

# STD100NH03L

# N-CHANNEL 30V - 0.005 Ω - 60A DPAK STripFET<sup>TM</sup> III POWER MOSFET

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	I <sub>D</sub>
STD100NH03L	30 V	< 0.0055 Ω	60 A(2)

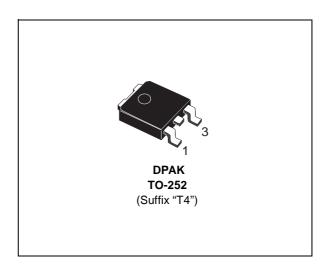
- TYPICAL  $R_{DS}(on) = 0.005 \Omega$  @ 10 V
- R<sub>DS(ON)</sub> \* Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- SURFACE-MOUNTING DPAK (TO-252) POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")

### **DESCRIPTION**

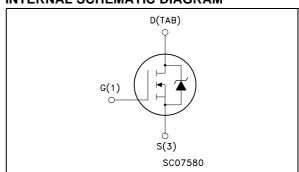
The **STD100NH03L** utilizes the latest advanced design rules of ST's proprietary STripFET™ technology. This is suitable fot the most demanding DC-DC converter application where high efficiency is to be achieved.

### **APPLICATIONS**

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTES



### INTERNAL SCHEMATIC DIAGRAM



### **Ordering Information**

SALES TYPE	MARKING	PACKAGE	PACKAGING
STD100NH03LT4	D100NH03L	TO-252	TAPE & REEL

### **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	30	V
V <sub>DGR</sub>	Drain-gate Voltage (R <sub>GS</sub> = 20 k $\Omega$ )	30	V
V <sub>GS</sub>	Gate- source Voltage	± 20	V
I <sub>D</sub> (2)	Drain Current (continuous) at T <sub>C</sub> = 25°C	60	А
I <sub>D</sub> (2)	Drain Current (continuous) at T <sub>C</sub> = 100°C	60	А
I <sub>DM</sub> (3)	Drain Current (pulsed)	240	А
P <sub>tot</sub>	Total Dissipation at T <sub>C</sub> = 25°C	100	W
	Derating Factor	0.66	W/°C
E <sub>AS</sub> (4)	Single Pulse Avalanche Energy	700	mJ
T <sub>stg</sub>	Storage Temperature	-55 to 175	°C
Tj	Max. Operating Junction Temperature	-55 to 175	

July 2003 1/11

### STD100NH03L

### THERMAL DATA

Rthj-case Rthj-amb Rthj-pcb	Thermal Resistance Junction-case Thermal Resistance Junction-ambient Thermal Resistance Junction-pcb(#)	Max Max Max	1.5 100 43	°C/W °C/W
Ťi	Maximum Lead Temperature For Soldering Purpose		275	°C

<sup>(#)</sup> When Mounted on 1 inch<sup>2</sup> FR-4 board, 2 oz of Cu.

# **ELECTRICAL CHARACTERISTICS** ( $T_{CASE} = 25~^{\circ}C$ UNLESS OTHERWISE SPECIFIED) OFF

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	$I_D = 250 \ \mu A, \ V_{GS} = 0$	30			V
I <sub>DSS</sub>	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	$V_{DS} = Max Rating$ $V_{DS} = Max Rating T_C = 125^{\circ}C$			1 10	μA μA
IGSS	Gate-body Leakage Current (V <sub>DS</sub> = 0)	$V_{GS} = \pm 20V$			±100	nA

# ON (\*)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
V <sub>GS(th)</sub>	Gate Threshold Voltage	$V_{DS} = V_{GS}$	I <sub>D</sub> = 250 μA	1	1.8	2.5	V
R <sub>DS(on)</sub>	Static Drain-source On Resistance	V <sub>GS</sub> = 10 V V <sub>GS</sub> = 5 V	I <sub>D</sub> = 30 A I <sub>D</sub> = 30 A		0.005 0.0060	0.0055 0.0105	Ω Ω

### **DYNAMIC**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
gfs (5)	Forward Transconductance	$V_{DS} = 10 \text{ V}$ $I_{D} = 30 \text{ A}$		40		S
C <sub>iss</sub> C <sub>oss</sub> C <sub>rss</sub>	Input Capacitance Output Capacitance Reverse Transfer Capacitance	$V_{DS} = 15V f = 1 MHz V_{GS} = 0$		4100 680 70		pF pF pF
R <sub>G</sub>	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.3		Ω

