







SLOS378A - SEPTEMBER 2001 - REVISED JULY 2003

FAMILY OF MICROPOWER RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS

FEATURES

- BiMOS Rail-to-Rail Output
- Input Bias Current . . . 1 pA
- High Wide Bandwidth . . . 160 kHz
- High Slew Rate . . . 0.1 V/μs
- Supply Current . . . 7 μA (per channel)
- Input Noise Voltage . . . 89 nV/√Hz
- Supply Voltage Range . . . 2.7 V to 16 V
- Specified Temperature Range
 - -40°C to 125°C . . . Industrial Grade
 - 0°C to 70°C . . . Commercial Grade
- Ultra-Small Packaging
 - 5 Pin SOT-23 (TLV27L1)

APPLICATIONS

- Portable Medical
- Power Monitoring
- Low Power Security Detection Systems
- Smoke Detectors

DESCRIPTION

The TLV27Lx single supply operational amplifiers provide rail-to-rail output capability. The TLV27Lx takes the minimum operating supply voltage down to 2.7 V over the extended industrial temperature range, while adding the rail-to-rail output swing feature. The TLV27Lx also provides 160-kHz bandwidth from only 7 μ A. The maximum recommended supply voltage is 16 V, which allows the devices to be operated from (±8-V supplies down to ±1.35 V) two rechargeable cells.

The rail-to-rail outputs make the TLV27Lx good upgrades for the TLC27Lx family—offering more bandwidth at a lower quiescent current. The TLV27Lx offset voltage is equal to that of the TLC27LxA variant. Their cost effectiveness makes them a good alternative to the TLC/V225x, where offset and noise are not of premium importance.

The TLV27L1/2 are available in the commercial temperature range to enable easy migration from the equivalent TLC27Lx. The TLV27L1 is not available with the power saving/performance boosting programmable pin 8.

The TLV27L1 is available in the small SOT-23 package—something the TLC27(L)1 was not—enabling performance boosting in a smaller package. The TLV27L2 is available in the 3mm x 5mm MSOP, providing PCB area savings over the 8-pin SOIC and 8-pin TSSOP.

SELECTION GUIDE

DEVICE	ν _S [۷]	l _Q /ch [μΑ]	V _{ICR} [V]	V _{IO} [mV]	I _{IB} [pA]	GBW [MHz]	SLEW RATE [V/μs]	V _n , 1 kHz [nV/√Hz]
TLV27Lx	2.7 to 16	11	−0.2 to V _S +1.2	5	60	0.18	0.06	89
TLV238x	2.7 to 16	10	-0.2 to V _S -0.2	4.5	60	0.18	0.06	90
TLC27Lx	4 to 16	17	-0.2 to V _S −1.5	10/5/2	60	0.085	0.03	68
OPAx349	1.8 to 5.5	2	−0.2 to V _S +0.2	10	10	0.070	0.02	300
OPAx347	2.3 to 5.5	34	−0.2 to V _S +0.2	6	10	0.35	0.01	60
TLC225x	2.7 to 16	62.5	0 to V _S -1.5	1.5/0.85	60	0.200	0.02	19

NOTE: All dc specs are maximums while ac specs are typicals.



