RADIOTRONIX.inc

EWM-900-FDTC
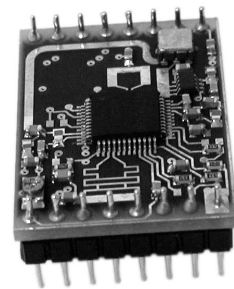
- **Designed for Voice and Data Applications**
- **1000 foot operating range is possible**
- **-109 dBm Sensitivity (12dB SINAD)**
- **19.2K Baud Data Rate**
- **56 Channels**
- **Full-Duplex Operation**

Low-Cost 902-928MHz Full-Duplex Audio/Data Transceiver

The EWM-900-FDTC is ideal for unlicensed voice and data applications. The transceiver module requires no external RF components except for the antenna. It generates virtually no emissions, making FCC and ETSI approvals easy.

The transceiver operates in full-duplex. It can transmit and receive data and/or voice simultaneously. The receiver section employs a direct-conversion, zero IF architecture, eliminating image frequency interference.

The manufacturing-friendly DIP style package and low-cost make the EWM-900-FDTC suitable for high volume applications.

APPLICATIONS

- Automated Meter Reading
- Wireless Headsets
- On-Site Paging
- Asset Tracking
- Wireless Alarm and Security Systems
- Long Range RFID
- Automated Resource Management

KEY FEATURES

- Low-Cost
- 56 channels
- 3V operation
- Small-size: 1.22' x .82"
- No external parts are required
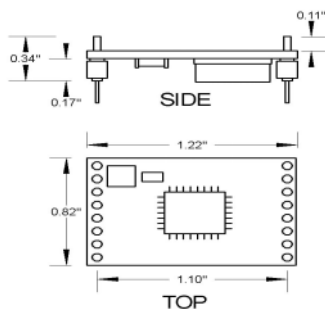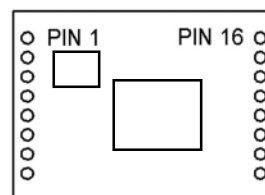- Simple serial programming interface
- RSSI

Figure 1. Mechanical Drawing

Figure 2. Pin-Out Diagram

EWM-900-FDTC

Absolute Maximum Ratings

Rating	Value	Units
Power Supply and All Input Pins	-0.3 to +12	VDC
Storage Temperature	-50 to +100	°C
Soldering Temperature (10 sec)	350	°C

Electrical Characteristics

Characteristic	Symbol	Min	Typ	Max	Units	Notes
Power Supply:						
Operating Voltage	Vcc	2.7	3.0	3.3	Vdc	
Sleep Current	Isleep			5	uA	
RX Current	Irx1		31		mA	LNA off
RX Current	Irx2		35		mA	LNA on
TX Current	Itx		26		mA	
Receiver:						
Audio Sensitivity			-109		dBm	note 1
Data Sensitivity			-100		dBm	note 2
Strong signal SINAD			44		dB	note 3
Input IP3			-1		dBm	
Input P1DB			-18		dBm	
Adjacent channel rejection			60		dB	
TX carrier suppression			45		dB	
RX LO feedthrough			-65		dBm	
Audio output level			600		mVp-p	
Data high voltage		Vcc-.7		Vcc	Vdc	
Data low voltage		0		.7	Vdc	
RSSI voltage range		0.1		2.1	Vdc	
RSSI gain			-32		mv/dB	
RSSI dynamic range			65		dB	
Transmitter:						
Output power			0		dBm	note 4
2nd Harmonic Power			-50		dBm	note 4
3rd Harmonic Power			-60		dBm	note 4
4th Harmonic Power			-70		dBm	note 4
Modulation bandwidth		0.3		10	kHz	
Data input high voltage		Vcc-.7		Vcc	Vdc	
Data input low voltage		0		0.7	Vdc	
Peak deviation			25		kHz pk	

Notes:

1). 12dB SINAD, 1kHz modulation tone, 25kHz p-p frequency deviation
2). 19.2kBit/second, 10^{-5} BER, 25kHz p-p deviation
3). -85dBm input level, 1kHz tone, 25 kHz p-p deviation
4). 50 ohm load