### **ELECTRICAL CHARACTERISTICS** (continued)

### **SWITCHING ON**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(on)</sub> t <sub>r</sub>	Turn-on Delay Time Rise Time	$\begin{aligned} V_{DD} &= 15 \text{ V} & I_D &= 30 \text{ A} \\ R_G &= 4.7 \ \Omega & V_{GS} &= 10 \text{ V} \\ \text{(Resistive Load, Figure 3)} \end{aligned}$		16 95		ns ns
Q <sub>g</sub> Q <sub>gs</sub> Q <sub>gd</sub>	Total Gate Charge Source Gate Charge Gate-Drain Charge	V <sub>DD</sub> = 15 V I <sub>D</sub> = 60 A V <sub>GS</sub> = 10 V		57 11.8 7.3	77	nC nC nC
Q <sub>oss</sub> (6)	Output Charge	V <sub>DS</sub> = 16 V V <sub>GS</sub> = 0 V		27		nC
Q <sub>gls</sub> (7)	Third-quadrant Gate Charge	V <sub>DS</sub> < 0 V V <sub>GS</sub> = 10 V		55		nC

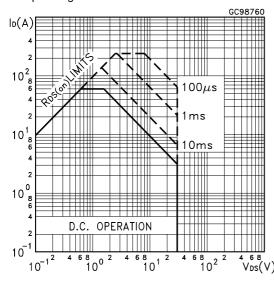
### **SWITCHING OFF**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(off)</sub>	Turn-off Delay Time Fall Time	$\begin{array}{ccc} V_{DD} = 15 \text{ V} & I_{D} = 30 \text{ A} \\ R_{G} = 4.7\Omega, & V_{GS} = 10 \text{ V} \\ \text{(Resistive Load, Figure 3)} \end{array}$		48 23		ns ns

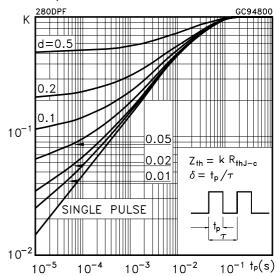
### **SOURCE DRAIN DIODE**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I <sub>SD</sub>	Source-drain Current Source-drain Current (pulsed)				60 240	A A
V <sub>SD</sub> (5)	Forward On Voltage	I <sub>SD</sub> = 30 A V <sub>GS</sub> = 0			1.4	V
t <sub>rr</sub> Q <sub>rr</sub> I <sub>RRM</sub>	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 60 \text{ A}$ di/dt = $100 \text{A}/\mu \text{s}$ $V_{DD} = 30 \text{ V}$ $T_j = 150 ^{\circ} \text{C}$ (see test circuit, Figure 5)		46 64 2.8	62 86	ns nC A

### Safe Operating Area



### Thermal Impedance

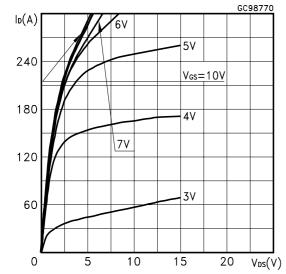


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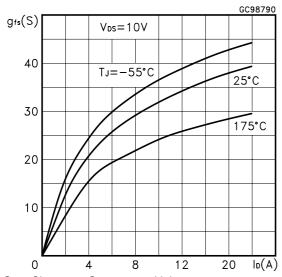
<sup>(2)</sup> Value limited by wire bonding  $^{(3)}$  Pulse width limited by safe operating area. (4) Starting  $T_j=25$  °C,  $I_D=30\text{A},\,V_{DD}=15\text{V}$ 

<sup>(5)</sup> Pulsed: Pulse duration = 300  $\mu$ s, duty cycle 1.5 %. (6)  $Q_{OSS} = C_{OSS}^* \Delta \ V_{in}$ ,  $C_{OSS} = C_{gd} + C_{ds}$ . See Appendix A (7) Gate charge for synchronous operation

### **Output Characteristics**

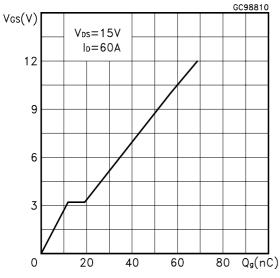


### Transconductance

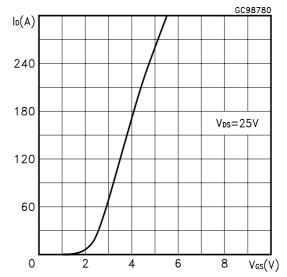


Gate Charge vs Gate-source Voltage

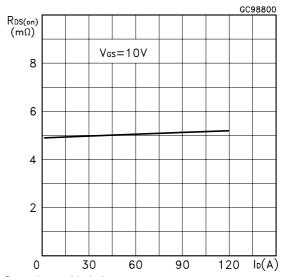
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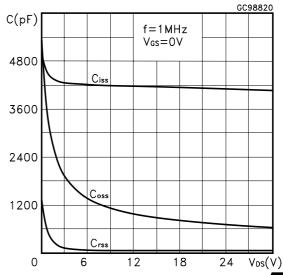
### **Transfer Characteristics**