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PACKAGE/ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE CODE	SYMBOL	SPECIFIED TEMPERATURE RANGE	ORDER NUMBER	TRANSPORT MEDIA	
TI VOZI 4 OD	0010.0		07)/40		TLV27L1CD	Tube	
TLV27L1CD	SOIC-8	D	27V1C	000 1- 7000	TLV27L1CDR	Tape and Reel	
TI.) (071 4000) (207.00	551	\/DIO	0°C to 70°C	TLV27L1CDBVR		
TLV27L1CDBV	SOT-23	DBV	VBIC		TLV27L1CDBVT	Tape and Reel	
TI) (071 41D	0010.0		071/41		TLV27L1ID	Tube	
TLV27L1ID	SOIC-8	D	27V1I	4000 / 40500	TLV27L1IDR	Tape and Reel	
TI VOTI 4100V	207.00	551	1/01/	–40°C to 125°C	TLV27L1IDBVR	T 15 1	
TLV27L1IDBV	SOT-23	DBV	VBII		TLV27L1IDBVT	Tape and Reel	
TI VOTI OOD	0010.0		071/00	000 / 7000	TLV27L2CD	Tube	
TLV27L2CD	SOIC-8	D	27V2C	0°C to 70°C	TLV27L2CDR	Tape and Reel	
TI VOTI 01D	0010.0		071/01	4000 / 40500	TLV27L2ID	Tube	
TLV27L2ID	SOIC-8	D	27V2I	–40°C to 125°C	TLV27L2IDR	Tape and Reel	

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage, V _S	16.5 V
Input voltage, V _I (see Note 1)	V _S
Output current, I _O	100 mÅ
Differential input voltage, V _{ID}	V _S
Continuous total power dissipation	See Dissipation Rating Table
Maximum junction temperature, T _J	150°C
Operating free-air temperature range, TA: C suffix	0°C to 70°C
I suffix	–40°C to 125°C
Storage temperature range, T _{stq}	–65°C to 125°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .	300°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: Relative to GND pin.

DISSIPATION RATING TABLE

PACKAGE	(_o C/M) _θ]C	^θ JA (°C/W)	T _A ≤ 25°C POWER RATING	T _A = 85°C POWER RATING
D (8)	38.3	176	710 mW	370 mW
DBV (5)	55	324.1	385 mW	201 mW
DBV (6)	55	294.3	425 mW	221 mW

recommended operating conditions

		MIN	MAX	UNIT
Complex of the man (A)	Dual supply	±1.35	±8	.,
Supply voltage, (V _S)	Single supply	2.7	16	V
Input common-mode voltage range		-0.2	V _S -1.2	V
Operating free circumparature T	C-suffix	0	70	°C
Operating free-air temperature, T _A	I-suffix	-40	125	C



electrical characteristics at recommended operating conditions, $V_S = 2.7 \text{ V}$, 5 V, and 10 V (unless otherwise noted)

dc performance

	PARAMETER	TEST CONDIT	TIONS	T _A †	MIN	TYP	MAX	UNIT
V	land effect values			25°C		0.5	5	\/
VIO	Input offset voltage	$V_{IC} = V_S/2,$ V_O $R_I = 100 \text{ k}\Omega,$ R_S	= $V_S/2$, = 50Ω	Full range			7	mV
αVIO	Offset voltage drift	11 - 100 132, 115	- 00 12	25°C		1.1		μV/°C
OMBB	Occurred to activation and	$V_{IC} = 0 \text{ V to } V_{S}-1.2 \text{ V},$ $R_{S} = 50 \Omega$		25°C	71	86		-ID
CMRR	Common-mode rejection ratio	$R_S = 50 \Omega$	$R_S = 50 \Omega$		70			dB
			$V_S = 2.7 V$,	25°C	80	100		
	Large-signal differential voltage	VO(PP)=VS/2,	5 V	Full range	77			40
AVD	amplification	$VO(PP)=VS/2,$ $R_L = 100 \text{ k}\Omega$	V- 15.V	25°C	77	82		dB
			$V_S = \pm 5 V$	Full range	74			

[†] Full range is –40°C to 125°C for I suffix.