Pin Description

| Pin Number	Name	Description
1	GND	Transceiver ground. Connect to ground plane
2	ANT	50 ohm antenna input
3	GND	Transceiver ground. Connect to ground plane
4	GND	Transceiver ground. Connect to ground plane
5	GND	Transceiver ground. Connect to ground plane
6	GND	Transceiver ground. Connect to ground plane
7	AUDIO	Receive audio output.
8	LNAEN	LNA control input. The LNA is on when this pin is high and off when this pin is low. Do not let this pin float.
9	GND	Transceiver ground. Connect to ground plane
10	CLK	Serial programming interface clock.
11	DAT	Serial programming interface data.
12	LE	Serial programming interface latch enable.
13	RSSI	Receive signal strength indicator
14	RXD	Receive data output
15	TXD/AUDIN	Transmit data/audio input
16	VCC	Power supply. Vcc should be bypassed with a .01uF ceramic capacitor and filtered with a 4.7uF tantalum capacitor. Noise on the power supply will degrade receiver sensitivity.
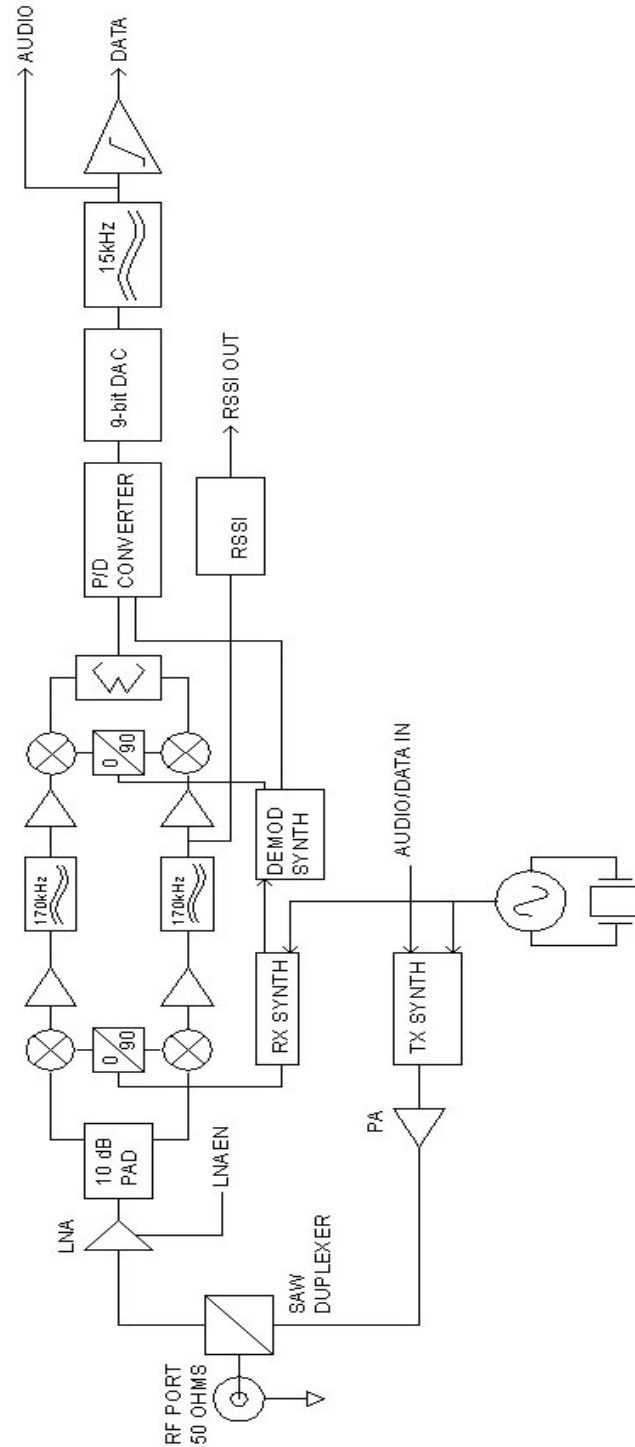
Figure 3: Block Diagram

General Description

The EWM-900-FDTC is a complete, fully-integrated FM/FSK transceiver module capable of full-duplex transmission and reception of voice and data. The transceiver operates on one of 56 channels in the 902-928MHz unlicensed band.