Static Drain-source On Resistance

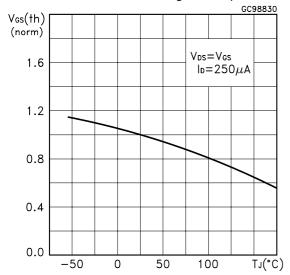


Capacitance Variations

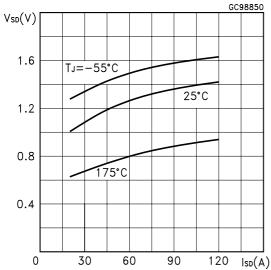


### STD100NH03L

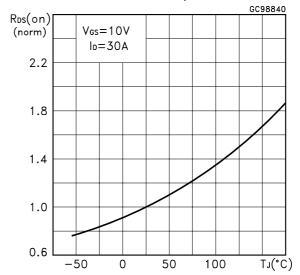
### Normalized Gate Threshold Voltage vs Temperature



### Source-drain Diode Forward Characteristics



### Normalized on Resistance vs Temperature



### Normalized Breakdown Voltage vs Temperature

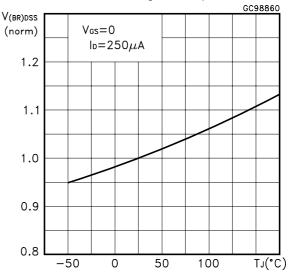


Fig. 1: Unclamped Inductive Load Test Circuit

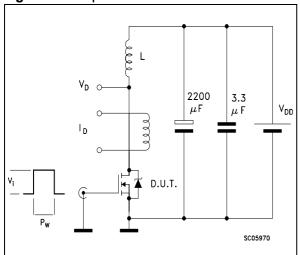
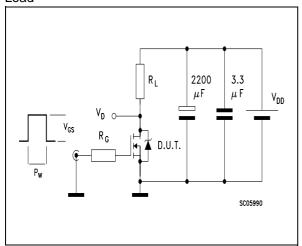


Fig. 3: Switching Times Test Circuits For Resistive Load



**Fig. 5:** Test Circuit For Inductive Load Switching And Diode Recovery Times

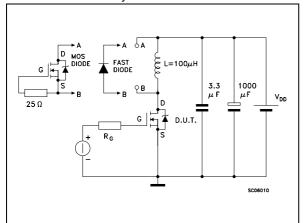


Fig. 2: Unclamped Inductive Waveform

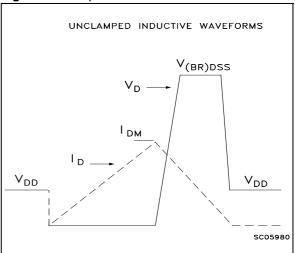
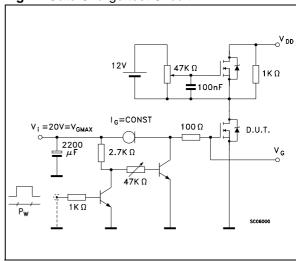
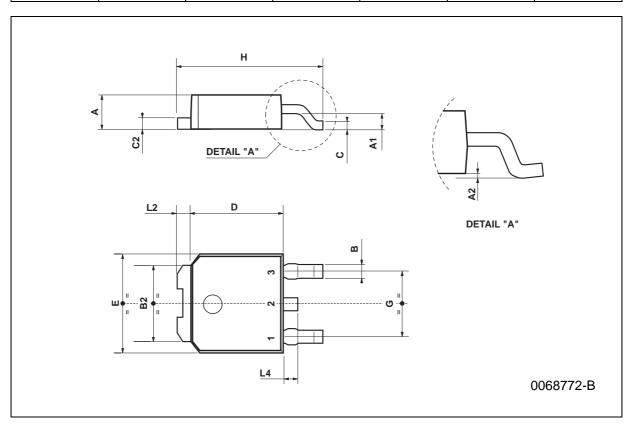


Fig. 4: Gate Charge test Circuit



# **TO-252 (DPAK) MECHANICAL DATA**

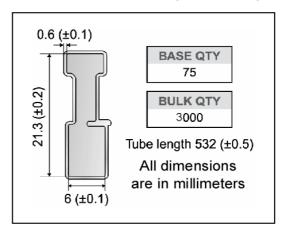
DIM.		mm			inch	
Diwi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A2	0.03		0.23	0.001		0.009
В	0.64		0.9	0.025		0.035
B2	5.2		5.4	0.204		0.212
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
Е	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	9.35		10.1	0.368		0.397
L2		0.8			0.031	
L4	0.6		1	0.023		0.039