input characteristics

	PARAMETER	TEST	CONDITIONS	TA	MIN	TYP	MAX	UNIT
				≤25°C		1	60	
liO	Input offset current			≤70°C			100	pА
		$V_{IC} = V_{S}/2,$	$V_0 = V_S/2$,	≤125°C			1000	
		$R_L = 100 \text{ k}\Omega$,	$R_S = 50 \Omega$	≤25°C		1	60	
l _{IB}	Input bias current			≤70°C			200	рΑ
				≤125°C			1000	
r _{i(d)}	Differential input resistance			25°C		1000		GΩ
C _{IC}	Common-mode input capacitance	f = 1 kHz		25°C		8		pF

power supply

	PARAMETER	TEST CONDITIONS	T _A †	MIN	TYP	MAX	UNIT
	Octobrond Comment (complete complete co	V- V /0	25°C		7	11	•
IQ	Quiescent current (per channel)	$V_O = V_S/2$	Full range			16	μΑ
DCDD	Down comply rejection retire (AV /AV)	V _S = 2.7 V to 16 V, No load,	25°C	74	82		J.
PSRR	Power supply rejection ratio (ΔV _S /ΔV _{IO})	V _{IC} = V _S /2 V	Full range	70			dB

[†] Full range is –40°C to 125°C for I suffix.



electrical characteristics at recommended operating conditions, V $_{\mbox{S}}$ = 2.7 V, 5 V, and ± 5 V (unless otherwise noted) (continued)

output characteristics

	PARAMETER	TEST CONDI	TIONS	T _A †	MIN	TYP	MAX	UNIT
			V 0.7.V	25°C	200	160		
		V10 = V0/2	V _S = 2.7 V	Full range	220			
			., -,,	25°C	120	85		
		I _{OL} = 100 μA	V _S = 5 V	Full range	200			
	Output valta as assis a frage vail		\/- \IF\/	25°C	120	50		\/
VO	Output voltage swing from rail		$V_S = \pm 5 V$	Full range	150			V
				25°C	800	420		
		$V_{IC} = V_S/2$,	$V_S = 5 V$	Full range	900			
		$V_{IC} = V_{S}/2,$ $I_{OL} = 500 \mu A$	V- 15.V	25°C	400	200		
			$V_S = \pm 5 V$	Full range	500			
IO	Output current	$V_O = 0.5 \text{ V from rail}$	V _S = 2.7 V	25°C		400		μΑ

[†] Full range is –40°C to 125°C for I suffix.

dynamic performance

	PARAMETER	TEST CONDITIONS	TA	MIN TYP MAX	UNIT
GBP	Gain bandwidth product	$R_L = 100 \text{ k}\Omega$, $C_L = 10 \text{ pF}$, $f = 1 \text{ kHz}$	25°C	160	kHz
			25°C	0.06	
SR	Slew rate at unity gain	$V_{O(pp)} = 1 \text{ V}, R_L = 100 \text{ k}\Omega,$ $C_L = 50 \text{ pF}$	–40°C	0.05	V/µs
		ОС = 30 рг	125°C	0.8	
φМ	Phase margin	$R_L = 100 \text{ k}\Omega$, $C_L = 50 \text{ pF}$	25°C	62	٥
	0.400	$V(STEP)_{DD} = 1 V$, $A_{V} = -1$, Rise	0500	62	
t _S	Settling time (0.1%)	$V(STEP)pp = 1 V$, $AV = -1$, Rise $C_L = 50 pF$, $R_L = 100 k\Omega$ Fall	25°C	44	μs

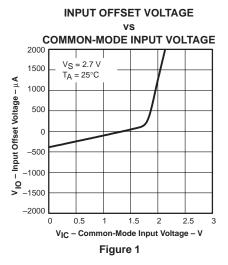
noise/distortion performance

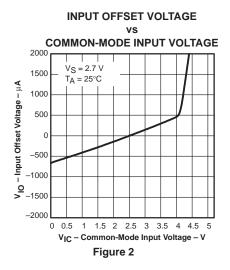
	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
Vn	Equivalent input noise voltage	f = 1 kHz	25°C		89		nV/√Hz
In	Equivalent input noise current	f = 1 kHz	25°C		0.6		fA/√Hz