No external RF components (except for the antenna) are required.

The transceiver is configured via a 3-wire serial programming interface comprised of LE, DAT, and CLK. Parameters that can be set using this interface include transmit channel, receive channel, transmit enable, and receive enable.

The unique zero-IF receiver architecture allows for a simple, low-cost solution that does not exhibit the image frequency interference problems of traditional super-het designs.

Theory of Operation

A block diagram of the EWM-900-FDTC transceiver is shown in figure 3.

The antenna input pin is connected directly to the SAW duplexer. The purpose of the duplexer is to separate the receive and transmit frequency bands, effectively combining them while isolating the transmit and receive circuitry. Table 3 shows the frequency plan for -BS (base station) and -HS (handset) versions of the module. It should be noted that -BS modules can only talk to -HS modules and visa-versa because of the complimentary frequency plan.

The receive port of the duplexer is connected to a low-noise-amplifier. The purpose of the amplifier is to compensate for the signal loss through the duplexer and to improve the noise figure of the receiver. The LNA is used when the signal level is low and should be turned off for strong signals. In strong signal conditions, the LNA is turned off to improve the linearity of the receiver.

After the LNA, the incoming carrier is directly converted to baseband (zero-IF) using a pair of quadrature mixers. Special DC offset correction circuitry is employed to ensure proper operation at zero-IF. After the mixers, the receive chain is split into a quadrature pair (I-chain and Q-chain).

Following the quadrature mixers, a pair of variable-gain-amplifiers and low-pass-filters are used to amplify and filter the low-level input signal. Because the receiver uses a zero-IF architecture, these filters can be realized on-chip using only resistors and capacitors, reducing the size and cost of the transceiver.

The gain of the quadrature down-conversion mixers and variable-gain-amplifiers is automatically controlled by an internal AGC circuit. This is done to maintain linearity in the receive chain.

The RSSI circuit derives the RSSI voltage from the Q receive chain.

Demodulation is achieved by up-converting the baseband to 140kHz and digitizing the resultant frequency spectrum. A special P/D circuit is used to demodulate the carrier and generate an analog waveform using a 9-bit D/A converter. In strong signal conditions, this will result in a 44dB SINAD.

The analog output is low-pass filtered to remove IF noise. The audio output is the output of the LPF. The data slicer converts the analog waveform at the audio output to a digital waveform for digital applications. The transceiver includes on-board frequency synthesizers for the transmitter and the receiver. These are programmed through the serial programming interface discussed later in this document.

The transmitter synthesizer includes a modulation input that can be driven with digital or analog information. There is a 150hz high-pass filter on this input. Therefore, DC voltage levels cannot be sent.

The output of the transmitter synthesizer is connected to a power amplifier that boosts the output power to +3dBm typical. After the losses through the duplexer, the output transmit power is 0dBm typical.

Configuration

The transceiver is configured by programming four registers via a three-wire serial interface comprised of CLK (pin 10), DAT (pin 11), and LE (pin 12).
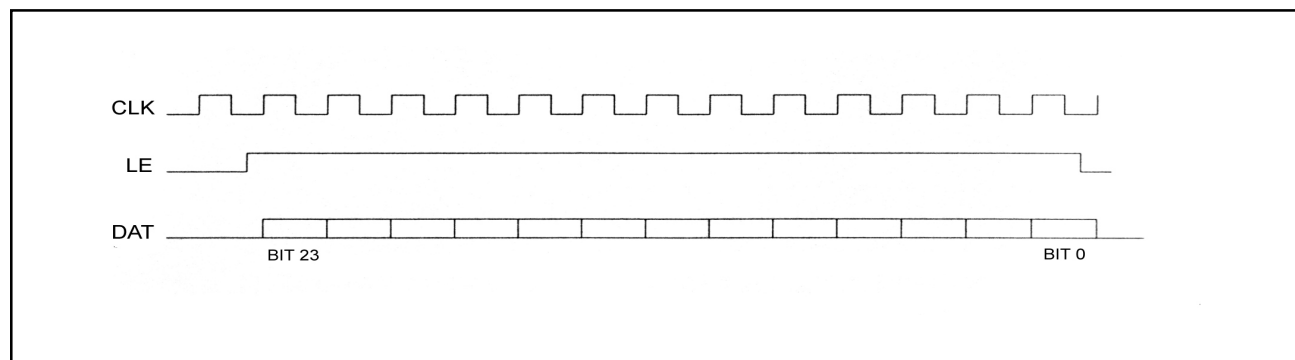

Fig. 4: SERIAL PROGRAMMING WAVEFORMS

When LE is high, bits are shifted in from DAT on each rising edge of CLK. The most significant bit is shifted in first.