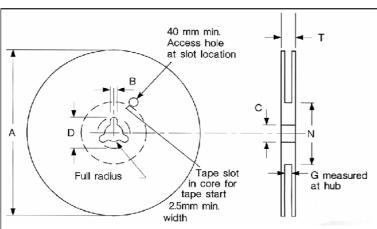
### **DPAK FOOTPRINT**

# 6.7 1.8 3.0 1.6 2.3 1.6 All dimensions are in millimeters

# **TUBE SHIPMENT (no suffix)\***



## TAPE AND REEL SHIPMENT (suffix "T4")\*

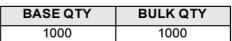


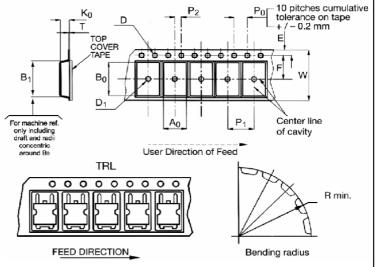
### REEL MECHANICAL DATA

рім.	mm		ine	ch
	MIN.	MAX.	MIN.	MAX.
Α		330		12.992
В	1.5		0.059	
С	12.8	13.2	0.504	0.520
D	20.2		0.795	
G	16.4	18.4	0.645	0.724
N	50		1.968	
Т		22.4		0.881

TAPE	MECH/	MICAL	DATA
			$\nu$

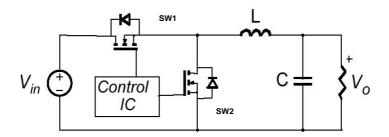
DIM.	mm		inch	
Dilvi.	MIN.	MAX.	MIN.	MAX.
A0	6.8	7	0.267	0.275
В0	10.4	10.6	0.409	0.417
B1		12.1		0.476
D	1.5	1.6	0.059	0.063
D1	1.5		0.059	
E	1.65	1.85	0.065	0.073
F	7.4	7.6	0.291	0.299
K0	2.55	2.75	0.100	0.108
P0	3.9	4.1	0.153	0.161
P1	7.9	8.1	0.311	0.319
P2	1.9	2.1	0.075	0.082
R	40		1.574	
W	15.7	16.3	0.618	0.641
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# **APPENDIX A Buck Converter: Power Losses Estimation**



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is converted to allow for a safer working junction temperature.

The low side (SW2) device requires:

- ullet Very low  $R_{DS(on)}$  to reduce conduction losses
- $\bullet \qquad Small \ Q_{gls} \ to \ reduce \ the \ gate \ charge \ losses$
- Small Coss to reduce losses due to output capacitance
- Small Q<sub>rr</sub> to reduce losses on SW<sub>1</sub> during its turn-on
- The  $C_{gd}/C_{gs}$  ratio lower than  $V_{th}/V_{gg}$  ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

- ullet Small  $R_g$  and  $L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate
- Small Q<sub>g</sub> to have a faster commutation and to reduce gate charge losses
- Low R<sub>DS(on)</sub> to reduce the conduction losses.

		High Side Switch (SW1)	Low Side Switch (SW2)
Pconduct	ion	$R_{DS(on)SW1} * I_L^2 * d$	$R_{DS(on)SW2} * I_L^2 * (1-d)$
Pswitchin	g	$V_{\text{in}} * (Q_{\text{gsth(SW1)}} + Q_{\text{gd(SW1)}}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
$P_{\text{diode}}$	Recovery	Not Applicable	<sup>1</sup> V <sub>in</sub> *Q <sub>rr(SW2)</sub> * f
	Conduction	Not Applicable	$V_{\text{f(SW2)}} * I_{\text{L}} * t_{\text{deadtime}} * f$
Pgate(QG	)	$Q_{g(SW1)} * V_{gg} * f$	$Q_{gls(SW2)}*V_{gg}*f$
P <sub>Qoss</sub>		$\frac{V_{in} *Q_{oss(SW1)} *f}{2}$	$\frac{V_{\text{in}} * Q_{\text{oss(SW2)}} * f}{2}$

Parameter	Meaning
d	Duty-cycle
Qgsth	Post threshold gate charge
$Q_{ m gls}$	Third quadrant gate charge
Pconduction	On state losses
Pswitching	On-off transition losses
Pdiode	Conduction and reverse recovery diode losses
Pgate	Gate drive losses
P <sub>Qoss</sub>	Output capacitance losses

<sup>&</sup>lt;sup>1</sup> Dissipated by SW1 during turn-on

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