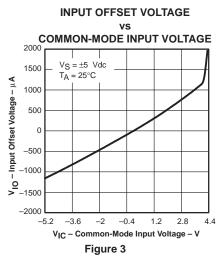


Table of Graphs

			FIGURE
VIO	Input offset voltage	vs Common-mode input voltage	1, 2, 3
I _{IB} /I _{IO}	Input bias and offset current	vs Free-air temperature	4
Vон	High-level output voltage	vs High-level output current	5, 7, 9
VOL	Low-level output voltage	vs Low-level output current	6, 8, 10
	0:	vs Supply voltage	11
lQ	Quiescent current	vs Free-air temperature	12
	Supply voltage and supply current ramp up		13
A _{VD}	Differential voltage gain and phase shift	vs Frequency	14
GBP	Gain-bandwidth product	vs Free-air temperature	15
фm	Phase margin	vs Load capacitance	16
CMRR	Common-mode rejection ratio	vs Frequency	17
PSRR	Power supply rejection ratio	vs Frequency	18
	Input referred noise voltage	vs Frequency	19
SR	Slew rate	vs Free-air temperature	20
VO(PP)	Peak-to-peak output voltage	vs Frequency	21
·	Inverting small-signal response		22
	Inverting large-signal response		23
	Crosstalk	vs Frequency	24







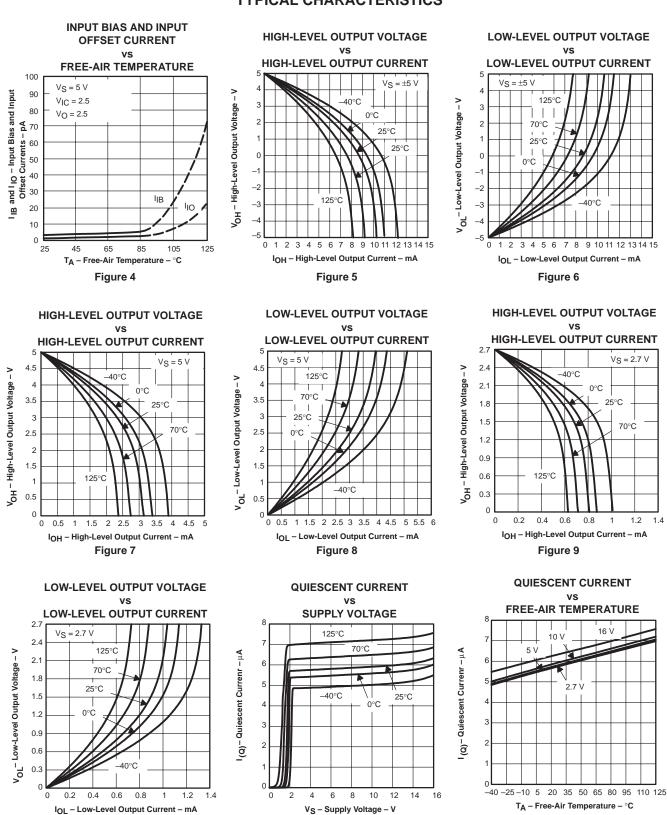
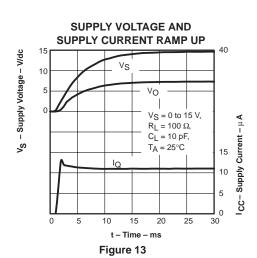




Figure 11

Figure 12

Figure 10



AND PHASE SHIFT FREQUENCY 120 V_S = 5 V $R_L = 100 \text{ k}\Omega$ $C_L = 10 \text{ pF}$ A _{VD} – Differential Voltage Gain – dB 100 0° T_A = 25°C 80 30° Shift 60 60° 40 90° 20 120° 0 150° -20 180° 0.1 1 10 100 1 k 10 k 100 k 1 M f - Frequency - Hz

