0V; Timing reference levels (CLK) 2.0V; All others 1.5V. Output load per test load circuit with C<sub>L</sub> = 40pF. Output transition is measured at V<sub>OH</sub> > 1.5V and V<sub>OL</sub> < 1.5V.</li>
- 6. Controlled via design or process parameters and not directly tested. Characterized upon initial design and after major process and/or design changes.

### **Test Load Circuit**



SWITCH  $\text{S}_{\text{1}}$  OPEN FOR  $\text{I}_{\text{CCSB}}$  and  $\text{I}_{\text{CCOP}}$  test

# **Timing Waveforms**





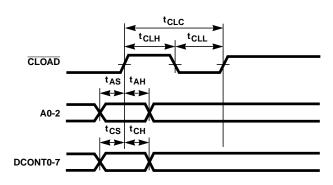


FIGURE 26. CONTROL INPUT TIMING

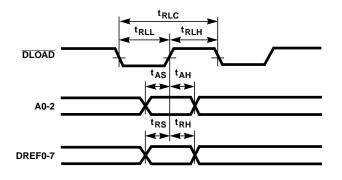
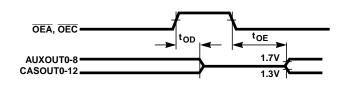


FIGURE 27. REFERENCE INPUT TIMING



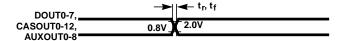


FIGURE 28. OUTPUT TIMING

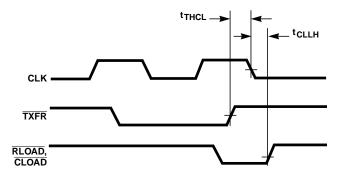


FIGURE 29. TRANSFER, LOAD TIMING WHEN BOTH OCCUR ON A SINGLE CYCLE

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