Bits 0 and 1 determine the register being programmed.

Bit 0	Bit 1	Register
0	0	Reference frequency control
0	1	RX VCO control
1	0	TX VCO Control
1	1	Mode Control

Table 1: REGISTER ADDRESS BITS

Each register is 22 bits. Therefore, 24 bits must be shifted into the DAT pin to successfully program a register,

When LE transitions high, the incoming bits are stored in a shift register until LE transitions low. On the falling edge of the transition, the contents of the shift register are transfered to the appropriate control register.

Reference Frequency Control Register

The value programmed into this register determines the reference frequency for the transceiver.

For normal operation, the reference frequency is set to 50kHz. The programming word is:

binary: 00000000 01000011 11000000

The two zero's on the end are the register select bits and are set for the reference frequency control register.

The reference frequency should be fixed at 50kHz and should not be changed.

Receive VCO Control Register

The receive VCO control register determines the operating channel of the receiver. The receive frequency is determined by the values of the A, F, and M counters. The trim bits are used to adjust the tuning range of the receive VCO. The trim value is determined at the factory and is provided with each module. It is clearly marked on the top of the module in indelible

Bit position								Byte
7	6	5	4	3	2	1	0	
A Count [0:2]			F Register [0:2]			0	1	0
M Count [0:5]						A Count [3:5]		1
0	0	Trim [0:1]		M Count [6:9]				2

Table 2: REFERENCE FREQUENCY

RADIOTRONIX.inc

Table 3: Receive/Transmit Channel Programming Table

Channel	RX: EWM-900-FDTC-BS TX: EWM-900-FDTC-HS				RX: EWM-900-FDTC-HS TX: EWM-900-FDTC-BS			
	Frequency	M	A	F	Frequency	M	A	F
0	902.250	563	29	0	924.850	578	1	0
1	902.300	563	30	0	924.900	578	2	0
2	902.350	563	31	0	924.950	578	3	0
3	902.400	564	0	0	925.000	578	4	0
4	902.450	564	1	0	925.050	578	5	0
5	902.500	564	2	0	925.100	578	6	0
6	902.550	564	3	0	925.150	578	7	0
7	902.600	564	4	0	925.200	578	8	0
8	902.650	564	5	0	925.250	578	9	0
9	902.700	564	6	0	925.300	578	10	0
10	902.750	564	7	0	925.350	578	11	0
11	902.800	564	8	0	925.400	578	12	0
12	902.850	564	9	0	925.450	578	13	0
13	902.900	564	10	0	925.500	578	14	0
14	902.950	564	11	0	925.550	578	15	0
15	903.000	564	12	0	925.600	578	16	0
16	903.050	564	13	0	925.650	578	17	0
17	903.100	564	14	0	925.700	578	18	0
18	903.150	564	15	0	925.750	578	19	0
19	903.200	564	16	0	925.800	578	20	0
20	903.250	564	17	0	925.850	578	21	0
21	903.300	564	18	0	925.900	578	22	0
22	903.350	564	19	0	925.950	578	23	0
23	903.400	564	20	0	926.000	578	24	0
24	903.450	564	21	0	926.050	578	25	0
25	903.500	564	22	0	926.100	578	26	0
26	903.550	564	23	0	926.150	578	27	0
27	903.600	564	24	0	926.200	578	28	0
28	903.650	564	25	0	926.250	578	29	0
29	903.700	564	26	0	926.300	578	30	0
30	903.750	564	27	0	926.350	578	31	0
31	903.800	564	28	0	926.400	579	0	0
32	903.850	564	29	0	926.450	579	1	0
33	903.900	564	30	0	926.500	579	2	0
34	903.950	564	31	0	926.550	579	3	0
35	904.000	565	0	0	926.600	579	4	0
36	904.050	565	1	0	926.650	579	5	0
37	904.100	565	2	0	926.700	579	6	0
38	904.150	565	3	0	926.750	579	7	0
39	904.200	565	4	0	926.800	579	8	0
40	904.250	565	5	0	926.850	579	9	0
41	904.300	565	6	0	926.900	579	10	0
42	904.350	565	7	0	926.950	579	11	0
43	904.400	565	8	0	927.000	579	12	0
44	904.450	565	9	0	927.050	579	13	0
45	904.500	565	10	0	927.100	579	14	0
46	904.550	565	11	0	927.150	579	15	0
47	904.600	565	12	0	927.200	579	16	0
48	904.650	565	13	0	927.250	579	17	0
49	904.700	565	14	0	927.300	579	18	0
50	904.750	565	15	0	927.350	579	19	0
51	904.800	565	16	0	927.400	579	20	0
52	904.850	565	17	0	927.450	579	21	0
53	904.900	565	18	0	927.500	579	22	0
54	904.950	565	19	0	927.550	579	23	0
55	905.000	565	20	0	927.600	579	24	0