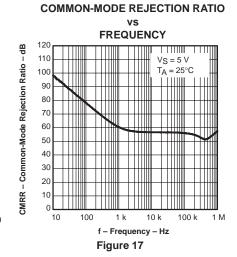
Figure 14

DIFFERENTIAL VOLTAGE GAIN

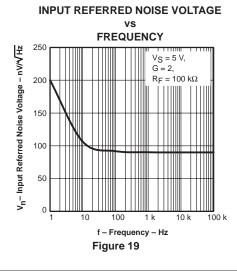
GAIN-BANDWIDTH PRODUCT vs FREE-AIR TEMPERATURE 170 GBP - Gain-Bandwidth Product - kHz 160 $V_S = 2.7 V$ 150 V_S = 5 V 140 130 120 110 100 -40 -25 -10 5 20 35 50 65 80 95 110 125 T_A – Free-Air Temperature – $^{\circ}C$ Figure 15

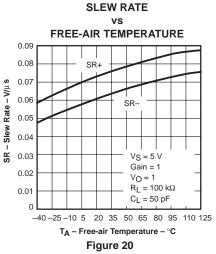
POWER SUPPLY REJECTION RATIO

PHASE MARGIN VS LOAD CAPACITANCE 80 V_S = 5 V 70 $R_L = 100 \text{ k}\Omega$ T_A = 25°C 60 Phase Margin – Degrees 50 40 30 20 10 0 100 1000 10 C_L - Load Capacitance - pF Figure 16

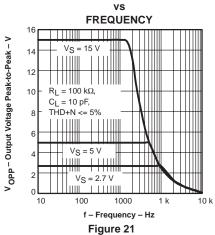


vs **FREQUENCY** V_S =±2.5 V 100 90 PSRR - Power Supply Rejection Ratio -T_A = 25°C 80 70 60 50 40 30 20 0 10 k 100 k 10 f – Frequency – Hz Figure 18





PEAK-TO-PEAK OUTPUT VOLTAGE



INVERTING SMALL-SIGNAL RESPONSE

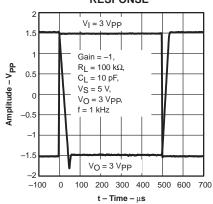


Figure 22

INVERTING LARGE-SIGNAL RESPONSE

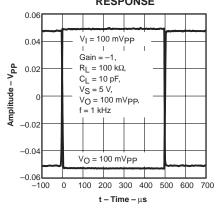


Figure 23

CROSSTALK

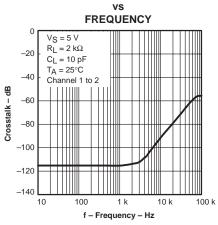


Figure 24



APPLICATION INFORMATION

offset voltage

The output offset voltage (V_{OO}) is the sum of the input offset voltage (V_{IO}) and both input bias currents (I_{IB}) times the corresponding gains. The following schematic and formula can be used to calculate the output offset voltage:

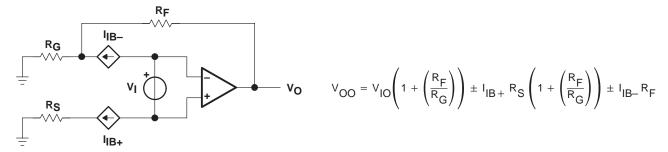


Figure 25. Output Offset Voltage Model

general configurations

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to accomplish this is to place an RC filter at the noninverting terminal of the amplifier (see Figure 26).

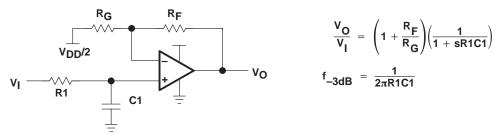


Figure 26. Single-Pole Low-Pass Filter

If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to do this can result in phase shift of the amplifier.

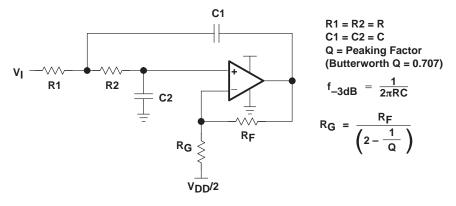


Figure 27. 2-Pole Low-Pass Sallen-Key Filter





APPLICATION INFORMATION

circuit layout considerations

To achieve the levels of high performance of the TLV27Lx, follow proper printed-circuit board design techniques. A general set of guidelines is given in the following.

- Ground planes—It is highly recommended that a ground plane be used on the board to provide all
 components with a low inductive ground connection. However, in the areas of the amplifier inputs and
 output, the ground plane can be removed to minimize the stray capacitance.
- Proper power supply decoupling—Use a 6.8-μF tantalum capacitor in parallel with a 0.1-μF ceramic capacitor on each supply terminal. It may be possible to share the tantalum among several amplifiers depending on the application, but a 0.1-μF ceramic capacitor should always be used on the supply terminal of every amplifier. In addition, the 0.1-μF capacitor should be placed as close as possible to the supply terminal. As this distance increases, the inductance in the connecting trace makes the capacitor less effective. The designer should strive for distances of less than 0.1 inches between the device power terminals and the ceramic capacitors.
- Sockets—Sockets can be used but are not recommended. The additional lead inductance in the socket pins
 will often lead to stability problems. Surface-mount packages soldered directly to the printed-circuit board
 is the best implementation.
- Short trace runs/compact part placements—Optimum high performance is achieved when stray series inductance has been minimized. To realize this, the circuit layout should be made as compact as possible, thereby minimizing the length of all trace runs. Particular attention should be paid to the inverting input of the amplifier. Its length should be kept as short as possible. This will help to minimize stray capacitance at the input of the amplifier.
- Surface-mount passive components—Using surface-mount passive components is recommended for high
 performance amplifier circuits for several reasons. First, because of the extremely low lead inductance of
 surface-mount components, the problem with stray series inductance is greatly reduced. Second, the small
 size of surface-mount components naturally leads to a more compact layout thereby minimizing both stray
 inductance and capacitance. If leaded components are used, it is recommended that the lead lengths be
 kept as short as possible.



APPLICATION INFORMATION

general power dissipation considerations

For a given θ_{JA} , the maximum power dissipation is shown in Figure 28 and is calculated by the following formula:

$$P_{D} = \left(\frac{T_{MAX}^{-T}A}{\theta_{JA}}\right)$$

Where:

P_D = Maximum power dissipation of TLV27Lx IC (watts)

T_{MAX} = Absolute maximum junction temperature (150°C)

 T_A = Free-ambient air temperature (°C)

 $\theta_{JA} = \theta_{JC} + \theta_{CA}$

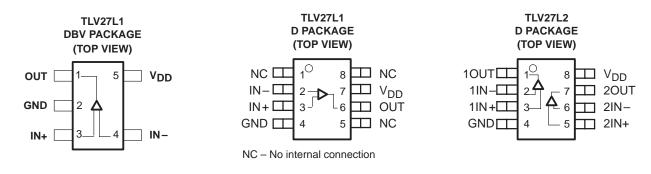
 θ_{JC} = Thermal coefficient from junction to case

 θ_{CA} = Thermal coefficient from case to ambient air (°C/W)

MAXIMUM POWER DISSIPATION FREE-AIR TEMPERATURE 2 TJ = 150°C PDIP Package Low-K Test PCB 1.75 θ_{JA} = 104°C/W Maximum Power Dissipation - W 1.5 MSOP Package Low-K Test PCB SOIC Package θ_{JA} = 260°C/W 1.25 Low-K Test PCB მ**J**Ā = 176°C/W 0.75 0.5 0.25 SOT-23 Package Low-K Test PCB θ_{JA} = 324°C/W -55-40-25-10 5 20 35 50 65 80 95 110 125 T_A - Free-Air Temperature - °C

NOTE A: Results are with no air flow and using JEDEC Standard Low-K test PCB.

Figure 28. Maximum Power Dissipation vs Free-Air Temperature





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