ink.

The actual receive frequency can be calculated using the following formula:

$$F = 50E3 * (32 * M) + A + (F/8)$$

Example: Calculate the A, M, and F values for a receive frequency of 926 MHz.

M=578
A=24
F=0

For this example, assuming a trim value of 2, the programming word will be:

binary: 00101010 00001011 00000001

Table 3 lists the receive frequency and register values for each of the 56 receive channels. The table lists these values for both the -BS and -HS versions of the transceiver.

Transmit VCO Control Register

The transmit VCO control register determines the operating channel of the transmitter. The receive frequency is determined by the values of the M,A, and F counters.

Bit position								Byte
7	6	5	4	3	2	1	0	
A Count [0:2]			F Register [0:2]			1	0	0
M Count [0:5]						A Count [3:5]		1
0	Trim [0:2]			M Count [6:9]				2

Table 4: TX VCO REGISTER

Note that the TX register has three trim bits, where the RX register has only two. The trim bits are used to adjust the tuning range of the transmit VCO. The trim

value is determined at the factory and is provided with each module. It is clearly marked on the top of the module in indelible ink.

The actual transmit frequency can be calculated using same formula as for the receive register. See the receive register description for more information on calculating the M, A, and F values.

The table on the facing page lists the M, A, and F values for each channel. Use this table to quickly determine the values for programming.

Mode Control Register

The mode control register determines the operating modes of the transceiver. With the register, the designer can turn on and off the transmitter and receiver sections of the transceiver.

| Bit position								Byte
7	6	5	4	3	2	1	0	
0	0	1	TXE	RXE	1	1	1	0
0	0	0	0	0	0	0	0	1
0	0	0	0	0	0	0	0	2

Table 5: MODE CONTROL REGISTER

RXE determines the state of the receiver. When the bit is set to 1, the receiver is on. When the bit is set to 0 the bit is off.

TXE determines the state of the transmitter. When the bit set to 1, the transmitter is on. When the bit is set to 0, the bit is off.

The transceiver is placed in power-down mode by turning both the transmitter and receiver off. In this mode, the transceiver will draw <5uA.

The on-board crystal oscillator is automatically enabled when either the transmitter or receiver is enabled.

Figure 5: Evaluation Board Schematic Diagram

RADIOTRONIX.inc.
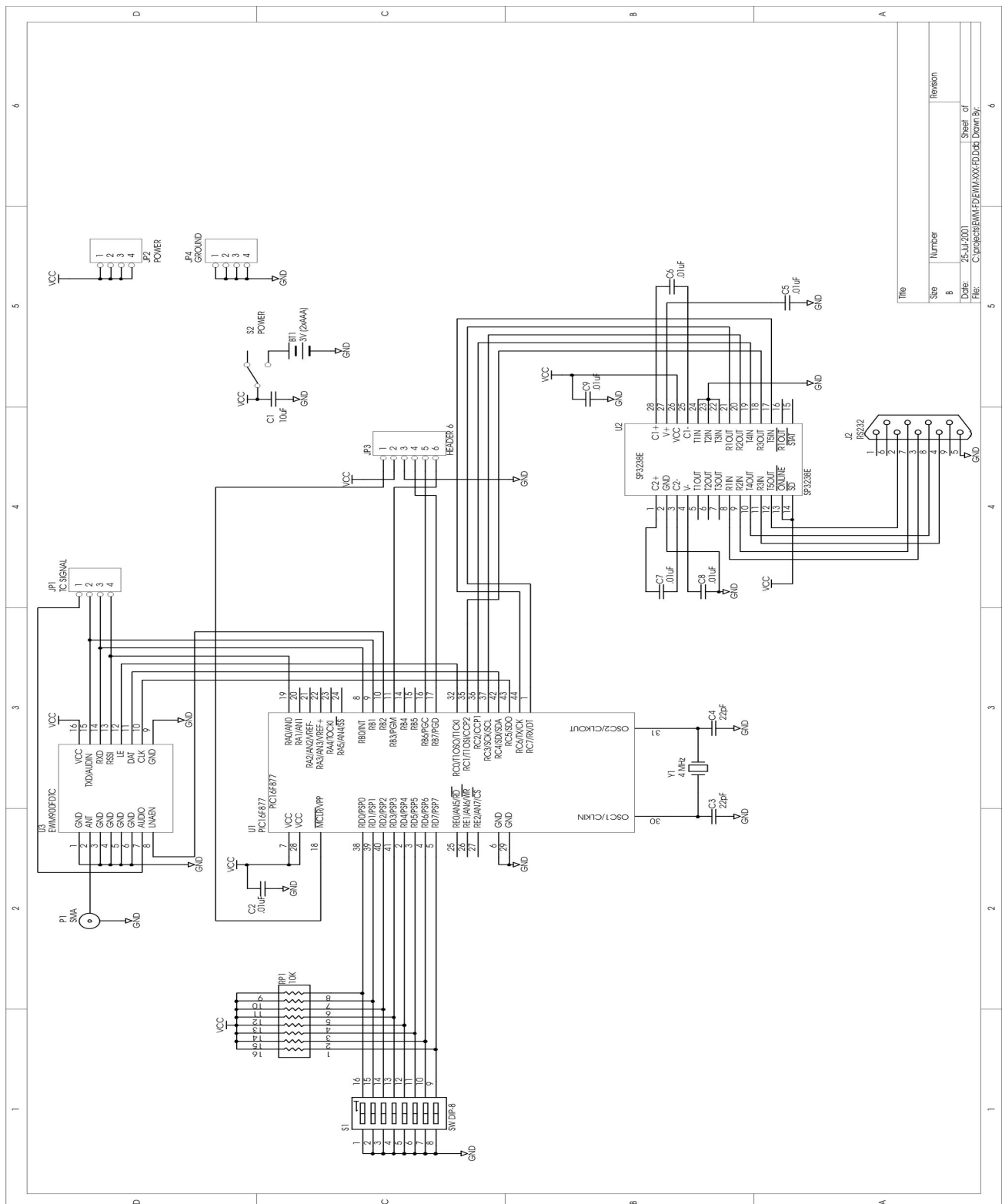
Applications Information

Figure 5 shows the schematic diagram of our evaluation board. All of the transceiver parameters in the specification sheet were determined on this board.

Power Supply

The transceiver is designed to operate from a 3V DC power supply. Power is provided to the transceiver on pin 16. This power supply should be noise-free. Noise on the power supply will degrade the receiver sensitivity, thereby decreasing range.

Ground

It is important to have a good ground system for any wireless design. The ground plane serves as a base reference for the operation of RF circuitry. All filters are referenced to ground. All oscillators are referenced to ground.

If the ground system design is bad, RF circuitry may not function correctly. The results may be reduced performance or spurious emissions.

As a rule of thumb, it is best to use a solid fill ground plane to connect all of the ground pins of the transceiver.

Transmitter Start-Up Time

The start-up time for the transmitter is determined mainly by the loop bandwidth of the PLL loop filter.

When the transmitter is enabled by setting the TXE bit in the mode register to a 1, it can take up to 20mSec for the transmitter to be on-channel and ready to transmit information.

Transmitter Channel-Change Time

The channel change time for the transmitter is the time it takes for the transmitter to be locked onto the new channel after programming the TX channel register.

For a 1MHz jump, the channel change time is 11mSec.

For a 100kHz jump, the channel change time is 7 mSec.

Transmitter Audio/Data Input Requirements

The TXD/AUDIN pin is connected to the transmit VCO. Voltage changes on this pin will frequency modulate the VCO.

The input impedence of this pin will vary from module to module as required to set the proper frequency deviation. Therefore, the pin must be driven by a low impedence source. If the driving circuit is digital, it must have a push-pull output. If the driving circuit is analog, the signal should be buffered by an operational amplifier.

A 3V p-p signal present at this pin will result in a +/- 25kHz frequency deviation.

There is a 150 hz high-pass response on this pin due to the TX PLL loop filter. Therefore, the transmitter is not capable of transmitting DC voltage levels. The bandwidth of the signal used to modulate the transmitter should be between 300 hz and 15kHz.

For data applications, the baud rate should be between 9600 bps and 19.2kbps.

Receiver Start-Up Time

Like the transmitter, the receiver start-up time is determined by the loop bandwidth of the PLL loop filter.

When the receiver is enabled by setting the RXE bit in the mode register to a 1, it can take up to 20mSec for the receiver to be on-channel and ready to receive information.

Receiver Channel-Change Time

The channel change time for the receiver is the time it takes for the receiver to be locked onto the new channel after programming the RX channel register.

For a 1MHz jump, the channel change time is 7mSec.

For a 100kHz jump, the channel change time is 5 mSec.

Receiver Audio Output

The receiver audio output is the analog signal directly from the FM demodulator. It is filtered by a 3rd order LPF with a cut-off frequency of 15kHz.

The audio signal is 600mV typically.

The receiver sensitivity can be increased for audio applications by using an external LPF with a 4kHz cut-off frequency.

Without this filter, the audio sensitivity is -109dBm for a 12dB SINAD. With this filter, the sensitivity should improve by 3-6dB.

Receiver Data Output

The receiver data output is the digital signal from the bit-slicer. It represents the binary data used to modulate the transmitter in data applications.

The data output is rail-to-rail and is compatible with CMOS and TTL signal levels.

RSSI

The RSSI pin indicates the received signal strength of the incoming carrier via a DC voltage.

The RSSI dynamic range is 65dB. However, the LNA extends this dynamic range because the LNA can be turned on and off.

When the LNA is on, the RSSI will indicate from -112 to -62 dBm. The voltage range is .1 to 2.1 V with .1V representing -62dBm.

When the LNA if off,the RSSI will indicate from -80 to -20dBm.

RADIOTRONIX.inc.

ORDERING INFORMATION

EWM-900-FDTC-BS Basestation Version TX: 924-928 MHz RX: 902-905 MHz
EWM-900-FDTC-HS Handset Version TX: 902-905 MHz RX: 924-928 MHz

Contact Information

Radiotronix Inc.
1141 SE Grand Suite 118
Oklahoma City, OK 73129

405-604-3093 phone
405-604-3095 fax

sales@radiotronix.com
support@radiotronix.com

Technical Support

Our preferred method of technical support is through the support forum. Using the support forum, you can search previous postings to see if your question has already been asked and answered. If not, you can post a new question and we will answer it in short order. In this way, you can quickly get answers to commonly asked questions.

We give preference to support questions posted on the forum. To access the forum, go to www.radiotronix.com and click on the Support tab.

Our second preferred method of providing technical support is through e-mail. You can e-mail us your question at support@radiotronix.com.

Our least preferred method of technical support is by phone. Supporting RF products such as ours requires our support engineers to spend time in the lab tracking down problems and answers to customer questions. This process is much more efficient when the support team can respond electronically to a support request.

However, we do recognize that there are some times where nothing less that human contact will do. If that is the case, please call 405-604-3093. Your support request will be placed in a que and handled in the order it was received.

NOTES: